



## Coordinated Control Strategy of DC Fault Ride-through for the WF Connected to the Grid Through the MMC-HVDC

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Aiming at the problems of DC fault isolation and power surplus in the HVDC system with large-scale wind farm (WF) integration after single-pole grounding fault, this article designs a modified current transfer modular multilevel converter (M-CT-MMC) topology with DC fault isolation and power dissipation functions. The DC fault current can be isolated through the coordination of each branch of the M-CT-MMC. In terms of surplus power consumption, the control mode switching strategy of the M-CT-MMC is designed to improve the power transmission capability of the non-fault pole and enable it to absorb surplus power independently. Furthermore, a coordinated control strategy of wind turbines and dissipation resistors is designed to absorb surplus power. With the advantages of fast response speed of wind turbines and reliable dissipation resistance, DC fault ride-through under different working conditions is realized. Finally, the effectiveness and feasibility of the coordinated control strategy for DC fault ride-through are verified.

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## INTRODUCTION

In the HVDC system with large-scale WF integration, when a short-circuit fault occurs at the DC side, the fault current will rise rapidly, and power surplus will occur (Lin et al., 2016; Song et al., 2018; Nian et al., 2020), which will lead to the overload of the converter station. Therefore, how to quickly isolate the fault current and dissipate the surplus power is the key issue to be resolved urgently to realize DC fault ride-through (Xue and Xu, 2014; Moawwad et al., 2016; Rizk-Allah et al., 2021).

At present, AC and DC circuit breakers are mainly used to isolate fault currents. However, the development of a hybrid DC circuit breaker is difficult and expensive (Liu et al., 2017; Xiaoguang et al., 2017). AC circuit breakers have poor speed (Li et al., 2017). Guo et al. (2017) designed asymmetric full-bridge submodules and relied on diode freewheeling to make partial capacitors reversible to achieve fault current suppression. The full-bridge submodules, clamped dual sub-modules, and hybrid sub-modules all can isolate DC fault (Li et al., 2016; Tao et al., 2016; Li et al., 2019; Li et al., 2021), but they have respective problems of high cost or complex control and voltage sharing, or the investment benefit is not clear. In addition, there are combined DC fault current transfer MMC topology with DC fault clearing capability. The topology works well in a pseudo bipolar system, but there is a DC bias problem in the bipolar system, which will improve the withstand voltage of the grounding switch and increase the number of devices.

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To deal with the problem of surplus power dissipation after failure, He et al. (2017) proposed a power transfer control strategy for a bipolar connection system, which can balance the surplus power after a fault, but it still needs to cut off the machine and load in the face of extreme conditions. Li et al. (2020) used the coordinated control between the WF and the MMC to actively reduce the output power of the WF by a fast communication method, but the influence of communication delay on response speed should be considered. Cao et al. (2019) designed an accurate switching dissipative resistance control strategy for the multiterminal bipolar system to absorb the surplus power after a fault and effectively improve the fault ride-through capability under the condition of excess power. At present, there is still not enough information to study the cooperative control strategy of wind turbine load-shedding and dissipation resistors absorbing surplus power based on the new topology.

The M-CT-MMC topology is designed for bipolar HVDC connected with the WF to realize DC fault isolation and power dissipation. Under the condition of giving full play to the active load-shedding capacity of wind turbines, a complete coordinated control strategy for DC fault ride-through is designed for self-absorption and non-self-absorption conditions. The bipolar M-CT-MMC control mode switching strategy, wind turbine

load-shedding control strategy, and dissipation resistors are used to realize the surplus power consumption, which can meet the needs of different surplus power consumption to realize DC fault ride-through. Finally, the effectiveness and feasibility of the coordinated control strategy for DC fault ride-through are verified.

## COORDINATED CONTROL STRATEGY OF DC FAULT RIDE-THROUGH FOR THE WF CONNECTED TO THE GRID THROUGH THE MMC-HVDC

# The DC System With WF Integration and M-CT-MMC Topology

#### The DC System With WF Integration

The system is shown in **Figure 1**. The M-CT-MMC is connected with a symmetrical bipolar metal return line. The DC line is the overhead line. The M-CT-MMC at the WF (WFMMC) side adopts constant frequency/AC voltage control to ensure voltage quality and frequency stability of the WF parallel point. The M-CT-MMC at the grid side adopts constant DC voltage/reactive power control to ensure voltage stability at the DC side and provide reactive power support for the AC grid at the receiving end.

