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# Graphene based microstrip patch antenna on photonic crystal substrate for 5G application

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One of the strongest and thinnest substances known to exist is graphene. It can be efficiently used for antenna fabrication and significant benefits can be analyzed such as size miniaturization, gain and return loss enhancement, flexibility and increase in bandwidth, etc. In this paper, a graphene patch antenna has been designed and fabricated for 5G applications at 28.3 GHz. Different 5G frequency allocation throughout the world is shown in tabular form. Photonic crystal substrate is used to reduce surface wave loss which in turn, increases the radiation efficiency of the antenna. The ground plane is made up of copper, treated as a Perfect Electric Conductor Several unique properties of graphene are analyzed and shown. Then, a thin graphene sheet is used as a patch and an antenna is designed and fabricated. The patch has dimensions of 3.4 mm by 2.38 mm. The simulated return loss is maximum at -33.69 dB at 28.3 GHz, whereas for the measured design it is -26.31 dB at 28.04 GHz. It is observed that graphene can be a good alternative to existing conductors, especially at higher frequencies. The proposed antenna is designed and simulated using HFSS software.

#### KEYWORDS

microstrip antenna, 5G, graphene, photonic crystal substrate, gain, return loss

### Introduction

Massive advancement in the field of communication leads towards the fifthgeneration technology commonly known as 5G. It is the wireless communication standard after the fourth generation (4G/IMT). The standard document of 5G had been published by 3GPP and officially approved in the Plenary Meeting in June 2018. It has been developed to provide next-generation user experiences, enable new deployment patterns, and deliver new services. 5G will push the mobile environment into new worlds with its fast speeds, greater reliability, and low latency.

Various countries have different frequency spectrum allocations for mobile communication. Millimeter wave bands (26, 28, 38, and 60 GHz) will be used for 5G communication in general (IMT-2020, 2020; Shimodaira et al., 2015; Marcus, 2016) as shown in Table 1. It is 24.5–29.5 GHz for India, while the FCC has allotted an extraordinary



7 GHz of un-channelized spectrum for license-free operation between 57 and 64 GHz (Kao et al., 2013; Wang et al., 2016). This makes it possible for anyone to create apps that operate in this frequency range in order to fully utilize this spectrum for their needs.

Designing an antenna at such a high frequency band is a real challenge due to its compact size and wide band requirement to incorporate the entire accessible band. The microstrip patch antenna is one of the most useful antennae due to its several advantages such as low weight, compact size, conformal to planer and non-planer surfaces and ease of fabrication and lower fabrication cost *etc.* It has various shapes e.g., rectangular, square, circle and can be easily etched on any PCB (Balanis, 2005).

The low efficiency, narrow bandwidth, and lower gain of microstrip antennas is a major operational drawback. To boost the bandwidth and gain of the microstrip patch antenna, some strategies such as raising the substrate thickness or patch size were used. However, a thicker substrate reduces radiation efficiency due to greater surface wave losses. Because it can limit surface wave propagation and improve the electrical performance of the microstrip patch antenna, photonic crystal substrate is a superior answer to this problem. The photonic crystal substrate is a superior answer to this problem. The photonic crystal is a non-natural material made of periodic implant structure within the surrounding medium, which might be dielectric or conductor. Scattering and diffraction properties of periodic structures greatly affect the electromagnetic wave propagation through them. So, antennas with photonic crystals have many unique characteristics (Ozbay et al., 2003; Gonzalo et al., 1999; Meade et al., 1993; Gonzalo et al., 21312138).

Graphene has recently attracted tremendous interest in various research areas due to its exceptional electrical and mechanical properties (Geim and Novoselov, 2007). Graphene is a hexagonally organized honeycomb lattice material with great mobility (8000 cm<sup>2</sup>/V), 5000 W/m/K thermal conductivity, and

observable flexibility (Novoselov et al., 2004; Geim and Novoselov, 2007; Song et al., 2018). In general, the conductivity of graphene is very frequency-dependent and can have completed different behaviour e.g., at microwave and THz. Recently, graphene has gained a lot of attention as a material and has various uses in the microwave to the mm-wave range (Warner et al., 2013; Bozzi et al., 2015). But, the application of graphene in passive guided devices and antennas from microwaves to THz has been by far less exploited.

