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Adaptive consensus tracking control of non-affine non-linear MASs based on Taylor decoupling technology and an event-triggered design strategy

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This research paper studies the consensus tracking control problem for a class of uncertain non-affine non-linear multi-agent systems (MASs). First, different from the separation design scheme using the mean value theorem in previous works, this research paper not only uses the mean value theorem but also introduces the Taylor decoupling method to decouple the complex unknown non-affine structure. Second, to solve the difficulty of unknown non-linear functions in non-linear MASs, an intelligent technique based on neural networks was used. In addition, compared with the existing traditional event-triggered control strategy based on the relative threshold, an improved event-triggered control strategy based on the decreasing function of error variables was introduced to reduce the waste of unnecessary resources. The theoretical result shows that the whole closed-loop system is stable under the action of the proposed control protocol. Finally, the simulation experiment verifies the effectiveness of our control method.

KEYWORDS

adaptive control, consensus tracking control, non-affine non-linear multi-agent system, Taylor decoupling, event-triggered strategy control

1 Introduction

In recent years, with the rapid development of computer technology, problems related to MASs have also been the focus of many scholars. The system is mainly used in the fields of robotics, transportation, and human-machine interactions [1–7]. Particularly, MASs have higher performance and efficiency compared with expensive single systems; however, their control is more complex. The large-scale complex control problem of MASs can be solved through information exchange and coordination among agents. One of the most significant and essential areas of study in MAS cooperative control is the consensus problem. Early studies conducted extensive research on the consensus tracking control of linear MASs [8–10]. However, in recent years, consensus tracking control of non-linear MASs has received increasing attention [11,12].

In real-world industrial production, many objects cannot be modeled as systems with affine forms; therefore, the control design of non-affine non-linear systems has always been a key problem [13–17]. Furthermore, due to the needs of some practical tasks, such as supersonic vehicles and magnetic levitation systems [18–20], theoretical research on non-affine non-linear MASs is more meaningful and some non-affine non-linear MAS control

methods have been proposed. Using a new class of implicit function and fuzzy logic technology, under the condition of switching topologies, the containment control problem of uncertain nonaffine non-linear MASs with many dynamic leaders has been addressed [21]. Regarding the control problem of non-affine non-linear MASs, Wang and Song [22] proposed a distributed neural adaptive control scheme under the condition that the control gain is uneven. The aforementioned research showed that the implicit function or median theorems are widely used for controller decoupling. In contrast, the Taylor method used in the present study provides a new approach for controller decoupling.

The previous literature has shown that the event-triggered control (ETC) strategy is a good way to reduce sample data and traffic to design control strategies. In recent years, many researchers have adopted the ETC strategy to design control strategies [23–28]. An ETC strategy that follows the switching threshold was introduced to save communication resources, and the tracking control problem for stochastic non-linear pure-feedback MASs was solved [29]. Wu et al [30] proposed an improved ETC strategy that included ETC input and tracking error reduction function to update the actual control input. However, the aforementioned event-triggered strategies do not take into account the triggering rate, which is worth considering in the development of more efficient ETC strategies, and which motivates our work.

Based on the aforementioned findings, this research paper focuses on the consensus tracking control problem for non-affine non-linear MASs. According to the Taylor decoupling technique, a scheme of control input separation design for non-affine non-linear MASs is proposed to ensure the boundedness of all signals and achieve good consensus tracking. By introducing an improved ETC strategy, unnecessary resource waste is reduced. The following is a summary of the contributions made by this research paper: 1) to solve the coupling problem of non-affine non-linear MASs, the Taylor decoupling technology was used to effectively decouple the non-linear coupling functions. In addition, an intelligent technology based on neural networks was used to approximate unknown nonlinear functions. 2) The previous literature used the fixed threshold ETC strategy to change the size of the control amplitude, with a constant measurement error [31]; in contrast, the relative threshold ETC considered in this study can adjust the system performance more flexibly. This research paper adopts an improved relative threshold ETC strategy to design the controller for each agent and introduces a decreasing function of error variables, which improves the efficiency of the ETC strategy by reducing the waste of communication resources.

2 Problem formulation and preliminaries

2.1 Graph theory

Consider the topological structure of a MASs with one leader and multiple followers, which is represented by a $\hat{G} = (\hat{\mathcal{V}}, \hat{\mathcal{E}})$ with $\hat{\mathcal{V}} = \{v_0, v_1, \dots, v_N\}$ representing the node set, where v_0 is an agent associated with the leader, and $\hat{\mathcal{E}} = \hat{\mathcal{V}} \times \hat{\mathcal{V}}$ denoting the edge set. An edge $(i, j) \in \hat{\mathcal{V}}$ in \hat{G} means that the agent *i* can get information from the agent *j* directly. The adjacency matrix is denoted as $A = [a_{ij}] \in \mathbb{R}^{N \times N}$ with $a_{ij} > 0$. The set of neighbors of node *i* is denoted by $N_i = \{j = (j,i) \in \hat{\mathcal{E}}\}$. The diagonal matrix $D = diag(d_1, d_2, \ldots, d_N) \in \mathbb{R}^{N \times N}$ is the definition of the in-degree matrix, where $\tilde{d}_i = \sum_{j \in N_i} a_{ij}$. The Laplacian matrix is defined as L = D - A, where $L \in \mathbb{R}^{N \times N}$.