#### M-CT-MMC Topology

The M-CT-MMC topology is shown in **Figure 1**. The normal working state of the M-CT-MMC is the same as that of the half-bridge MMC; only three auxiliary branches are added, which can isolate the fault current and dissipate the surplus power. The components of each branch are described below.

- 1) Isolating branch: It is composed of the isolating switch T composed of several diodes and IGBTs in series after an antiparallel connection and an ultrafast mechanical switch K.
- Line power dissipation branch: It is composed of freewheeling diodes D and an absorption resistor R<sub>1</sub>.
- 3) Bridge arm absorption branch: All bridge arm inductors of the MMC are equipped with an absorption branch at both ends, which is composed of the dissipation resistor  $R_2$  and bridge arm switch  $T_{arm}$  composed of *n* antiparallel thyristors in series. The leads between the two switches of each phase are connected to construct the AC flow path.

## Fault Current Isolating Control Strategy for the M-CT-MMC Topology

For **Figure 1** system, the isolating process of the M-CT-MMC fault current is introduced by taking mono-pole grounding fault as an example:

- 1) The fault occurs at  $t_0$ , the capacitor of the fault pole M-CT-MMC discharges rapidly, and the fault current rises sharply. With a delay of 1 ms, the fault pole M-CT-MMC is locked at  $t_1$ , and  $T_{arm}$  is triggered. All dissipative resistors are put into operation to reduce the fault current flowing through the converter valve, and the voltage after T is disconnected.
- 2) T is turned off at  $t_2$ , and the fault current is transferred to the line power dissipation branch and the bridge arm absorption branch.
- 3) After a short delay, the current flowing through submodules of the bridge arm and the isolating branch is reduced to zero, and K is turned off. After about 2 ms, at t<sub>3</sub>, the ultrafast mechanical switch action is completed to achieve physical isolation of the M-CT-MMC and fault point.
- 4) At *t*<sub>4</sub>, T<sub>arm</sub> has switched to the dissipation resistor adaptive switching control mode.

The M-CT-MMC can reduce the current stress of the bridge arm converter valve and can also provide breaking conditions for T to achieve effective isolation of DC faults. The three-phase neutral point scheme is constructed by the interconnection of bridge arm absorption branch outgoing lines, which saves the grounding investment and does not have the problem of bearing DC bias.

### Coordinated Control Strategy of Surplus Power Consumption

After fault isolation, the output power of the WF will be completely borne by the non-fault pole, which may lead to non-fault pole overload. The problem of power surplus must be solved.

#### Control Mode Switching Strategy of the M-CT-MMC

To improve the power transfer capability of the non-faulty pole M-CT-MMC of the WFMMC, the control mode switching strategy is designed as follows:

- The faulty pole M-CT-MMC control is switched to constant DC voltage/AC voltage control and put into operation again to maintain the voltage stability of the grid-connected bus of the WF.
- 2) The non-fault pole M-CT-MMC control is switched to the constant frequency/reactive power—voltage droop control (The reactive power reference value is set to 0 to improve the power transfer capability.)—to absorb the power of the fault pole and maintain the frequency stability of the WF side.

Under the condition that the surplus power can be completely transferred from the non-fault pole M-CT-MMC (self-absorbing condition), the WF and the receiving end system can quickly return to the normal state based on this method.

#### Control Strategy of Load-Shedding for Wind Turbines

When the non-fault pole M-CT-MMC cannot fully transfer surplus power (the non-self-absorption condition), it should be operated at full capacity. Surplus power is absorbed by wind turbine load-shedding and the dissipation resistors of the bridge arm.

The load shedding control strategy of the wind turbine is a coordinated control strategy between the M-CT-MMC and WF based on the idea of frequency control. Using the frequency control capability of the non-fault pole M-CT-MMC, we improve the frequency of the wind farm side accurately during fault crossing. Then, using the frequency response ability of the wind turbine, the accurate load shedding of the wind turbine is realized. The load-shedding power  $\Delta P_{WF}$  of the WF should meet the following equation:

$$\Delta P_{\rm WF} = P_{\rm WF} - S_{\rm N},\tag{1}$$

where  $P_{WF}$  is the output power of WF and  $S_N$  is the rated capacity of the non-fault pole M-CT-MMC.

According to the principle of primary frequency regulation of the synchronous generator, the equivalent adjustment coefficient K doubly fed induction generator (DFIG) is defined as follows:

$$K = \frac{\Delta f}{\Delta P_{\rm WF}},\tag{2}$$

where  $\Delta f$  is the frequency variation of the sending end system.