The use of graphene material as conducting substance for the patch is speculated to enhance the bandwidth and radiation efficiency of the antenna due to its stand-out functionalities and electromagnetic properties. So, several works have been published using graphene as antennas, couplers, filters, amplifiers and other telecommunication devices.

Wang et al. (2015) presented the concept of the graphenebased two-dimensional leaky-wave antenna (LWA), allowing both frequency tuning and beam steering in the terahertz (THz) band. Goyal and Vishwakarma (Goyal and Kumar Vishwakarma, 2018) investigated the effect of graphene's chemical potential on return loss and resonant frequency of the antenna; Mojtaba and Carey (Dashti and David Carey, 2018) show a graphene antenna with dynamic conductivity to produce a surface impedance microstrip antenna with a bandwidth of hundreds of GHz. In Qu et al. (2018) proposed a frequencyselective surface (FSS) with tunable band stop performance.

The choice of substrate is also very important for obtaining efficient antennas. Photonic band gaps (PBGs) have been increasingly used in order to improve the performance and characteristics of antenna designs (Choudhury et al., 2016).

Conventional metallic patch antennas suffer from high propagation and scattering losses in the environment. So, graphene-based antennas can be used as an alternative. Azizi et al. (2017) compared the different characteristics of a graphene-patch antenna with the conventional copper-patch antenna in the THz frequency band. It shows the return loss increases from -15 to -29 dB and gain improves to 7.16 dB when graphene is used as patch material while it was 5.63 dB when the conventional copper patch was used.

This paper presents a microstrip antenna on a photonic crystal substrate in which graphene may be used as patch material. But due to fabrication limitations, copper is used instead of graphene. The rest of this paper is organized as follows: Section 2 describes graphene and its properties. Antenna design on photonic crystal substrate is discussed in Section 3. Simulated and measured results are discussed in Section 4. Section 5 concludes the paper.

### Graphene and its properties

Graphene is used as conducting patch material which is the two-dimensional version of graphite. In graphene, the



energy-momentum relationship for electrons is linear over a wide range of energies, rather than quadratic, so electrons in graphene behave as massless relativistic particles "Dirac fermions" with an energy-independent velocity (Wallace, 1947; Geim and Novoselov, 2007; Charlier et al., 2008). It is a zero-overlap semimetal (with both holes and electrons as charge carriers) with very high electrical conductivity as shown in Figure 2. High charge carrier mobility, for which values of 10,000 cm<sup>2</sup>/Vs., in some cases even 200,000 cm<sup>2</sup>/Vs were reported. Its surface conductivity is frequency dependent and can be described using Kubo's formula approximation (Gusynin et al., 2007) in its most simplified form given by

$$\sigma(i\omega,\mu_{c},\Gamma,\Gamma) = -j\frac{q_{c}^{2}(\omega+2j\Gamma)}{\Pi\hbar^{2}} \left[ \frac{q_{c}^{2}}{(\omega+2j\Gamma)^{2}} \int_{0}^{\infty} \varepsilon \left( \frac{\partial f_{d}(\varepsilon)}{\partial \varepsilon} - \frac{\partial f_{d}(-\varepsilon)}{\partial \varepsilon} \right) d\varepsilon - \int_{0}^{\infty} \frac{f_{d}(-\varepsilon) - f_{d}(\varepsilon)}{(\omega+2j\Gamma)^{2} - \left(\frac{4E}{\hbar}\right)^{2}} d\varepsilon \right]$$
(1)

where -e is the electronic charge,  $\hbar = h/2\Pi$  is the reduced Planck's constant.