2.2 System formulation

We consider the following class of non-affine non-linear MASs:

$$\dot{x}_{i,k} = g_{i,k}(\Delta_{i,k})x_{i,k+1} + f_{i,k}(\Delta_{i,k}) + \varphi_{i,k}^{T}(\Delta_{i,k})\eta_{i,k}, \dot{x}_{i,n_{i}} = f_{i,n_{i}}(\Delta_{i}, u_{i}) + \varphi_{i,n_{i}}^{T}(\Delta_{i})\eta_{i,n_{i}} + d_{i}(t), y_{i} = x_{i,1},$$
(1)

where $\Delta_{ik} = [x_{i,1}, x_{i,2}, \dots, x_{ik}]^T \in \mathbb{R}^k, \Delta_i = [x_{i,1}, x_{i,2}, \dots, x_{in_i}]^T \in \mathbb{R}^{n_i}$ are the system state vectors. $y_i \in \mathbb{R}$, $u_i \in \mathbb{R}$, $d_i \in \mathbb{R}$ are the control output, the input, and the additive disturbance, $f_{i,k}(\cdot)$; $g_{i,k}(\cdot)$: $\mathbb{R}^k \to \mathbb{R}$ represents the known smooth functions, f_{in_i} ; and $\varphi_{i,k}(\cdot)$ represents the unknown smooth functions. $\eta_{i,k} \in \mathbb{R}^p$ denotes the unknown parameter vector.

Our goal is to ensure that: 1) all signals in the closed-loop system fall within the specified compact set; and 2) the system output tracking error $e_1 = y - y_d$ converges to zero.

Assumption 1: The external disturbance d_i , the reference signal y_d , and its *k*th-order derivatives $y_d^{(k)}$, k = 1, 2, ..., n, are all continuous and bounded. In addition, $|y_d| \le y_d^*$, $|y_d^{(k)}| \le y_d^{(k)}$, and $|d_i| \le d_i^*$, where $y_d^*, y_d^{(k)^*}$ and d_i^* are the unknown upper bounds.

Assumption 2: \hat{G} contains a spanning tree, the root which is called the leader y_d .

Assumption 3 [30]: Based on Assumption 1, for a given compact set $\Omega_{\Delta} \in \mathbb{R}^n$, there exist two positive constants f_a^* and f_b^* such that this research paper deals with a class of non-affine non-linear MASs tracking control systems with uncertainties

$$0 \le f_a^* \le \frac{\partial f_{i,n_i}(\Delta, 0)}{\partial u} \le f_b^*, \tag{2}$$

where arbitrary $\Delta \in \Omega_{\Delta}$.

2.3 Preliminaries

Lemma1: Let Ω_{Δ} be given compact set of \mathbb{R}^{n_i} , then the nonlinear coupling function $f_{i,n_i}(\Delta_i, u_i)$ can be changed into

$$f_{i,n_{i}}(\Delta_{i}, u_{i}) = f_{i,n_{i}}(\Delta_{i}, 0) + g_{i,n_{i}}(\Delta_{i}, u_{i})u_{i}.$$
(3)

Then, we use Taylor's theorem to separate u_i from g_{i,n_i}

$$g_{i,n_i}(\Delta_i, u_i) = g_{i,n_i}(\Delta_i, 0) + \frac{\partial g_{i,n_i}(\Delta_i, 0)}{\partial u_i} u_i + \frac{1}{2!} \frac{\partial^2 g_{i,n_i}(\Delta_i, 0)}{\partial u_i^2} u_i^2 + \cdots + \frac{1}{n_i!} \frac{\partial^n g_{i,n_i}(\Delta_i, 0)}{\partial u_i^n} u_i^n + \Re_{i,n_i+1}(\Delta_i, u_i),$$
(4)

where $g_{i,n_i}(\Delta_i, u_i) = \left(\frac{\partial f(\Delta_i, u_i)}{\partial u_i}\right|_{u=u_c}$ with $u_c = cu, c \in (0, 1)$ and $\Re_{i,n_i+1}(\Delta_i, u_i) = \frac{1}{(n_i+1)!} \frac{\partial^{n+1}g_{n_i}(\Delta_i, \zeta)}{\partial u_i^{n+1}} u_i^{n+1}$ with $0 < \zeta < u_i$. Substituting Eq. 4 into Eq. 3, we obtain

$$f_{i,n_i}(\Delta_i, u_i) = f_{i,n_i}(\Delta_i, 0) + g_{i,n_i}(\Delta_i, 0)u_i + m_{i,n_i}(\Delta_i, u_i)$$

= $g_{i,n_i}(\Delta_i, 0)u_i + C(\Delta_i),$ (5)

where $C(\Delta_i) = f_{i,n_i}(\Delta_i, 0) + m_{i,n_i}(\Delta_i, u_i), m_{i,n_i}(\Delta_i, u) = \frac{\partial g_{i,n_i}(\Delta_i, 0)}{\partial u_i} u_i^2 + \frac{1}{2!} \frac{\partial^2 g_{i,n_i}(\Delta_i, 0)}{\partial u_i^2} u_i^3 + \dots + \frac{1}{n_!!} \frac{\partial^2 g_{i,n_i}(\Delta_i, 0)}{\partial u_i^n} u_i^{n+1} + \Re_{i,n_i+1}(\Delta_i, u_i) u_i.$

Therefore, from Eqs 3–5, Eq. 1 can be rewritten in the following affine form:

$$\dot{x}_{i,k} = g_{i,k}(x_i)x_{i,k+1} + f_{i,k}(x_i) + \varphi_{i,k}^{T}(x_i)\eta_{i,k}, \dot{x}_{i,n_i} = g_{i,n_i}(\Delta_i, 0)u_i + C(\Delta_i) + \varphi_{i,n_i}^{T}(x_i)\eta_{i,n_i} + d_i(t),$$

$$y_i = x_{i,1}.$$
(6)

