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*CORRESPONDENCE Ali Ghadami, ali.ghadami@mech.sharif.edu

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Dynamic modeling and analysis of flexible micro-porous piezoelectric sensors applicable in soft robotics

Ali Ghadami^{1,2}*, Hamid Reza Mirdamadi^{2,3} and Hamideh Khanbareh⁴

¹Department of Mechanical Engineering, Sharif University of Technology, Tehran, Iran, ²Department of Mechanical Engineering, Isfahan University of Technology, Isfahan, Iran, ³Department of Control Engineering and System Analysis, Active Structures Lab, Universite Libre de Bruxelles, Brussels, Belgium, ⁴Department of Mechanical Engineering, Materials and Structures Centre (MAST), University of Bath, Bath, United Kingdom

With recent advances in system integration technologies, numerous efforts have been made to develop soft piezoelectric sensors for various engineering and healthcare applications. Using flexible and sensitive materials is crucial for designing soft sensors in order to maximize their efficiency and integrability. Micro-porous PU-PZT composite is a recently designed piezoelectric particulate composite material with an improved flexibility and piezoelectric voltage coefficient over common piezoelectric ceramics that makes it a promising candidate for application in soft sensors. In this study, we investigate the dynamic response and sensitivity of the micro-porous PU-PZT composite for applications in soft sensors in both 33 and 31 modes using energy methods. By using the effective field method, the microporous PU-PZT composite material properties were extracted and optimized based on the partially experimentally measured properties in order to get a complete picture of the properties of the material. In addition, the effects of changing the sensor geometry by varying the thickness and adding an extra layer between the piezoelectric layers are studied. Finally, a large area sensor based on micro-porous PU-PZT composite is simulated in finite element software, and the effect of several parameters on sensor's performance is investigated. Dynamic analysis of the sensor shows high sensitivity in both 31 and 33 modes which is a significant improvement compared to the commonly used bulk piezoelectric ceramics. This work has demonstrated that due to the high output voltage and structural flexibility of the microporous PU-PZT composite, a flexible large-area sensor would be a suitable choice for artificial skins and smart gloves.

KEYWORDS

large-area sensor, soft robotics, piezoelectric, smart skin, flexible sensors

1 Introduction

Piezoelectric materials are widely used in many applications as sensors and actuators due to their unique properties that generate electric charge in response to mechanical input and *vice versa*. Application of piezoelectric sensors ranges from pressure and vibration sensors (Chuang et al., 2008; Shih et al., 2017; Yaghootkar et al., 2017; Kim and Kim, 2018) to sensors applicable in minimally invasive surgery (Wang et al., 2011), monitoring human health (Yeo and Lim, 2016), and active vibration control (Biglar et al., 2014; Biglar and Mirdamadi, 2014; Biglar and Mirdamadi, 2016). Touch sensors and robotics applications have benefited greatly from piezoelectric sensors due to their durability, sensitivity, and affordability.

In particular, with the recent advances in soft robotics, piezoelectric sensors have become an inseparable part of the designed robots. Soft robotics is an evolving field and aims to design robots composed of soft and stretchable materials for applications, including handling delicate materials using soft grippers and safe human-robot interaction, which are not possible using traditional rigid robots. The solid-state sensors traditionally used in robotics cannot capture the highdimensional deformation of soft systems (Thuruthel et al., 2019). Embedded soft sensors are an appropriate solution for this approach because they are thin, flexible, cover a large area like the human skin, and resist external forces and chemical pollution (Chuang et al., 2008). According to Dr. Mandayam A. Srinivasan, founder of the MIT Touch Lab, an ideal soft sensor "essentially has to be what skin is: flexible, with the ability to sense dynamically and with high spatial resolution, and physically robust-it shouldn't break, it shouldn't wear out." (MIT Technology Review J, 2006).