The expression of the maximum operating power  $P_1$  of the WF in the steady state and the output power  $P_{\min}$  when the rotor speed of the unit reaches the upper limit is shown in Eq. 3.

$$\begin{bmatrix} \boldsymbol{P}_1 = \boldsymbol{k}_{mppt}\boldsymbol{\omega}_1^3 \\ \boldsymbol{P}_{min} = \boldsymbol{k}_{min}\boldsymbol{\omega}_2^3 = \boldsymbol{k}_{min} \left(1.2\boldsymbol{\omega}_N\right)^3, \quad (3)$$

where  $k_{mppt}$  and  $k_{min}$  are the tracking coefficients of the corresponding power curve,  $\omega_1$  is the current rotor speed,  $\omega_2$ 



is the upper limit of rotor speed, and  $\omega_N$  is the rated rotor speed; usually  $\omega_2 = 1.2\omega_N$ .

DFIG operates at a maximum power point in the steady state. When the system frequency fluctuates, the output power of the wind turbine meets the following relationship:

$$P_{1} = P_{1} - \Delta P_{e} = k_{\text{mppt}} (\omega_{1} - \Delta \omega)^{3}$$
  
=  $k_{\text{mppt}} (\omega_{1}^{3} - 3\omega_{1}^{2}\Delta\omega_{1} + 3\omega_{1}\Delta\omega^{2} - \Delta\omega^{3}),$  (4)

When the speed adjustment range is not large, it can be considered that  $\omega_1 >> \Delta \omega$ . (4) is simplified:

$$\Delta P_{\rm e} \approx k_{\rm mppt} 3\omega_1^2 \Delta \omega = \frac{\Delta P_{\rm max}}{f_{\rm max} - f_N} \Delta f, \qquad (5)$$

where  $f_{\text{max}}$  is the peak value of the system's allowable operating frequency, the maximum is 50.5Hz;  $\Delta P_{\text{max}}$  is the limit frequency modulation power at the current wind speed, which can be expressed as the difference between  $P_1$  and  $P_{\min}$  at the wind speed.

$$\Delta \boldsymbol{P}_{\max} = \boldsymbol{P}_1 - \boldsymbol{P}_{\min} = \boldsymbol{k}_{mppt} \boldsymbol{\omega}_1^3 - \boldsymbol{k}_{\min} \left(1.2 \boldsymbol{\omega}_N\right)^3, \quad (6)$$

Substituting  $\Delta P_{\text{max}}$  into (6) and sorting it out, we can get

$$\Delta f = K \Delta P_{\rm e} = K \times k_{\rm mppt} 3\omega_1^2 \Delta \omega, \qquad (7)$$

$$K = \frac{f_{\text{max}} - f_{\text{N}}}{k_{\text{mppt}}\omega_1^3 - k_{\min}\left(1.2\omega_{\text{N}}\right)^3}.$$
(8)

When the system finally runs to a new steady-state point,  $\Delta P_{e}$  is the load shedding power of WF  $\Delta P_{WF}$ , which satisfies  $\Delta \omega \propto \Delta P_{WF} \propto \Delta f$  within the frequency adjustment range. The *K* value of the corresponding wind speed can be determined by using the current speed and relevant tracking coefficient of the wind turbine. Finally, combined with the frequency control of the non-fault pole M-CT-MMC, the precise load shedding of the WF can be realized. The specific implementation methods are as follows:

- 1) The non-fault pole M-CT-MMC determines  $\Delta f$  according to  $\Delta P_{WF}$  and *K*. By adding  $\Delta f$  to the frequency reference value  $f_N$  of the non-fault pole M-CT-MMC, the system frequency on the WF side is controlled to reach the new steady-state value  $f_N + \Delta f$ .
- 2) The WF monitors the system frequency f in real time. When  $f > f_N$ , it immediately enters the overspeeding load-shedding control mode. According to Eq. 1, the frequency modulation  $\Delta P_{WF}$  can be calculated. Combined with the functional relationship  $F(\mathbf{x})$  between the output power of the WF and

the speed of the WF in Eq. 5,  $\Delta \omega$  is obtained.  $\Delta \omega$  is superimposed on  $\omega_1$  to obtain a new rotational speed control command  $\omega_{ref}$  to realize the accurate load-shedding control of the WF. As shown in Figure 2, when  $f < f_N$ ,  $\Delta \omega = 0$ , it does not participate in the system power regulation. The WF continues to work in the maximum power tracking mode to improve the efficiency of wind energy utilization. In this process, the inherent delay of communication between the WF and M-CT-MMC is eliminated, and the load shedding response speed of the WF after the fault is improved.