Lemma 2 [32]: Define the diagonal matrix $\tilde{B} = diag\{\tilde{b}_i\} \in \mathbb{R}^{N \times N}$, then $L + \tilde{B}$ is non-singular.Lemma 3 [32]: Define $E_1 = (e_{1,1}, e_{2,1}, \dots, e_{N,1})^T$, $Y = (y_1, y_2, \dots, y_N)^T$, $Y_c = (y_c, y_c, \dots, y_c)^T$, then

$$||Y - Y_c|| \le ||E_1|| \beta (L + \tilde{B}),$$
 (7)

where $\beta(L + \tilde{B})$ is the minimum singular value of $L + \tilde{B}$. Lemma 4 [32]: For any constant $\alpha \in \mathbb{R}$ and any variable $\varepsilon > 0$, the following inequality holds:

$$0 \le |\alpha| - \alpha \tanh\left(\frac{\alpha}{\varepsilon}\right) \le \kappa\varepsilon,\tag{8}$$

where $\kappa = 0.2785$.

2.4 Radial basis function neural networks

Radial basis function neural networks (RBFNNs) can approximate arbitrary non-linear functions [11,33–35]. Specifically, the unknown non-linear functions $F(\Gamma)$ can be approximated over a compact set $\Gamma \subset \Omega_{\Gamma} \subset \mathbb{R}^{l}$

$$F(\Gamma) = \Phi^{*T} \overline{S}(\Gamma) + \delta(\Gamma), \qquad (9)$$

where $\Phi^* = [\Phi_1, \Phi_2, \dots, \Phi_l]^T \in \mathbb{R}^l$ is the ideal weight vector, $\delta(\Gamma)$ is the approximation error satisfying $|\delta(\Gamma)| \leq \tau$ with a precision level $\tau > 0$. $\bar{S}(\Gamma) = [\bar{S}_1(\Gamma), \bar{S}_2(\Gamma), \dots, \bar{S}_l(\Gamma)]^T \in \mathbb{R}^l$ is the basis function, where l > 1 is the node number of s RBFNNs. Particularly, the basis function can be chosen as

$$\bar{S}_{i}(\Gamma) = \exp\left[-(\Gamma - \xi_{i})^{T}(\Gamma - \xi_{i})/\eta_{i}^{2}\right], \quad i = 1, \dots, l, \qquad (10)$$

where $\xi_i = [\xi_{i1}, \dots, \xi_{il}]^T$ is the center of the receptive field center and η_i is the width of the Gaussian function.

3 Main result

This section provides an efficient adaptive ETC strategy based on the adaptive neural approximation technique and a backstepping scheme.

The following error variables are defined:

$$e_{i,1} = \sum_{j=1}^{N} a_{ij} (y_i - y_j) + \tilde{b}_i (y_i - y_d), \qquad (11)$$

$$e_{i,k} = x_{i,k} - u_{i,k-1}, \tag{12}$$

where $u_{i,k-1}$ is the virtual controller designed in step k.

Step 1: First, the derivation of $e_{i,1}$ along (Eq. 11) is

$$\dot{e}_{i,1} = \left(\tilde{b}_i + \tilde{d}_i\right) \left(g_{i,1}\left(x_i\right) x_{i,1} + f_{i,1}\left(x_i\right) + \varphi_{i,1}^T\left(x_i\right) \eta_{i,1}\right) \\ -\tilde{b}_i \dot{y}_d - \sum_{j=1}^N a_{ij} \left(g_{j,1}\left(x_j\right) x_{j,1} + f_{j,1}\left(x_i\right) + \varphi_{j,1}^T\left(x_j\right) \eta_{j,1}\right).$$
(13)

The Lyapunov function is

$$V_{i,1} = \frac{1}{2}e_{i,1}^2 + \frac{1}{2\gamma_{i,1}}\tilde{\theta}_{i,1}^{2}, \qquad (14)$$

where $\gamma_{i,1}$ is a positive design parameter, $\hat{\theta}_{i,1}$ is the estimation of $\theta_{i,1}$, and $\tilde{\theta}_{i,1} = \theta_{i,1} - \hat{\theta}_{i,1}$.

From Eqs 13, 14, the derivative of $V_{i,1}$ is computed as

$$\dot{V}_{i,1} = e_{i,1}\dot{e}_{i,1} - \frac{\theta_{i,1}}{\gamma_{i,1}}\dot{\hat{\theta}}_{i,1}
= e_{i,1} \Big[\left(\tilde{b}_i + \tilde{d}_i \right) \Big(g_{i,1}(x_i) x_{i,2} + f_{i,1}(x_i) + \varphi_{i,1}^T(x_i) \eta_{i,1} \Big)
- \tilde{b}_i \dot{y}_d - \sum_{j=1}^N a_{ij} \Big(g_{j,1}(x_j) x_{j,2} + f_{j,1}(x_i)
+ \varphi_{j,1}^T(x_j) \eta_{j,1} \Big) \Big] - \frac{\tilde{\theta}_{i,1}}{\gamma_{i,1}} \dot{\hat{\theta}}_{i,1}.$$
(15)

Consequently, taking Eq. 15 into account yields

$$\dot{V}_{i,1} = e_{i,1} \bigg[\left(\tilde{b}_i + \tilde{d}_i \right) \left(g_{i,1} \left(x_i \right) e_{i,2} + g_{i,1} \left(x_i \right) u_{i,1} \right) + \dot{F}_{i,1} \left(\Gamma_i \right) - \frac{e_{i,1}}{2} \bigg] - \frac{\tilde{\theta}_{i,1}}{\gamma_{i,1}} \dot{\hat{\theta}}_{i,1},$$
(16)

where

$$\hat{F}_{i,1}(\Gamma_i) = \left(\tilde{b}_i + \tilde{d}_i\right) \left(f_{i,1}(x_i) + \varphi_{i,1}^T(x_i)\eta_{i,1}\right) - \tilde{b}_i \dot{y}_d
- \tilde{d}_i \left(g_{j,1}(x_j)x_{j,2} + f_{j,1}(x_i)
+ \varphi_{j,1}^T(x_j)\eta_{j,1}\right) + \frac{e_{i,1}}{2}.$$
(17)

