



# Design and Hardware Experiment of Concentrating Magnetic Ring of Current Sensors Used in Open-Loop Fluxgate

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In order to improve the ability of fluxgate current sensor to detect weak residual current, starting from the magnetic ring, which is the key magnetic concentrating element, and adopting the method of Comsol simulation analysis, the open-loop fluxgate is introduced from the aspects of magnetic ring material selection and structural design, etc. The key technology of the design of the magnetic concentrating ring of the current sensor is verified by experiments that the magnetic ring can increase the sensitivity of the current measurement by 15.4 times, and the frequency can reach 6.7 kHz, which can meet the AC and DC complex leakage current detection. Provide technical reference for designers engaged in this type of sensor.

**Keywords:** fluxgate magnetic field sensor, open loop, magnetic concentrator, magnetic ring, residual current

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## 1 INTRODUCTION

Due to the continuous increase of DC equipment such as photovoltaic power generation, new energy vehicles, and DC charging piles, the residual current generated contains various DC components and is becoming more and more complex. In order to solve the problem of residual current detection containing DC signal, it is more important to realize the effective detection and leakage protection of AC and DC residual current (Young et al., 2007). Literature (Takahiro and Dev, 2011) proposed a simplified magnetic modulation AC and DC residual current detection method. Literature (Ponjavic and Duric, 2007) proposed to realize the detection of AC and DC residual current through the excitation current. Literature (Xu and Min, 2009) introduced a detection and protection of pulsating DC residual current signal through hardware design. Reference (Min, 2011) combined the “State Grid Core” AC-DC hybrid leakage monitoring technology with monitoring and protection modules in the photovoltaic power generation system. Reference (Xu et al., 2015) applied full current-sensitive RCD in the charging device of electric vehicle to detect AC and DC residual current to protect equipment and personal safety. In recent years, with the in-depth research of magnetic modulation residual current detection technology, A-type residual current protector can detect residual current in the form of DC and pulsating DC. In addition, many scholars at home and abroad have conducted in-depth research on the detection method of residual current and the identification method of current waveform, which provides theoretical support for the design of residual current detection system with complex waveform (O’Shaughnessy et al., 2021).

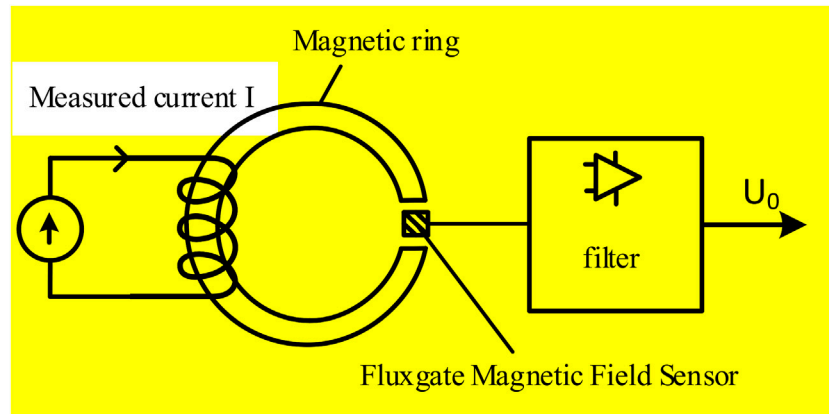


FIGURE 1 | Schematic diagram of fluxgate current sensor measurement.

TABLE 1 | Performance index of common magnetic materials.

| Performance          | Iron-based Amorphous Alloy | Silicon Steel | Iron-based nanoalloys | Ferrite   |
|----------------------|----------------------------|---------------|-----------------------|-----------|
| Bs(T)                | 1.56                       | 2.03          | 1.25                  | 0.5       |
| Maximum permeability | 45 × 104                   | 4 × 104       | 25 × 104              | 0.2 × 104 |
| Hc(A/m)              | <3                         | <30           | <1                    | 6         |
| Frequency(Hz)        | 15 K                       | 100           | 50 K                  | 20 K      |

## 2 THE STRUCTURE AND PRINCIPLE OF FLUXGATE CURRENT SENSOR

As a weak magnetic field measurement device with good comprehensive performance (Telukunta et al., 2017), the fluxgate sensor can measure the magnetic field generated by DC and AC, and also has the advantages of good zero-point stability, small temperature drift and high resolution. It has great advantages in measuring magnetic fields.