 $f_d(\varepsilon) = [e \text{ kBT}+1]^{-1}$  is the Fermi-Dirac distribution and  $k_B$  is the Boltzmann's constant

ω is radian frequency,  $μ_c$  is chemical potential, Γ is a phenomenological scattering rate, that is, assumed to be independent of energy ε and T is temperature. The first term in Eq. 1 is due to intraband contributions, and the second term to interband contributions. Surface conductivity dependency on frequency is shown in graphical form, which is achieved using MATLAB simulation. Here,  $μ_c$  is taken as 0ev which represents the worst case (Hanson, 2008).

Figure 1 shows conductivity plot with frequency, keeping the chemical potential as zero for a graphene flake within the GHz-THz band. Because of its two-dimensional crystalline structure and exceedingly thin layer height, graphene is thought of as having a lattice structure. The minimal thickness is considered from the perspective of







electromagnetic simulation so that the EM field distribution may observe the thickness of the graphene and wave propagation can happen.

From the figure, it can be seen that the conductivity of graphene shows inductive nature and it allows infinite graphene sheet to support Transverse Magnetic surface waves, which is referred to as Surface Plasmon Polariton (SPP), since the conductivity results from the plasma-like behaviour of electrons [25].

An electron attached to an atom in an insulator or semiconductor can only break free if it receives enough energy from heat or a passing photon to cross the "band gap". However, with graphene, the gap is infinitesimally FIGURE 5 Size approximation of fabricated Patch Antenna and compare with rupee coin.

small, as illustrated in Figure 2A. This is the primary reason why graphene electrons flow so readily and quickly. Figure 2B depicts a freestanding graphene with the valence and conduction energy bands, denoted by  $\Pi$ and  $\Pi$ \* bands, exhibiting a linear (conical) relationship between electron energy and momentum and intersecting at particular momentum positions K and K' (left). Conrad and colleagues (Geim and Novoselov, 2007; Nevius et al., 2015) demonstrated that, while the first carbon layer of samples grown epitaxially on a silicon carbide substrate at around 1340°C is electronically inert and thus lacks a band structure (middle), at growth temperatures 20°C higher, the layer becomes active and demonstrates and  $\Pi$  and  $\Pi *$ bands separated by a gap of 0.5 electron volts around the K and K'. Because of this peculiar dispersion relationship, graphene is a zero-gap substance that conducts electrons like a metal.

Geographical location	5G frequency spectrum
Europe	For commercial installations beginning in 2020, the frequency range will be 24.25-27.5 GHz
China	Studies at 24.25-27.5 and 37-43.5 GHz
Japan	Trials between 27.5-28.28 GHz are anticipated to begin in 2017, with commercial installations beginning in 2020
Korea	Trials at 26.5-29.5 GHz in 2018 and industrial deployments in 2019
United States	Pre-commercial deployments at 27.5-28.35 and 37-40 GHz in 2018
INDIA	24.5-29.5 GHz

TABLE 1 5G frequency band spectrum of different geographical areas.

TABLE 2 Optimized dimensions of Patch antenna.

Design parameter	Value	Design parameter	Value
Resonant Frequency	28.3 GHz	Patch Length	5 mm
Relative Permittivity	4.4	Patch width	5 mm
FR4 Substrate thickness	0.254 mm	Feedline Length	4.71 mm
Substrate width	10 mm	Feedline width	0.32 mm
Substrate length	10 mm	_	_

The particles in each valley can formally be described by Dirac-like Hamiltonian (Geim and Novoselov, 2007)

$$H = \hbar \nu_f \begin{pmatrix} 0 & k_x - ik_y \\ k_x + ik_y & 0 \end{pmatrix} = \hbar \nu_f \sigma. K$$
(2)

where K is the quasiparticle momentum,  $\sigma$  is the 2D Pauli matrix and k-independent Fermi velocity  $v_F$  plays the role of the speed of light.