Due to $\dot{F}_{i,1}(\Gamma_i)$ contains unknown functions. Hence, the RBFNN is introduced to approximate the unknown functions

$$\dot{F}_{i,1}(\Gamma_{i}) = \Phi_{i,1}^{*T} \bar{S}_{i,1}(\Gamma_{i}) + \delta_{i,1}(\Gamma_{i}), \left|\delta_{i,1}(\Gamma_{i})\right| \le \tau_{i,1},$$
(18)

where $\tau_{i,1} > 0$, $\Gamma_i = [x_{i,1}^T, x_{j,1}^T, y_d, \dot{y}_d]^T \in \Omega$.

Furthermore, combining Lemma 4 with Eq. 18 and Young inequality results in

$$e_{i,1}\acute{F}_{i,1}(\Gamma_i) \le \frac{\theta_{i,1}}{2c_{i,1}^2} e_{i,1}^2 \vec{S}_{i,1}^T(\Gamma_i) \vec{S}_{i,1}(\Gamma_i) + \frac{c_{i,1}^2}{2} + \frac{e_{i,1}^2}{2} + \frac{\tau_{i,1}^2}{2},$$
(19)

where $c_{i,1}$ is a positive constant.

The virtual control $u_{i,1}$ is constructed as

$$u_{i,1} = \frac{1}{\left(\tilde{b}_{i} + \tilde{d}_{i}\right)g_{i,1}(x_{i})} \left[-a_{i,1}e_{i,1} - \frac{\hat{\theta}_{i,1}}{2c^{2}_{i,1}}e_{i,1}\bar{S}_{i,1}^{T}(\Gamma_{i})\bar{S}_{i,1}(\Gamma_{i}) \right], \quad (20)$$

where $a_{i,1}$ is a positive constant.

According to Assumption 3 and Eqs 17-20, we obtain

$$\dot{V}_{i,1} \leq -a_{i,1}e_{i,1}^{2} + \left(\tilde{b}_{i} + \tilde{d}_{i}\right)\left(g_{i,1}\left(x_{i}\right)e_{i,1}e_{i,2}\right) \\
- \frac{\tilde{\theta}_{i,1}}{\gamma_{i,1}}\left(\frac{\gamma_{i,1}}{2c^{2}_{i,1}}e_{i,1}^{2}\bar{S}_{i,1}^{T}\left(\Gamma_{i,1}\right)\bar{S}_{i,1}\left(\Gamma_{i,1}\right) - \dot{\hat{\theta}}_{i,1}\right) \\
+ \frac{c^{2}_{i,1}}{2} + \frac{\tau^{2}_{i,1}}{2}.$$
(21)

Then, the adaptive law $\hat{\theta}_{i,1}$ and the positive design parameters $\mu_{i,1}$ are

$$\dot{\hat{\theta}}_{i,1} = \frac{\gamma_{i,1}}{2c_{i,1}^2} e_{i,1} \bar{S}_{i,1}^T (\Gamma_{i,1}) \bar{S}_{i,1} (\Gamma_{i,1}) - \hat{\theta}_{i,1}, \qquad (22)$$

$$\mu_{i,1} = \frac{c_{i,1}^2}{2} + \frac{\tau_{i,1}^2}{2} + \frac{\theta_{i,1}^2}{2\gamma_{i,1}}.$$
(23)

Substituting Eqs 22, 23 into Eq. 21, we obtain

$$\dot{V}_{i,1} \le -a_{i,1}e_{i,1}^2 + \left(\tilde{b}_i + \tilde{d}_i\right) \left(g_{i,1}\left(x_i\right)e_{i,1}e_{i,2}\right) - \frac{\tilde{\theta}_{i,1}^2}{2\gamma_{i,1}} + \mu_{i,1}.$$
(24)

Step $k(2 \le k \le n_i - 1)$: We choose the Lyapunov function as

$$V_{i,k} = V_{i,k-1} + \frac{1}{2}e_{i,k}^2 + \frac{1}{2\gamma_{i,k}}\tilde{\theta}_{i,k}^2.$$
 (25)

Similar to Eqs 14–17 in Step 1, the derivative of $V_{i,k}$ can be computed as

$$\dot{V}_{i,k} = \dot{V}_{i,k-1} + e_{i,k} \bigg[g_{i,k}(x_i) x_{i,k+1} + f_{i,k}(x_i) + \varphi_{i,k}^T \eta_{i,k} \\ - \sum_{l=1}^{k-1} \sum_{j \in N_j} \frac{\partial u_{j,k-1}}{\partial x_{j,l}} \big(g_{j,l}(x_j) x_{j,l+1} + f_{j,l}(x_j) + \varphi_{j,l}^T \eta_{j,l} \big)$$
(26)
$$- \frac{\partial u_{i,k-1}}{\partial y_d} \dot{y}_d - \sum_{l=1}^{k-1} \frac{\delta u_{i,k-1}}{\delta \hat{\theta}_{i,l}} \dot{\hat{\theta}}_{i,l} \bigg] - \frac{\tilde{\theta}_{i,k}}{\gamma_{i,k}} \dot{\hat{\theta}}_{i,k},$$

where $\gamma_{i,k}$ is an arbitrary constant.