This paper proposes an AC and DC leakage current detection method, the principle of which is shown in Figure 1.

The fluxgate current sensor is mainly composed of a magnetic ring, an energized coil, a fluxgate magnetic field sensor and an amplifying filter circuit. The fluxgate magnetic field sensor can output a voltage signal proportional to the magnetic field, and indirectly measure the magnitude of the current by measuring the magnetic induction intensity generated by the residual current.

The magnitude of the magnetic induction is related to the total current passing through the loop and the magnetic permeability in vacuum. For the residual current, the safe current through the human body is not allowed to exceed 30 mA, and the magnetic permeability of the air is very small, so the magnetic induction intensity generated by the DC residual current in the air is very weak, and it is difficult for the fluxgate sensor to measure the residual current. The value of the current, which requires the sensor to have a high sensitivity (Grim et al., 2020).

It can be seen from the above analysis that adding a magnetic core with higher magnetic permeability can improve the sensitivity of the sensor, or winding more coils on the magnetic ring, so that the weak

DC residual current in the ring magnetic circuit can generate a large magnetic induction intensity. The method studied in this paper is to use a magnetic ring with high magnetic permeability to increase the ability of the magnetic focusing ring, which can more easily measure the magnetic field generated by the weak DC residual current. Ampere’s loop theorem in the presence of a magnetic medium:

$$\oint_L B \cdot dl = \mu_0 \sum I \tag{1}$$

where  $\mu = \mu_r \mu_0$ ,  $\mu$  is the magnetic permeability of the magnetic ring.

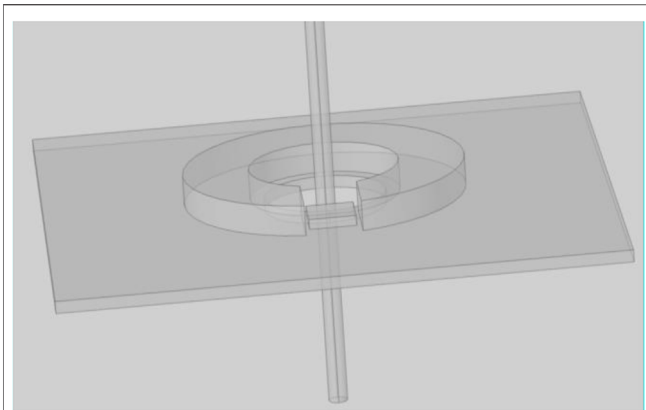
## 3 MAGNETIC RING DESIGN OF OPEN-LOOP FLUXGATE CURRENT SENSOR

$$H = B/\mu \tag{2}$$

Substitute Eq. 2 into Eq. 1 to get:

$$\oint H \cdot dl = \sum I \tag{3}$$

The above formula is to analyze the current in a wire. If the wire under test is evenly wound on the magnetic ring, then the Ampere loop theorem is also satisfied when there is a magnetic



**FIGURE 2** | Simulation model built.

medium (Velasco-Quesada et al., 2016). At this time, the formula can be transformed into:

$$Hl = NI \tag{4}$$

Where  $N$  is the number of turns of the coil and  $l$  is the average length of the magnetic circuit. At this time, the output of the sensor can be expressed as:

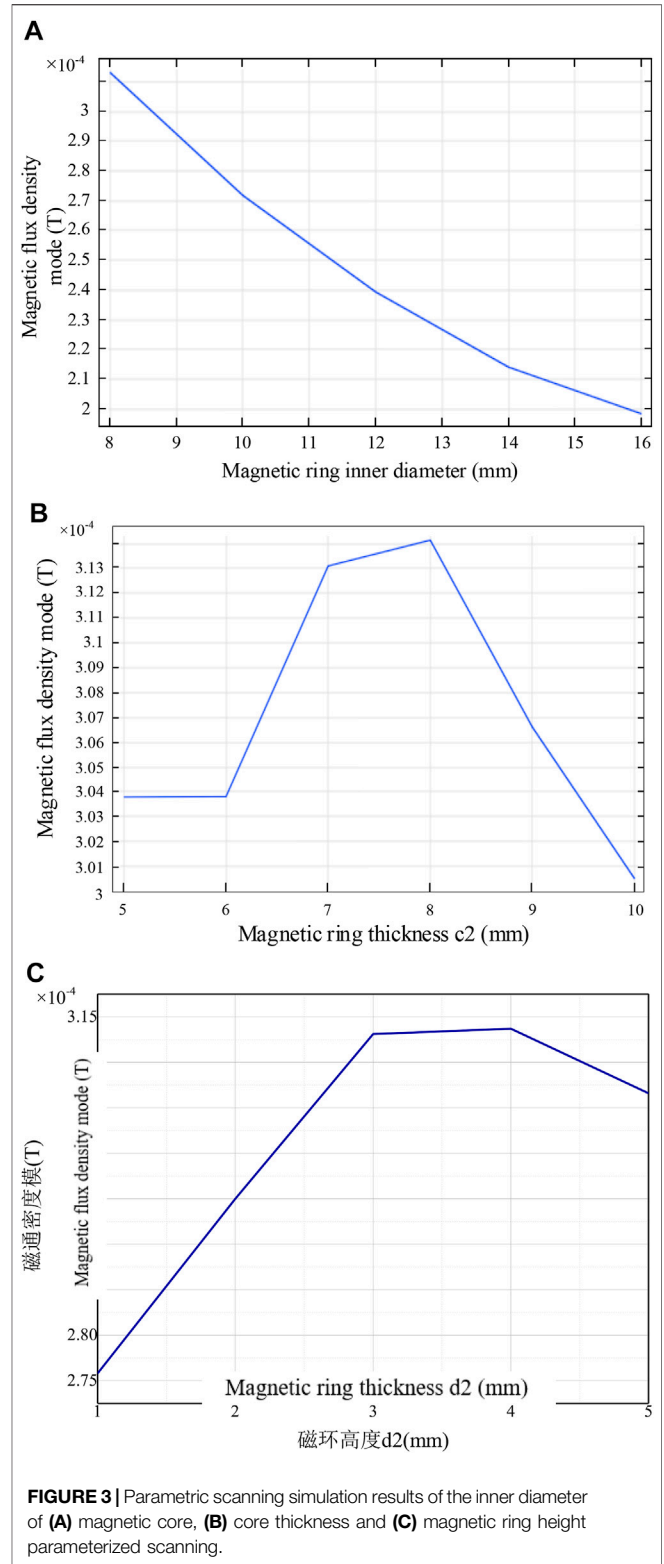
$$V_{VOUT} = B \times k = \frac{\mu NI}{l} \times k \tag{5}$$

Among them,  $k$  is the proportional coefficient, which is related to the characteristics of the sensor. According to the above formula, when the magnetic permeability of the magnetic ring is constant, the output of the sensor has a linear relationship with the current to be measured (Cao et al., 2019). Due to the hysteresis characteristics of the magnetic ring, the magnetic permeability of the magnetic ring is a constant value only when it works in the linear range, so a magnetic ring with low coercivity and high saturation strength must be selected (Velasco-Quesada, 2010). In this paper, the parameters of the magnetic core are designed by means of simulation, and the maximum value of the magnetic induction intensity at the opening of the magnetic core is obtained.

### 3.1 Selection of Magnetic Ring Materials

It can be seen from the above conclusions that the magnetic ring with low coercivity, high magnetic saturation strength and high magnetic permeability should be selected (Liang et al., 2019). Firstly, the material of the magnetic ring is selected. Common high permeability materials are: iron-based amorphous alloy, silicon steel, ferrite, iron-based nanoalloys etc., The performance indicators are shown in **Table 1**.