### Graphene fabrication

A unique process for fabricating patch antenna with graphene is presented by (Sajal and Braaten, 2015) at the resonant frequency of 2.6 GHz. In this paper, a thin graphene sheet is pated on 100  $\mu$ m thick paper and cut in the shape of a patch using g commercially available micro-cutter. Another fabrication approach is given by Zhou et al. (2020) in which a conductive ink was made by mixing graphene flakes with N-Methyl-pyrrolidone (NMP) then, the desired pattern was printed on a paper substrate using screen-printing technology. The operating frequency is 5.8 GHz for wireless communication applications. A graphene-based bow-tie antenna is designed by (Avsar Aydin, 2021) for biomedical applications using 3-D printing technology. Conductive Graphene PLA filament for the core of the antenna and standard PLA for the dielectric substrate parts is used in this design. It provides an ultra-wide

bandwidth of 18 GHz in the frequency range of 2–20 GHz for ultra-wideband applications.

### 4 Antenna design and fabrication

A microstrip antenna is made up of a metallic patch on top of a dielectric substrate with a metallic conductor beneath it to serve as a ground plane. Different characteristics of antenna such as resonant frequency, bandwidth, gain, return loss, radiation pattern *etc.* depends on the substrate, patch dimension and feed type. This proposed antenna is designed on a photonic crystal using FR4 material ( $\varepsilon r = 4.4$  and loss tangent tan  $\delta =$ 0.002) which acts as a substrate as shown in Figure 3.

Figure 3 shows the proposed photonic crystal with an air gap in the dielectric slab. FR4 (relative permittivity of 4.4 and dielectric loss tangent of 0.002) is used as a substrate. An air gap hole of radius 0.3 mm with a period of 1 mm is taken and embedded in the pattern of six rows along length and six columns along the width, i.e., a total number of 36 cylindrical air holes are created as shown in figure. A much closer view of the photonic band gap crystal is shown in Figure 3B where cylindrical radius and periods are mentioned. The proposed substrate has length, width, and height of 10 mm, 10 mm and 0.254 mm respectively. As shown in Figure 4 & Figure 5.

The performance of the microstrip antenna is dependent on its size, operating frequency, radiation efficiency,





FIGURE 7

(A) FR4 substrate with a copper coating (B) etched FR4 (C) cylindrical holes of 0.3 mm is drilled in the substrate (D) Copper patch is pasted on Substrate.



directivity, and other criteria (Balanis, 2005). For effective radiation, the following equations are utilized in calculations:

The practical width of the patch is calculated using equation (Gonzalo et al., 21312138)

$$W = \frac{1}{2f_r \sqrt{\mu_0 \varepsilon_0}} \sqrt{\frac{2}{\varepsilon_r + 1}} = \frac{\nu_0}{2f_r} \sqrt{\frac{2}{\varepsilon_r + 1}}$$
(3)

Where  $v_0$  is the free space velocity of light.

The effective dielectric constant,  $\varepsilon_{eff}$  is given (Balanis, 2005):

#### TABLE 3 Summarizes the effect of change in radius of cylindrical holes.

Structure having photonic substrate with air gap radius	Resonant frequency (GHz)	Return loss (S11 in dB)	Bandwidth (GHz)	Gain(dB)
R = 0.1 mm	25.94	-22.99	0.78	2.14
R = 0.20 mm	26.18	-24.8	1.1	2.73
R = 0.30 mm	28.3	-33.69	1.30	3.13

TABLE 4 Cylindrical air gap volume Vs. FR4 Volume correlation of proposed antenna.