In the same way, as in Eq. 15, we get

$$\dot{V}_{i,k} = \dot{V}_{i,k-1} + e_{i,k} \left[g_{i,k} \left(x_i \right) \left(e_{i,k+1} + u_{i,k} \right) + \dot{F}_{i,k} \left(\Gamma_i \right) - \left(\hat{b}_i + \hat{d}_i \right) g_{i,k-1} e_{i,k-1} - \frac{e_{i,k}}{2} \right] - \frac{\tilde{\theta}_{i,k}}{\gamma_{i,k}} \dot{\hat{\theta}}_{i,k},$$
(27)

where

$$\hat{F}_{i,k}(\Gamma_{i}) = f_{i,k}(x_{i}) + \varphi_{i,k}^{T} \eta_{i,k}
- \sum_{l=1}^{k-1} \sum_{j \in N_{j}} \frac{\partial u_{j,k-1}}{\partial x_{j,l}} (g_{j,l}(x_{j}) x_{j,l+1} + f_{j,l}(x_{j})
+ \varphi_{j,l}^{T} \eta_{j,l}) - \frac{\partial u_{i,k-1}}{\partial y_{d}} \dot{y}_{d} - \sum_{l=1}^{k-1} \frac{\delta u_{i,k-1}}{\delta \hat{\theta}_{i,l}} \dot{\hat{\theta}}_{i,l}
+ \frac{e_{i,k}}{2} + (\dot{b}_{i} + \dot{d}_{i}) g_{i,k-1} e_{i,k-1},$$
(28)

where for k = 2, take $(\tilde{b}_i + \tilde{d}_i) = (\dot{b}_i + \dot{d}_i)$, and for $3 \le k \le n_i - 1$, take $\dot{b}_i + \dot{d}_i = 1$. Similar to Eq. 18, the equation $\dot{F}_{i,k}(\Gamma_i) = \Phi_{i,k}^*(\Gamma_i) \bar{S}_{i,k}(\Gamma_i) + \delta_{i,k}(\Gamma_i), |\delta_{i,k}(\Gamma_i)| \le \tau_{i,k}$ can be obtained easily.

Therefore, we obtain

$$e_{i,k}\dot{F}_{i,k}(\Gamma_i) \le \frac{\theta_{i,k}}{2c_{i,k}^2} e_{i,k}^2 \bar{S}_{i,k}^T(\Gamma_i) \bar{S}_{i,k}(\Gamma_i) + \frac{c_{i,k}^2}{2} + \frac{e_{i,k}^2}{2} + \frac{\tau_{i,k}^2}{2}.$$
 (29)

Designing the virtual control $u_{i,k}$ as

$$u_{i,k} = \frac{1}{g_{i,k}(x_i)} \left[-a_{i,k} e_{i,k} - \frac{\hat{\theta}_{i,k}}{2c^2_{i,k}} e_{i,k} \bar{S}_{i,k}^T(\Gamma_i) \bar{S}_{i,k}(\Gamma_i) \right], \quad (30)$$

where $c_{i,k} > 0$ is the design constant.

We then get

$$\dot{V}_{i,k} \leq -\sum_{l=1}^{k} a_{i,l} e_{i,l}^{2} + g_{i,k} (x_{i}) e_{i,k} e_{i,k+1} - \sum_{l=1}^{k-1} \frac{\tilde{\theta}_{i,l}^{2}}{2y_{i,l}} \\
- \frac{\tilde{\theta}_{i,k}}{\gamma_{i,k}} \left(\frac{\gamma_{i,k}}{2c^{2}{}_{i,k}} e_{i,k}^{2} \bar{S}_{i,k}^{T} (\Gamma_{i,k}) \bar{S}_{i,k} (\Gamma_{i,k}) - \dot{\hat{\theta}}_{i,k} \right) \\
+ \sum_{l=1}^{k-1} \mu_{i,k-1} + \mu_{i,k}.$$
(31)

The adaptive law $\dot{\hat{\theta}}_{i,k}$ and the positive design parameters $\mu_{i,k}$ are designed as

$$\dot{\hat{\theta}}_{i,k} = \frac{\gamma_{i,1}}{2c^2_{i,k}} e_{i,k} \bar{S}^T_{i,k} \left(\Gamma_{i,k} \right) \bar{S}_{i,k} \left(\Gamma_{i,k} \right) - \hat{\theta}_{i,k},$$
(32)

$$\mu_{i,k} = \frac{c_{i,k}^2}{2} + \frac{\tau_{i,k}^2}{2} + \frac{\theta_{i,k}^2}{2\gamma_{i,k}}.$$
(33)

Substituting Eqs 28-33 into Eq. 31 yields

$$\dot{V}_{i,k} \leq -\sum_{l=1}^{k} a_{i,l} e_{i,l}^2 + g_{i,k}(x_i) e_{i,k} e_{i,k+1} - \sum_{l=1}^{k} \frac{\tilde{ heta}_{i,l}^2}{2\gamma_{i,l}} + \sum_{l=1}^{k} \mu_{i,l}.$$

Step n_i : At this step, define $e_{i,n_i} = x_{i,n_i} - u_{i,n_{i-1}}$. We add an unidentified positive constant D such that $|C(x_i) + d_i| \le |C(x_i)| + |d_i| \le D$ for all $\Delta_i \in \Omega_{\Delta}$.