This method does not require communication, which improves the response speed of the wind turbine load-shedding. And it can achieve accurate load-shedding and maximum power transmission, but the frequency threshold is small and the load-shedding capacity is small, which cannot meet the power dissipation demand of the WF under full-power operation.

#### Control Strategy of Dissipative Resistors Absorbing Surplus Power

The input capacity  $P_{hs}$  (neglecting loss) of dissipation resistors can be expressed as follows:

$$\boldsymbol{P}_{\rm hs} = \boldsymbol{P}_{\rm WF} - \boldsymbol{S}_{\rm N} = \boldsymbol{\lambda} \boldsymbol{P}_{\rm N} - \boldsymbol{S}_{\rm N}. \tag{9}$$

where  $P_N$  and  $\lambda$  ( $0 \le \lambda \le 1$ ), respectively, represent the rated power and output coefficient of the WF.

For bipolar HVDC connected by the WF, the rated capacity of the unipolar M-CT-MMC is usually half of the rated capacity of the WF. Therefore, according to **Eq. 9**, if  $\lambda \leq 0.5$ , the non-fault pole is used to transmit surplus power. If  $0.5 < \lambda \leq 1$ , the input capacity of dissipation resistors shall be greater than and close to  $P_{\text{HS}}$  to eliminate surplus power and reduce the influence on AC voltage and DC transmission power. Input dissipative resistance  $R_{\text{eq2}}$  can be calculated by **Eq. 10**. When  $\lambda = 1$ ,  $R_{\text{eq2}}$  is minimum, dissipative resistance  $R_2$  ( $R_{\text{eq2}|\lambda=1}$ ) is designed as  $MR_{\text{SM2}}$  (dissipation resistance of each group) in parallel, as shown in **Eq. 11**.

$$\boldsymbol{R}_{\text{eq}2} = \frac{\boldsymbol{U}_{\text{L}}^2}{(\boldsymbol{\lambda}\boldsymbol{P}_{\text{N}} - \boldsymbol{S}_{\text{N}})} = \frac{\boldsymbol{U}_{\text{L}}^2}{(\boldsymbol{\lambda} - 0.5)\boldsymbol{P}_{\text{N}}},$$
(10)

$$\boldsymbol{R}_{\mathbf{SM2}} = 2\boldsymbol{M}\boldsymbol{R}_{\mathbf{eq2}|\lambda=1},\tag{11}$$

where  $U_{\rm L}$  is the effective value of line voltage borne by dissipation resistors.

The dissipation resistors of the fault pole M-CT-MMC are cut adaptively according to the actual output coefficient of the WF to maintain power balance, as given in the formula **Eq. 12**.



$$\begin{cases} M_{\text{out}} = M & 0 \le \lambda \le 0.5 \\ M_{\text{out}} = M - \left\lceil \frac{(\lambda - 0.5)P_{\text{N}}}{\frac{0.5P_{\text{N}}}{M}} \right\rceil = M - \left\lceil 2M(\lambda - 0.5) \right\rceil & 0.5 < \lambda \le 1 \end{cases},$$
(12)

where  $M_{out}$  is the number of dissipation resistor groups to be removed after fault isolation; [] represents an upward integer operation.

#### Coordinated Control Strategy of Load-Shedding for Wind Turbine and Dissipative Resistors Absorbing Surplus Power

The basic idea of the coordinated control strategy based on wind turbine load-shedding and dissipation resistor absorbing surplus power is as follows: first, fault isolation shall be carried out according to the working process of the M-CT-MMC described in *Fault Current Isolating Control Strategy for the M-CT-MMC Topology section*. After the fault is isolated, the M-CT-MMC control strategy is switched, and the bridge arm absorption branch is switched to the adaptive cutoff mode. At the same time, the WF will also operate under load-shedding. The two strategies are carried out at the same time. With the increase of load-shedding power, the number of groups of dissipative resistance is increased to achieve power balance and fault ride-through. The flow block diagram of fault ride-through is shown in **Figure 3**.

## SIMULATION VERIFICATION

Based on the Matlab/Simulink simulation platform, the simulation model as shown in **Figure 1** is built. The WF is paralleled by two 500 MV A-rated DFIGs. The main parameters





of the M-CT-MMC can be found in the study by Li et al. (2021), where  $R_{SM2} = 550\Omega$ , M = 5. The effectiveness of the abovementioned strategy is verified by simulation. Figure 4 shows the simulation results of the DC fault ridethrough-coordinated control strategy.