Air gap radius (mm)	Air cylindrical volume	FR4 cuboid available volume	Ratio (Air/FR4)
0.10	$\pi r^2 h = 3.14^* (0.1 \text{ mm})^{2*} 0.254 \text{ mm} = 0.008 \text{ mm}^3$	L*W*h -36* $\pi r^2$ h = 10 mm*10 mm*0.254 mm -36 * [3.14*(0.1 mm) <sup>2</sup> * 0.254 mm] = mm <sup>3</sup>	1/2896
0.1	0.008 mm <sup>3</sup>	25.112 mm <sup>3</sup>	1/3139
0.15	0.018 mm <sup>3</sup>	24.752 mm <sup>3</sup>	1/1375
0.2	0.032 mm <sup>3</sup>	24.248 mm <sup>3</sup>	1/755
0.25	0.049 mm <sup>3</sup>	23.636 mm <sup>3</sup>	1/482
0.3	0.072 mm <sup>3</sup>	22.808 mm <sup>3</sup>	1/316

$$\varepsilon_{eff} = \frac{\varepsilon_r + 1}{2} + \frac{\varepsilon_r - 1}{2} \left[ 1 + \frac{12h}{w} \right]^{-\frac{1}{2}}$$
(4)

Where  $\varepsilon_r$  is the dielectric constant of the material, *w* is the width of the patch, and *h* is the height of the substrate.

Prolonged electrical length,  $\Delta L$  of the patch due to the fringing effect can be calculated using (Balanis, 2005)

$$\Delta L = 0.412h \frac{\left(\varepsilon_{eff} + 0.3\right)\left(\frac{w}{h} + 0.264\right)}{\left(\varepsilon_{eff} - 0.258\right)\left(\frac{w}{h} + 0.8\right)}$$
(5)

The effective length,  $L_{eff}$  is given (Balanis, 2005)

$$L_{eff} = \frac{\nu_0}{2f_r \sqrt{\varepsilon_{eff}}} - 2\Delta L \tag{6}$$

The proposed antenna has a rectangular radiating patch excited with strip line feeding for 50  $\Omega$  characteristic impedance. Conducting patch has a dimension of 3.4 mm × 2.38 mm and a thin copper sheet of thickness 0.01 mm is used for it. FR4 is used as the substrate (dielectric constant  $\varepsilon_r = 4.4$  and loss tangent tan  $\delta = 0.002$ ) having the dimension of 10 × 10 × 0.254 mm. The patch is excited through a strip line having dimensions of 0.32 mm × 4.71 mm × 0.01 mm.

### Fabrication steps

Due to fabrication limitations, FR4 with copper coating is used instead of graphene. This antenna was fabricated using

photolithographic process. Table 2 shows the dimensions of the proposed antenna. This process constitutes an Ultraviolet light (UV ray) of the corresponding wavelength and desired photo-resist. Negative and positive photo-resist materials are there. The exposed portion of negative photo-resist hardens when the photo-resist developer is applied to it. Whereas in the case of positive photo-resist the exposure to the UV light changes the chemical structure of the resistor so that it becomes more soluble in the developer.

Flow chart of the fabrication steps is shown in Figure 6. In this fabrication, we use positive photo-resist (DNQ-Novolac photo-resist) for fabrication. Firstly, a double-sided copper coated FR4 substrate is taken. DNQ thin film is applied as lamination on overall substrate and copper layer on both side is completely etched out using FeCl<sub>3</sub> solution. This substrate in properly rinsed in clean running water to remove any impurities and then dried. In the next step, cylindrical holes of radius 0.3 mm are drilled in a periodic manner with separation of 1 mm. Substrate is properly cleaned again to eliminate any particle on it so that resonant frequency of antenna remains unaltered. With the help of non-dielectric adhesive solution, a very thin copper sheet is pasted on lower side of the substrate which acts as ground plane and patch of calculated size is pasted over it. Temperature is maintained at 200°C for proper pasting of sheets. Since non-dielectric adhesive solution is used, relative permittivity remains unaltered. Step by step fabrication process is illustrated in Figure 7.