The Lyapunov function is

$$V_{i,n_i} = V_{i,n_i-1} + \frac{1}{2}e_{i,n_i}^2 + \frac{1}{2\gamma_{i,n_i}}\tilde{\theta}_{i,n_i}^2.$$
 (34)

Then,

 $\dot{V}_{i,n_{i}} \leq \dot{V}_{i,n_{i}-1} + e_{i,n_{i}} \Big(g_{i,n_{i}}(x_{i},0)u_{i} + \varphi_{i,n_{i}}^{T}(x_{i})\eta_{i,n_{i}} \\ + D - \dot{u}_{i,n_{i-1}} \Big) - \frac{\tilde{\theta}_{i,n_{i}}}{\gamma_{i,n_{i}}} \hat{\theta}_{i,n_{i}},$ (35)

where

$$\dot{u}_{i,n_{i-1}} = \sum_{l=1}^{n-1} \frac{\partial u_{i,n_{i}-1}}{\partial x_{i,l}} \left(g_{i,l}(x_{i})x_{i,l+1} + f_{i,l}(x_{i}) + \varphi_{i,l}^{T}\eta_{i,l} \right) + \sum_{l=1}^{n-1} \sum_{j \in N_{j}} \frac{\partial u_{j,n_{l}-1}}{\partial x_{j,l}} \left(g_{j,l}(x_{j})x_{j,l+1} + f_{j,l}(x_{j}) \right) + \varphi_{j,l}^{T}\eta_{j,l} + \frac{\partial u_{i,n_{l}-1}}{\partial y_{d}} \dot{y}_{d} + \sum_{l=1}^{n-1} \frac{\delta u_{i,n_{l}-1}}{\delta \hat{\theta}_{i,l}} \dot{\hat{\theta}}_{i,l}.$$
(36)

From Eqs 35, 36, the derivative of $\dot{V}_{i,n}$ is computed as

$$\dot{V}_{i,n_{i}} \leq \dot{V}_{i,n_{i}-1} + e_{i,n_{i}} \left(g_{i,n_{i}} (x_{i},0) u_{i} - g_{i,n_{i}-1} e_{i,n_{i}-1} + \dot{F}_{i,n_{i}} (\Gamma i) - \frac{e_{i,n_{i}}}{2} \right) - \frac{\tilde{\theta}_{i,n_{i}}}{\gamma_{i,n_{i}}} \dot{\hat{\theta}}_{i,n_{i}},$$
(37)

where

$$\dot{F}_{i,n_{i}}(\Gamma_{i}) = \varphi_{i,n_{i}}^{T}(x_{i})\eta_{i,n_{i}} + D - \sum_{l=1}^{n-1} \frac{\partial u_{i,n_{i}-1}}{\partial x_{i,l}} \\
\times \left(g_{i,l}(x_{i})x_{i,l+1} + f_{i,l}(x_{i}) + \varphi_{i,l}^{T}\eta_{i,l}\right) \\
- \sum_{l=1}^{n-1} \sum_{j \in N_{j}} \frac{\partial u_{j,n_{i}-1}}{\partial x_{j,l}} \left(g_{j,l}(x_{j})x_{j,l+1} + f_{j,l}(x_{j}) + \varphi_{j,l}^{T}\eta_{j,l}\right) \\
+ \varphi_{j,l}^{T}\eta_{j,l}\right) - \frac{\partial u_{i,n_{i}-1}}{\partial y_{d}}\dot{y}_{d} - \sum_{l=1}^{n-1} \frac{\delta u_{i,n_{i}-1}}{\delta \hat{\theta}_{i,l}}\dot{\hat{\theta}}_{i,l} \\
+ \frac{e_{i,n_{i}}}{2} + g_{i,n_{i}-1}e_{i,n_{i}-1}.$$
(38)

Furthermore,

$$e_{i,n_{i}}\dot{F}_{i,n_{i}}\left(\Gamma_{i}\right) \leq \frac{\theta_{i,n_{i}}}{2c_{i,n_{i}}^{2}}e_{i,n_{i}}\bar{S}_{i,n_{i}}^{T}\left(\Gamma_{i}\right)\bar{S}_{i,n_{i}}\left(\Gamma_{i}\right) + \frac{c_{i,n_{i}}^{2}}{2} + \frac{e_{i,n_{i}}^{2}}{2} + \frac{\tau_{i,n_{i}}^{2}}{2}.$$
(39)

Hence, the virtual control signal is designed as

$$u_{i,n_{i}} = -a_{i,n_{i}}e_{i,n_{i}} - \frac{\theta_{i,n_{i}}}{2c^{2}_{i,n_{i}}}e_{i,n_{i}}\bar{S}_{i,n_{i}}^{T}(\Gamma_{i})\bar{S}_{i,n_{i}}(\Gamma_{i}), \qquad (40)$$

where a_{i,n_i} is the positive constant.

Substituting Eqs 34-40 into Eq. 35, we have

$$\dot{V}_{i,n_{i}} \leq -\sum_{l=1}^{n_{i}} a_{i,l} e_{i,l}^{2} - \sum_{l=1}^{n_{i}} \frac{\tilde{\theta}_{i,l}^{2}}{2\gamma_{i,l}} + \sum_{l=1}^{n_{i}} \mu_{i,l} + e_{i,n_{i}} (g_{i,n_{i}}(x_{i})u_{i} - u_{i,n_{i}}).$$
(41)