Among the above four materials, ferrite is the most widely used magnetic core material in switching power supplies, but compared with other materials, the magnetic permeability and magnetic saturation induction intensity are the lowest, so it is not selected; although iron-based nanoalloys have high magnetic permeability, they are not suitable for line-segment sensors; silicon steel sheet has high magnetic permeability and



**FIGURE 3** | Parametric scanning simulation results of the inner diameter of (A) magnetic core, (B) core thickness and (C) magnetic ring height parameterized scanning.

magnetic saturation strength, but its resistivity is too small, and the eddy current loss is large at high frequencies, so it is widely used in low frequency occasions (Qi et al., 2019). Therefore, when selecting the magnetic core material,

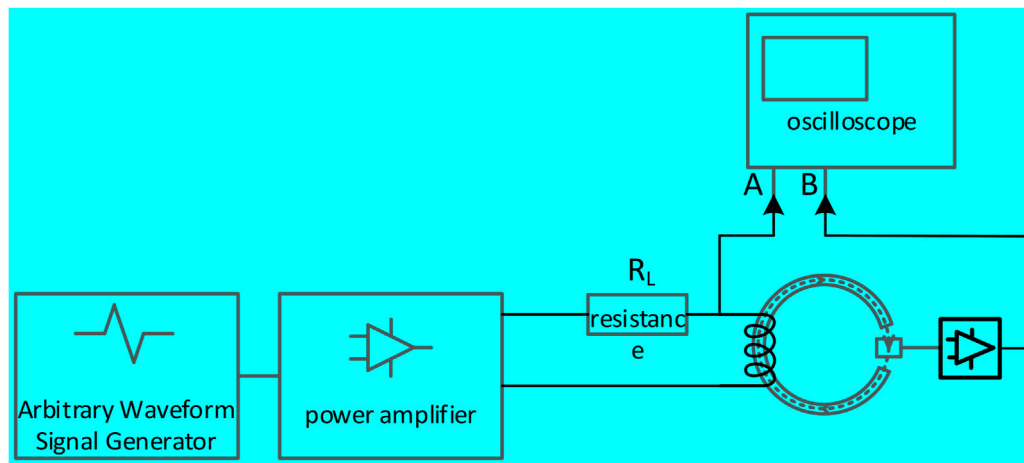


FIGURE 4 | Test principal diagram.

considering the processing difficulty, price, performance, etc., the magnetic material is finally selected as an iron-based amorphous alloy.

### 3.2 Design of Size Parameters of Magnetic Ring

After the magnetic ring material is determined, the size of the magnetic ring needs to be further determined. The inner diameter of the magnetic ring, the thickness of the magnetic ring, the height of the magnetic ring and the opening size of the magnetic ring will not only affect the size of the induced magnetic field, but also affect the measurement of the leakage current. Therefore, it is necessary to obtain the optimal parameters by controlling the variables.

#### 3.2.1 Magnetic Ring Opening Size Design

The magnetic concentrating ring is open-loop, but magnetic leakage will occur after the open-loop, reducing the magnetic flux detected by the magnetic field sensor, so an appropriate opening size should be designed. The size of the magnetic field sensor is 5 mm\*5 mm\*1 mm, the opening width of the magnetic ring should be at least 6 mm, the thickness of the magnetic ring should not be less than 5 mm, and the height of the magnetic ring should not be less than 1 mm, so as to ensure that the entire side of the magnetic field sensor is in the magnetic field. The inside of the opening of the ring (Lei, Lei, Zhou).

#### 3.2.2 Simulation of Inner Diameter Change of Magnetic Ring

Wrap two turns of wire around the magnetic ring to increase its sensitivity. It is determined that the inner diameter of the magnetic ring is 16 mm, and the minimum thickness of the magnetic ring is 5 mm, so the minimum outer diameter of the magnetic ring is 26 mm (Macrelli, 2014). As a measuring element, the magnetic field sensor does not contain magnetic permeability

components and has little effect on the measured magnetic field, so its relative magnetic permeability is set to 1. The established simulation model is shown in Figure 2.

Set the parameterized scanning range, the scanning range of the limited inner diameter is from 16 to 32mm, that is, the radius is from 8 to 16mm, and the scanning step is 2 mm. According to the theoretical analysis, when the inner diameter of the magnetic core increases, the magnetic path length of the magnetic core will increase, resulting in a decrease in the sensitivity of the sensor. The simulation results of parametric scanning of the inner diameter of the magnetic core are drawn as shown in Figure 3A.