## Simulation Verification of Self-Absorption Condition

The output power of the WF is set to 500 MW, and the positive and negative pole of the converter station shall bear 250 MW active power and 50MVar reactive power, respectively. At 8.0s, a unipolar metallic short-circuit grounding fault occurred at the DC outlet of the WF side. At  $t_1$ , the fault pole M-CT-MMC is isolated and triggers  $T_{arm}$ ;  $t_1$ - $t_2 = 0.4$  ms;  $t_2$ - $t_3 = 2$  ms; and  $t_3$ - $t_4 =$ 0.6 ms. When t = 8.0035s, the control mode of the M-CT-MMC is switched.

After the fault occurs, the submodule current and isolating branch current increase rapidly. After adopting the fault current isolating control, the fault current drops rapidly and transfers to the bridge arm absorption branch and the line energy dissipation branch. The branch current decays to 0 after 1.09 ms and within 50 ms, respectively. As shown in **Figure 4A**,**a**,**c**, the M-CT-MMC can effectively isolate the DC side fault. **Figure 4A**,**b** shows the voltage of  $T_{arm}$  without DC voltage bias, which verifies the effectiveness of the proposed three-phase neutral point scheme.

After the fault is isolated, the power output of the WF remains unchanged. WF output active power (500 MW) and system reactive power (100MVar) can only be borne by the nonfaulty pole M-CT-MMC, which leads to its overload operation. After switching the M-CT-MMC control, the faulty pole M-CT-MMC provides 0.2p.u. reactive power support for the system. It can avoid overload operation of the non-fault pole and ensure maximum transmission of active power at the same time, as shown in **Figure 4A,d,e**.

# Simulation Verification of the Non–self-absorption Condition

The output power of the WF is 772 MW. Under this condition, the proposed coordinated control strategy is compared with the method of only using dissipative resistance to absorb surplus power, and the simulation results are shown in **Figure 4B**. When the wind turbine load is reduced from 72 to 100 MW (maximum), the input capacity of dissipation resistors is 200 MW. Therefore, to reduce the frequency change of the sending end, the non-fault pole M-CT-MMC frequency reference value adopts 50.3 Hz.

Before control, all output power of the WF is injected into the non-fault pole to enable its active power reach 1.544pu. The non-fault pole overload is shown in **Figure 4B,a**. After control, the system frequency finally stabilizes from 50 to 50.3 Hz; load-shedding of the wind turbine is reduced by 72 MW; 200 MW dissipation resistors are put into operation. The remaining 500 MW power is borne by the non-faulty pole M-CT-MMC

to ensure the maximum transmission of active power and reduce the impact of the fault on the receiving end AC system, as shown in **Figure 4B,c**. When only dissipative resistors are put into operation, the output power of the WF is not changed; when three groups of dissipative resistors are put into operation, 300 MW active power is consumed, and the loss increases. The non-fault pole transmits 472 MW power and cannot operate at full load, as shown in **Figure 4B,b**, which increases the probability of load rejection. By comparison, the proposed coordinated control method has obvious advantages.

## CONCLUSION

Aiming at the problem of DC fault ride-through in the bipolar DC system with WF integration, the M-CT-MMC with fault isolation and power dissipation is designed, and the coordinated control strategy of surplus power consumption is proposed. The following conclusions are obtained through simulation:

- 1) In terms of fault isolation, the M-CT-MMC can effectively isolate the fault current and avoid DC voltage bias, which can also dissipate surplus power, transfer fault current, and provide a new discharge path for the bridge arm inductance current.
- 2) In terms of surplus power consumption, the bipolar M-CT-MMC control mode switching strategy, wind turbine load shedding control strategy, and dissipation resistors are used to absorb the surplus power, which can comprehensively play the rapidity of wind turbine load-shedding and the reliability of dissipation resistors to meet the demand of surplus power dissipation under different conditions, and effectively reduce the fault influence range and improve the ability of safe and stable operation of the system.

## DATA AVAILABILITY STATEMENT

The original contributions presented in the study are included in the article/Supplementary Material; further inquiries can be directed to the corresponding author.

## **AUTHOR CONTRIBUTIONS**

BZ: conceptualization. HL: methodology. PX: software. SJ: validation. LZ: formal analysis. YX: data curation. All authors contributed to the article and approved the submitted version.

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