Furthermore, the actual ETC input strategy is as follows:

$$v_{i}(t) = \varsigma_{0}^{-1} \left(1 + \varrho_{0}\right) \left(u_{i,n_{i}} tanh\left(\frac{u_{i,n_{i}}e_{i,n_{i}}}{\varepsilon}\right) + \sigma_{1} tanh\left(\frac{\sigma_{1}e_{i,n_{i}}}{\varepsilon}\right) + e_{n} tanh\left(\frac{e_{n}e_{i,n_{i}}}{\varepsilon}\right)\right),$$

$$(42)$$

$$u_i(t) = v_i(t_k), \quad \forall t \in [t_k, t_{k+1}),$$
(43)

$$t_{k+1} = \inf\{t \in \mathbb{R}^+ \mid |e_c(t)| \ge \varrho_0 |u(t)| + \omega_0 e_n + \nu_0\}, \qquad (44)$$

where $e_c(t) = v(t) - u(t)$, $e_n = \left[1/\sum_{k=1}^n |e_k(t)| + \kappa_1\right]$ and $\kappa_1 > 0$, $\varsigma_0 > 0$, $0 < \varrho_0 < 1$, $\omega_0 > 0$, and $v_0 > 0$ are positive design parameters such that $\varsigma_0 \le f_a^*$, $\omega_0 f_b^* < 1 - \varrho_0$, $v_0 f_b^* < \sigma_1 (1 - \varrho_0)$.

According to Eq. 44, $v(t) - u(t) = \lambda_0(t) (\varrho_0 | u(t) | + \omega_0 e_n(t) + v_0)$ where $\lambda_0(t)$ is a continuous function and $\lambda_0(t_k) = 0, \lambda_0(t_{k+1}) = \pm 1, \lambda_1(t) = \pm \lambda_0(t), |\lambda_0(t)| \le 1, \text{ and } |\lambda_1(t)| \le 1, \forall t \in t_k, t_{k+1}.$ Since $a \in \mathbb{R}, \varepsilon > 0, -a \tanh(\frac{a}{\varepsilon}) \le 0, e_{i,n_i}v_i(t) \le 0, \text{ and } \frac{e_{i,n_i}v_i(t)}{1+\lambda_1(t)\varrho_0} \le \frac{e_{i,n_i}v_i(t)}{1+\varrho_0}.$

Then,

$$u_{i}(t) = \frac{v_{i}(t)}{1 + \lambda_{1}(t)\varrho_{0}} - \frac{\lambda_{0}(t)\varrho_{0}e_{0}}{1 + \lambda_{1}(t)\varrho_{0}} - \frac{\lambda_{0}(t)v_{0}}{1 + \lambda_{1}(t)\varrho_{0}}.$$
 (45)

Substituting Eqs 42-45 into Eq. 41, one obtains

$$\begin{split} \dot{V}_{i,n_{i}} &\leq -\sum_{l=1}^{n_{i}} a_{i,l} e_{i,l}^{2} - \sum_{l=1}^{n_{i}} \frac{\tilde{\theta}_{i,l}^{2}}{2\gamma_{i,l}} - |e_{n}| |e_{i,n_{l}}| - \sigma_{1} |e_{i,n_{l}}| \\ &+ \sum_{l=1}^{n_{i}} \mu_{i,l} + \frac{f_{b}^{*} \omega_{0} |e_{0}| |e_{i,n_{l}}|}{1 - \varrho_{0}} + \frac{f_{b}^{*} v_{0} |e_{i,n_{l}}|}{1 - \varrho_{0}} + 3\kappa\varepsilon \qquad (46) \\ &\leq -\sum_{l=1}^{n_{i}} a_{i,l} e_{i,l}^{2} - \sum_{l=1}^{n_{i}} \frac{\tilde{\theta}_{i,l}^{2}}{2\gamma_{i,l}} + \sum_{l=1}^{n_{i}} \mu_{i,l} + 3\kappa\varepsilon. \end{split}$$

Remark 1: The newly introduced decreasing function $e_n(t)$ gives a higher triggering threshold when the tracking error e_k , k = 1, 2, ..., n is very small. According to $|e_c(t)| \ge \varrho_0 |u(t)| + \omega_0 e_n + v_0$, choosing the fixed threshold v_0 and parameters appropriately, ω_0 and ϱ_0 can achieve the expected tracking performance.

4 Stability analysis

We are now prepared to state the main results of this research after the analysis mentioned previously.

Theorem 1: Consider the non-linear MASs (Eq. 1) satisfying Assumption 2. For bounded initial conditions, the virtual control signals (Eqs 20, 30, 40), adaptive laws (Eqs 22, 31), and the

tracking control protocol (Eq. 43) based on Assumptions 1–3 are obtained. The whole controller design process ensures that the signals of all closed-loop systems are bounded.

Proof: The derivative of V_{I,n_i} is rewritten as

$$\dot{V}_{i,n_i} \le -\bar{\omega}_i V_{i,n_i} + \beta_i, \tag{47}$$

where $\beta_i = \sum_{l=1}^{n_i} \mu_{i,l} + 3\kappa\varepsilon$, $\varpi_i = min\{2a_{i,l}, \gamma_{i,l}\}$. The total Lyapunov candidate function V is $V = \sum_{i=1}^{N} V_{i,n_i}$.

From Eq. 47, one obtains

$$\dot{V} \le -\,\mathfrak{O}V + \beta,\tag{48}$$

where $\varpi = \min{\{\varpi_i, i = 1, 2, ..., N\}}$ and $\beta = \sum_{i=1}^N \beta_i$. Furthermore, Eq. 48 satisfies

$$0 \le V(t) \le e^{\bar{\omega}t} V(0) + \frac{\beta}{\bar{\omega}} \left(1 - e^{\bar{\omega}t}\right).$$
(49)

From Eq. 49,

$$\|E_1\|^2 \le 2e^{-\tilde{\omega}t}V(0) + \frac{2\beta}{\tilde{\omega}}\left(1 - e^{-\tilde{\omega}t}\right).$$
(50)

Theoretically, the following inequality can be made to hold by choosing the design parameters $a_{i,k}$, c_i , γ_i correctly based on the definitions of ϖ and β

$$\frac{\beta}{\omega} \le \frac{\varsigma^2}{2} \left(\underline{\beta} \left(L + \tilde{B} \right) \right)^2, \tag{51}$$

where arbitrary $\varsigma > 0$.