Given that flexibility and sensitivity are two important factors affecting the functionality of the sensors, composite materials are promising candidates for fabricating a sensor that satisfies our needs. Composite piezoelectric materials are chosen because one can optimize their electrical and mechanical properties as desired for a particular application. In addition, properties that cannot be achieved by a single material can be achieved with composite materials. Due to the need for embedding sensors in flexible structures, the demands for soft sensors have been increased over recent years, and several approaches have been investigated to reach this flexibility. For instance, in a research performed by Kim et al. (2020), soft gloves with integrated piezoelectric sensors have been developed to estimate hand motion for human-machine interaction applications. A PVDF film was used as a sensor, and the sensor was fabricated with a uniform beam structure with thin PVDF and Mylar layers. The glove was fabricated using Ecoflex 00–30 silicone rubber. Finally, using deep learning, the motion of the hand was predicted with sensors output voltage and used in a virtual reality interface system. Using discrete hard PZT ceramics in a flexible substrate can be another way to reach flexibility in the sensor structure. Acer et al. (2015) presented a solution by embedding PZT elements in an elastic silicon substrate that provided enough flexibility and softness and could even be used as a wearable sensor for contact detection and force measurement.

Micro-porous PU-PZT composite is a novel soft piezoelectric sensor that was manufactured by dispersing PZT5A4 ceramic powders (mean particle size of 1.8μ m) in a dense polymeric matrix (Smooth-on-Econ 80 polyurethane) and adding a third phase (gaseous) to the system. The presence of gaseous pores not only increases the sensor's flexibility, but also increases the piezoelectric voltage coefficient (g₃₃) by decreasing the dielectric permittivity. A further improvement in d₃₃ was achieved by dielectrophoresis (DEP), a process used to align piezoelectric particles and form long chains in an AC electric field, and an intermediate state between 0–3 and 1–3 connectivity was achieved. The g₃₃ coefficient was measured to be 165 mV·m/N which is a significant improve from the bulk PZT and PU composites (Khanbareh et al., 2017; Ali and Hamideh, 2021).

In this study, we consider the recently developed micro-porous PU-PZT composite material in a modeled multilayer circular sensor. The comprehensive properties of the micro-porous PU-PZT composite material required for dynamic analysis of the sensor response in the mathematical and COMSOL[®] model will be estimated using the effective field method (Glushanin et al., 2006) and partially available experimental data. The sensor resolution of the skins is an essential factor that should be taken into account. Using soft tissues and soft sensory materials and leveraging robust algorithms like neural networks, it will be possible to reach high sensing resolutions with limited sensors. As bending plays an essential role in soft materials sensing, and it is not investigated for the aforementioned material, it will be specially considered in this work, as both 33 and 31 modes can affect the output significantly.

First, we compare the micro-porous PU-PZT composites to other common piezoelectric materials in 33 mode to show its improvements as mentioned in the main paper. The functionality of the micro-porous PU-PZT composite material will be also investigated in the 31 mode as this mode plays an important role in soft sensors. Notably, the flexibility of the sensor is essential for integrating the sensor within a system, and the micro-porous PU-PZT composite has the highest flexibility among common materials used in sensor design. In addition, the effect of the piezoelectric layer and middle layer thickness on the performance of the micro-porous PU-PZT composite sensor is of interest which is beneficial to find the optimum rigidity and functionality of this sensor for the desired application. In the next step, as we aim to design a large area flexible sensor, a more complicated model of this sensor is developed and analyzed in COMSOL[®]. This theoretical simulation could be an important step in the practical implementation of this sensor and system.

2 Materials and methods

In this section, the formulation and procedures used in this study are described in detail. First, the constitutive equations of the piezoelectric materials are provided, and the relationship for output voltage in the 33 mode is calculated. Then the circular bending 31 sensor system is developed using energy methods. Finally, an approach for predicting the properties of the particulate piezoelectric materials is proposed.

2.1 Piezoelectric constitutive equations in 33 mode

Piezoelectric materials are widely used as sensors or energy harvesters in their 33 mode, meaning that the sensor is under stress in the poling direction and the induced voltage is in the same direction. Thus, it is essential to evaluate their performance in this situation. Here, it is desired to model a simple sensor mathematically and review the constitutive equations of the piezoelectric materials.