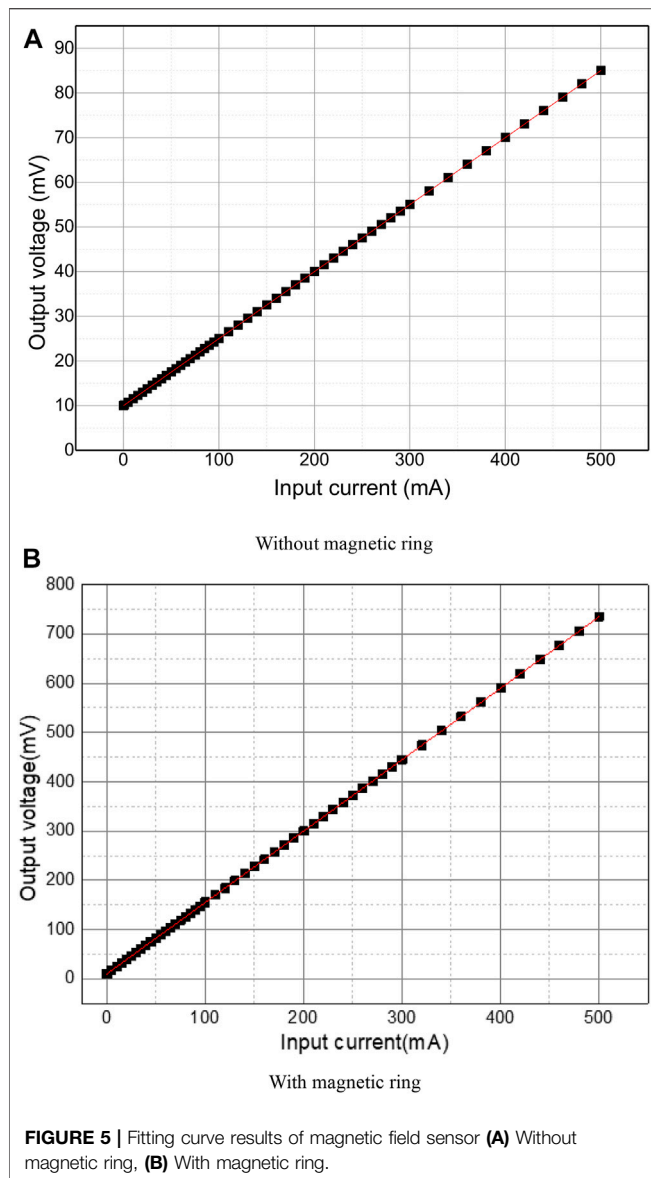
It can be seen from the figure that after the thickness of the magnetic core is fixed, with the increase of the inner diameter of the magnetic core, the average value of the magnetic induction intensity at the cross-section of the magnetic field sensor gradually decreases. Therefore, when the thickness of the magnetic ring is constant, the smaller the inner diameter of the magnetic core, the better the magnetization effect, and the greater the magnetic induction intensity generated at the opening, so the value of the inner diameter should be set to 16 mm.

#### 3.2.3 Simulation of Thickness Change of Magnetic Ring

The difference between the inner and outer diameters of the magnetic ring is the thickness of the magnetic core. Therefore, to determine the outer diameter of the magnetic core, it is necessary to analyze the influence of the thickness of the magnetic core on the magnetic induction intensity at the opening.

According to theoretical analysis, the closer the current position is to the greater the magnetic induction intensity, there is magnetic flux leakage at the edge and less magnetic flux leakage at the center of the magnetic core opening, so the best position for the magnetic field sensor to be placed is in the center of the magnetic core.

Due to the existence of magnetic leakage, the overall magnetic induction intensity at the opening of the magnetic ring is much



lower than that inside the magnetic ring, while the magnetic induction intensity inside the opening is stronger than the outside, which is consistent with the previous theoretical analysis. The simulation results of 4 show that the sensor is best placed in the center of the opening of the magnetic core.

The scan range to define the core thickness is 5–10 mm in 1 mm steps. The distribution of the average magnetic induction intensity is analyzed, and the results are shown in **Figure 3B**. With the increase of the thickness of the magnetic ring, the average magnetic induction intensity of the section first increases slightly, and then begins to gradually decrease. As the thickness of the magnetic ring continues to increase, the size of the opening of the magnetic ring also increases, and the length of the magnetic circuit of the magnetic ring also increases, which will lead to more and more magnetic flux leakage, and the detected magnetic induction intensity will decrease.