Lemma 3 states that the result $\lim_{t\to\infty} ||Y-Y_c|| \leq \zeta$ may be obtained by selecting the proper parameters, which implies that the system output is guaranteed to converge to a tiny finite error.

We can find $t^* > 0$ such that the $\{t_{k+1} - t_k\} \ge t^*$, $\forall k \in z^+$. For $e_c(t) = v(t)-u(t), \forall t \in [t_k, t_{k+1})$, we get

$$\frac{\mathrm{d}}{\mathrm{d}t}|e_c| = \mathrm{sign}\left(e_c\right)\dot{e}_c \le \left|\nu\left(t\right)\right|. \tag{52}$$

We know that $\dot{v}(t)$ is continuously bounded. Consequently, there is a positive constant v such that $|\dot{v}(t)| \leq v$. Furthermore, $e_c(t_k) = 0$ and $\lim_{t \to t_k} e_c(t) = m$, the lower bound t^* that satisfies $t^* \geq \frac{m}{v}$ can be obtained. The issue with the Zeno behavior is, therefore, resolved.

5 Simulation study

In this section, we will verify the effectiveness of the designed control strategy through a numerical example. Consider the following second-order non-affine non-linear system. The system's communication structure is shown in Figure 1, where node 0 represents a virtual leader. It is obvious that only follower 1 is capable of receiving the leader's signal. The system model is given by the following formula:

$$\begin{aligned} \dot{x}_{i,1} &= g_{i,1}(x_i) x_{i,2} + f_{i,1}(x_i) + \varphi_{i,1}^T(x_i) \eta_{i,1} \\ \dot{x}_{i,2} &= f_{i,2}(\Delta_i, u_i) + \varphi_{i,2}^T(x_i) \eta_{i,2} + d_i(t) \\ y_i &= x_{i,1}. \end{aligned}$$
(53)

The aforementioned non-linear functions are $g_{11} = 1 + \sin(x_{11} + 1)$, $f_{11} = 0.6 \cos(x_{11})$, $g_{12} = 1 + \sin(x_{12})$, $C_1 = 2 \cos(x_{12})$, $d_1 = 0.1 \sin(x_{12})$, $g_{21} = \sin(x_{21} + 1)$, $f_{21} = 0.8 \cos(x_{21})$, $g_{22} = 1 - \sin(x_{22})$, $C_2 = 0.8 \cos(x_{22})$, $d_2 = 0.1e^{-\frac{x_{22}}{2}}$, $g_{31} = \sin(x_{31} + 1)$, $f_{31} = 0.8 \cos(x_{31} + 1)$, $f_{31} = 0.8 \cos(x_{32} + 1)$, $f_{31} = 0.8 \cos(x_{31} + 1)$, f_{31}

















0.5 $\cos(x_{31})$, $g_{32} = 1 + \sin(x_{32} + 1)$, $C_3 = 0.5 \cos(x_{32})$, $d_3 = 0.3 \sin(x_{32})$, and $\varphi_{i,j} = e^{-x_{i,j}}$, i = 1, 2, 3, j = 1, 2, 3. The signal of the leader is chosen as $y_d = 0.6 \sin(t) + 1$. We choose the initial values $(0) = [1,1]^T$, $x_2(0) = [1,1]^T$, $x_3(0) = [1,1]^T$, $x_4(0) = [0,0]^T$, $x_5(0) = [0,0]^T$, and $x_6(0) = [1,1]^T$. The design parameters are selected as $a_{11} = 1$, $a_{12} = 10$, $a_{21} = 1$, $a_{22} = 1$, $a_{31} = 1$, $a_{31} = 1$, $\gamma_{i,j} = 1$, $i = 1, 2, 3, j = 1, 2, 3, k_{11} = 30$, $k_{12} = 40$, $k_{21} = 28$, $k_{22} = 35$, $k_{31} = 40$, and $k_{32} = 30$.

Concomitantly, we get the simulation results in Figures 2–7. Figure 2 shows that the actual output of the studies' systems can track well with the expected trajectory y_r . Figure 3 shows the error between the output signals and the expected signal. Figure 4 shows the adaptive parameter curves of each follower. The curves of the controller are shown in Figures 5–7. Figure 8 shows

the event-triggered times and the threshold value comparisons of the two methods.

6 Conclusion

This research investigates the consensus tracking control problem for a class of non-affine non-linear MAS and proposes a design scheme for control input separation. The Taylor decoupling technology is used to successfully decouple the control inputs with the non-affine non-linear terms. Then, the unknown non-linear functions that exist in the non-affine non-linear MASs are approximated using RBFNNs. Moreover, an improved ETC strategy is proposed, which introduces a decreasing function to improve the performance of the ETC strategy. This ETC strategy significantly reduces the computational burden of the communication process and achieves better control objectives. The designed control strategy ensures the boundedness of all signals and achieves good consensus tracking performance. In the future, we will focus on extending the proposed method to MASs with more general structures and malicious attacks.

Data availability statement

The original contributions presented in the study are included in the article/Supplementary material. Further inquiries can be directed to the corresponding author.

Author contributions

LW, ZS, and CL contributed to the study idea and design. LW wrote the first draft of the manuscript. LW organized the literature. ZS designed the figures. LC verified the experimental design. All authors contributed to the article and approved the submitted version.

Conflict of interest

The authors declare that the research was conducted in the absence of any commercial or financial relationships that could be construed as a potential conflict of interest.

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