The relationship between applied force, the displacement, and induced electrical charge can be described as (Leo, 2007):

$$u_3 = \frac{ts_{33}^E}{A}F + d_{33}V, \quad q = d_{33}F + \frac{\varepsilon_{33}^TA}{t}V, \quad (1)$$

where,

Considering open circuit condition, we can simply set V to zero,

$$u_3 = \frac{ts_{33}^E}{A}F, \ q = d_{33}F, \tag{2}$$

Resulting in the output voltage as follows:

$$V = \frac{d_{33}t}{\varepsilon A}F = \frac{g_{33}t}{A}F.$$
 (3)

The term g_{33} , which is the piezoelectric voltage constant, directly affects the output voltage and can be used to compare the sensitivity of sensors composed of different piezoelectric materials.

2.2 Modeling a 31 bending piezoelectric sensor

In this section, we first derive an analytical solution for a circular laminated sensor composed of a piezoelectric material and investigate its performance under a dynamic excitation. In this mode, the sensor is under stress perpendicular to the poling direction and the induced voltage is in the poling direction. The relationship between applied force and output voltage as well as deflection will be evaluated. In addition, the resonant frequencies of the system are calculated.

2.2.1 Theory and hypothesis

We employ the classical plate theory (CPT) for the analyses (Reddy, 2017). Because only the small bending deformation of plates are considered, the displacement field in the absence of tension loads reduces to:

$$u_r = -z \frac{\partial w}{\partial r}, u_\theta = -z \left(\frac{1}{r} \frac{\partial w}{\partial \theta}\right), u_z = w(r, \theta, z).$$
(4)

With this displacement field, for small displacements, the only nonzero stains are: (Reddy, 2017)

$$\varepsilon_{rr} = -z \left(\frac{\partial^2 w}{\partial r^2} \right), \varepsilon_{\theta\theta} = -\frac{z}{r} \left(\frac{\partial w}{\partial r} + \frac{1}{r} \frac{\partial^2 w}{\partial \theta^2} \right),$$
$$2\varepsilon_{r\theta} = -\frac{2z}{r} \left(\frac{\partial^2 w}{\partial r \partial \theta} - \frac{1}{r} \frac{\partial w}{\partial \theta} \right).$$
(5)

Here we analyze a non-piezoelectric layer in the middle, sandwiched between two piezoelectric layers and assumed to have a simply supported constraint at r = a as shown in Figure 1. This middle layer can be used to adjust the rigidity of the sensor, or it can be seen as electrode mounted between sensors. This configuration also allows us to investigate using sensors that sandwich a tissue.

2.2.2 Dynamic 31 response

Given the material properties of the micro-porous PU-PZT composites, we aim to identify the relation between distributed dynamic load, deflection, induced voltage (sensor output), and resonant frequencies for the considered sensor geometry (as shown in Figure 1) and compare its dynamic response with sensors composed of other selected piezoelectric materials. According to the Classical Plate Theory (CPT) the effect of transverse normal stress is neglected and the output voltage is only dependent on the 31 mode. To perform a dynamic analysis on this system, we employ Hamilton's principle as well as the Ritz method as bellow:

$$\delta \int_{t_1}^{t_2} \left(T - U - W^{external} \right) dt = 0, \tag{6}$$

where T, U, and $W^{external}$ represent kinetic energy, potential energy, and external load potential, respectively. We can also define: (Leo, 2007)

$$U = \int \frac{1}{2} \vec{S}^{T} c_{s} \vec{S} \, dV_{middle} + \sum_{1}^{2} \int \left(\frac{1}{2} \vec{S}^{T} c^{D} \vec{S} - \vec{S}^{T} h \vec{D} + \frac{1}{2} \vec{D}^{T} \beta^{S} \vec{D}\right) dV_{piezo}$$

$$W^{ext-mechanical} = \int q_{0} \vec{u}(r) \, dA$$

$$(7)$$



$$W^{ext-electrical} = \sum_{i=1}^{2} v^{i} q^{i}$$
$$T = \sum_{1}^{3} \int \frac{1}{2} \left(\frac{\partial \vec{u}}{\partial t} \right)^{T} \left(\frac{\partial \vec{u}}{\partial t} \right) \rho^{i} dV,$$