The increase of the thickness of the magnetic core will only slightly increase the magnetic induction intensity detected by the magnetic field sensor, which basically reaches the maximum value when the thickness is 8 mm, that is, the outer diameter is 32 mm.

### 3.2.4 Simulation of the Height Change of the Magnetic Ring

Analyze the effect of the height of the magnetic ring on the magnetic field at the opening of the magnetic ring. According to the previous analysis, the magnetic ring should ensure that the magnetic field sensor is inside its opening, its height is at least 1 mm, and the scanning step length is 1 mm. The change of the average magnetic induction intensity on the cross section of the magnetic field sensor is plotted, and the simulation results are shown in **Figure 3C**:

In the process of changing the height of the magnetic ring, the value of the average magnetic induction intensity detected by the cross-section first increases slightly, and then begins to gradually decrease. This is because the magnetic field sensor is always located 1 mm away from the bottom of the magnetic core, and the opening edge of the magnetic core will have a certain magnetic flux leakage. As the core height increases, the magnetic field sensor moves away from the open edge of the core, so the average magnetic flux density increases slightly. However, as the height of the magnetic core increases, the opening area and magnetic flux leakage will gradually increase. As can be seen from the figure, when the height of the magnetic core is 3 mm, the average magnetic induction intensity of the cross-section has reached the maximum value, so the height of the magnetic core is set to 3 mm.

Through the above simulation, it can be determined that the parameters of the iron-based amorphous alloy magnetic ring are as follows: the inner diameter is 16 mm, the outer diameter is 32 mm, the thickness is 8 mm, and the height is 3 mm.

## 4 EXPERIMENTAL VERIFICATION

In order to obtain the frequency characteristic and sensitivity of the open-loop fluxgate sensor, the sensor was measured experimentally. The experimental test principle is shown in **Figure 4**. The output of the arbitrary signal generator is connected to a power amplifier, and the output of the power amplifier is connected to a current-limiting resistor, and a loop is formed through a magnetic ring to simulate the leakage of different waveforms and current values. The first channel of the oscilloscope uses the current clamp meter probe to collect the current waveform in the line, and the second channel collects the output voltage of the sensor.

In the experiment: adjust the arbitrary signal generator to output 0–10 kHz sine wave sweep frequency, adjust the resistance RL to make the current value in the line 30 mA. By outputting different DC current values, the sensitivity test is carried out on the magnetic field sensor with magnetic focusing ring and the ordinary sensor. The experimental data curve is shown in **Figure 5**.

Fitting the data in (Figure 5A), the obtained linear formula is  $y = 0.15x$ , that is, the sensitivity is 0.15 mV/mA, and the linearity error is 0.0205%; the same formula in (Figure 5B) is  $y = 1.54x$ , that is, the sensitivity is 1.54 mV/mA, and the linearity error is 0.0021%. The sensitivity of the magnetic field sensor after adding the magnetic ring is 15.4 times higher than that of the sensor without the magnetic ring, and the linear error is reduced by 0.0184%.

## 5 CONCLUSION

The iron-based amorphous alloy material is selected for the open-loop magnetic focusing ring proposed in this paper, and the inner diameter of the magnetic ring is 16 mm, the outer diameter is 32 mm, the thickness is 8 mm, and the height is 3 mm through simulation. The experimental results show that the fluxgate current sensor with the open-loop magnetic concentrating ring structure can detect the residual current with a bandwidth of 6.7 kHz, a sensitivity of 1.54 mV/mA, and a linearity error of 0.0021%, which can meet the accurate detection of complex residual current waveforms.

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## DATA AVAILABILITY STATEMENT

The original contributions presented in the study are included in the article/Supplementary Material, further inquiries can be directed to the corresponding author.

## AUTHOR CONTRIBUTIONS

WY: Conceptualization, Writing-Reviewing and Editing; XX: Writing-Original draft preparation, Investigation; QY: Supervision; JH: Visualization and contributed to the discussion of the topic.

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**Conflict of Interest:** Author XX was employed by the company Yunnan Electric Power Technology Co., Ltd. Authors QY, JH, and JW were employed by the company Yunnan Power Grid Co., Ltd.

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