# **Results and discussion**

First of all, the proposed antenna is designed on a normal FR4 substrate and return loss is analyzed. Then, photonic bandgap substrate is formed by implanting the cylindrical holes within the substrate periodically. The hole radius varies from 0.1 mm to 0.3 mm with a step size of 0.1 mm and its effect is shown in Figure 8. It can be seen that results are improved in terms of return loss, gain and bandwidth. A tabular comparison is shown in Table 3. As the hole radius changes, different characteristics of the antenna also change which is analyzed

here and the 0.30 mm radius has the lowest return loss -33.69 dB. This lowest return loss occurs at 28.3 GHz having a bandwidth 1.30 GHz and gain of 3.13 dB, also these two parameters are higher in comparison to other parametric optimization results.

As the radius of the cylindrical air gap changes, air FR4 substrate volume also changes. With the increase in the air gap radius, the total available FR4 substrate volume decreases which, in terms, affects the relative permittivity of the substrate and this can be determined by observing Table 4. As mentioned air gap radius of 0.3 mm has a total available FR4 substrate volume of 22.808 mm<sup>3</sup>.





Figure 9 shows simulated *versus* measured return loss of proposed antenna. Solid lines show the simulated result whereas the dashed line represents the measured one. Simulated return loss is found to be -33.69 dB at 28.3 GHz whereas measured in S11 is -26.31 dB at 28.04 GHz as shown in Figure 9. The simulated bandwidth is found as 1.3 GHz whereas the measured bandwidth is 0.89 GHz which is in close proximity to the simulated one.

The far field radiation patterns (both E-plane and H-plane) of the fabricated antenna were measured in a fully equipped anechoic chamber as shown in Figure 10. Antenna Under Test is placed on an ETS Lindgren 2090 positioner and aligned to the reference horn antenna with adjustable polarization.

Surface current density distribution is shown in Figure 11. The current distribution is zero at the ends of the patch because the current can't flow "off" the patch. The voltage is out of phase with the current. The current density at the resonant frequency shows a greater amount of accumulated current on the radiating patch. Simulated VSWR is shown in Figure 12A and is 1.047 at the resonant frequency of 28.29 GHz which is a well-accepted value. Also, Figure 12B, shows the Gain result where the maximum Gain can be seen as 3.6 dB at zero degrees theta. By varying the photonic crystal hole radius on the substrate, the lowest return loss -33.69 dB has been obtained among other optimization. The resonant frequency for the lowest return loss value is 28.3 GHz having a bandwidth 1.30 GHz and gain of 3.13 dB and is higher among other results. The result shows the radius of holes 0.03 mm is useful in designing. Further, this fabricated antenna can be used for recent 5G applications since resonant frequency lies within India's 5G frequency spectrum. In the future, such photonic crystals required to test and analyze in different natural environments and temporal conditions.

# Conclusion

This research article shows the design and development of a rectangular microstrip patch antenna on the photonic crystal designed and simulated using HFSS with a resonant frequency of 28.3 GHz. Then, this antenna is fabricated using a photolithography process which is explained in brief. Electrical characterization of the antenna is done in the semianechoic chamber and measured results are in close proximity to the simulated one. Parametric optimization of the radius of cylindrical holes summarizes as the hole radius changes, different characteristics of the antenna also change which is analyzed here. The lowest return loss -33.69 dB occurs at 28.3 GHz having a bandwidth 1.30 GHz and gain of 3.13dB, and is higher in comparison to other parametric optimization results. With the increase in the air gap radius, the total available FR4 substrate volume decreases which, in terms, affects the relative permittivity of the substrate. Moreover, the surface current density has been observed and the current distribution is zero at the ends of the patch because the current can't flow "off" the patch. Further, this fabricated antenna can be used for recent 5G applications since resonant frequency lies within India's 5G frequency spectrum.

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### Data availability statement

The original contributions presented in the study are included in the article/Supplementary Material, further inquiries can be directed to the corresponding author.

### Author contributions

All authors listed have made a substantial, direct, and intellectual contribution to the work and approved it for publication.

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## Conflict of interest

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