In these equations, *i* represents the layer index, q_0 is force per unit area, *h* is piezoelectric constant, β^S is impermittivity, S is planar strain vector, and D is the electrical displacement vector. We approximate the displacement in the *z*-direction with a set of admissible shape functions, which satisfy the system's essential boundary conditions. In particular, the shape functions for this system are defined as:

$$u_z = \sum_{i=1}^N c_i \varphi_i(r, \theta), \tag{8}$$

where $\varphi_i(r, \theta)$ are polynomials generated by the Gram-Schmidt process as follows: (Bhardwaj et al., 2012)

$$\varphi_{i} = \frac{\Psi_{i}}{\sqrt{\langle \Psi_{i}, \Psi_{i} \rangle}}, \Psi_{1} = \left(1 - \left(\frac{r}{a}\right)^{2}\right).$$
(9)
$$\Psi_{i} = f_{i}(r, \theta)\Psi_{1} - \sum_{j=1}^{i-1} \frac{\langle \Psi_{1}f_{i}, \Psi_{j} \rangle}{\langle \Psi_{j}, \Psi_{j} \rangle} \Psi_{j}$$
$$f_{i}(r, \theta) = \left(\frac{r}{a}\right)^{m_{i}+n_{i}} \cos^{m_{i}}\theta \sin^{n_{i}}\theta,$$

where "a" is the radius of the sensor, $\langle f, g \rangle = \int f g dA$, and "n" and "m" are non-negative integers.

The strain-displacement relationship with respect to CPT and for cylindrical coordinates are:

$$S = L_{u}u(r) \rightarrow L_{u} = \begin{bmatrix} 0 & 0 & -z\left(\frac{\partial^{2}}{\partial r^{2}}\right) \\ 0 & 0 & -\frac{z}{r}\left(\frac{\partial}{\partial r} + \frac{1}{r}\frac{\partial^{2}}{\partial \theta^{2}}\right) \\ 0 & 0 & -\frac{2z}{r}\left(\frac{\partial^{2}}{\partial r\partial \theta} - \frac{1}{r}\frac{\partial}{\partial \theta}\right) \end{bmatrix},$$
(10)

According to Eq. 8, one obtains:

$$S = L_{u} \begin{bmatrix} 0 & 0 & 0 & \dots & 0 \\ 0 & 0 & 0 & \dots & 0 \\ \varphi_{1} & \varphi_{2} & \varphi_{3} & \dots & \varphi_{n} \end{bmatrix} \begin{bmatrix} c_{1} \\ c_{2} \\ c_{3} \\ \vdots \\ c_{n} \end{bmatrix} = B_{r}\vec{c},$$
(11)

-

where generalized coordinates are related to the strain vector. By this simplification, we can rewrite the terms for energy and work as follows:

$$U = \frac{1}{2}\vec{c}^{T}K_{middle}\vec{c} + \frac{1}{2}\vec{c}^{T}K_{piezo}^{D}\vec{c} - \vec{c}^{T}\theta q + \frac{1}{2}q^{T}C_{p}^{S^{-1}}q \qquad (12)$$
$$W^{ext-mechanical} = \int \vec{q_{0}}N_{r}(r)dA\vec{c}$$
$$W^{ext-electrical} = B_{v}v\vec{q}$$
$$T = \frac{1}{2}\vec{c}^{T}M\dot{c}.$$

Substituting Eq. 12 into the Hamilton principle and using proper shape functions, the governing equation for the system are obtained as follows,

$$(M_1 + M_2 + M_3)\vec{c} + \left(K_{piezo}^D + K_{middle}\right)\vec{c} = \theta\vec{q} + \int N_r^T(r)q_0^T dA$$
(13)

and,

$$-C^{S^{-1}}\vec{q}+\theta^T\vec{c}+B_v\nu=0,$$

where M, k, $C^{S^{-1}}$, and θ are mass, stiffness, the inverse of capacitance, and coupling matrices of the structure which are expressed below. In addition, $N_r(r)$ represents the interpolation matrix and B_v is set as $B_v = [1 - 1]^T$.

$$M_{i} = \sum_{i=1}^{3} \int N_{r}^{T} N_{r} \rho^{i} dV \qquad (14)$$

$$K_{piezo}^{D} = \int B_{r}^{T}(r) c^{D} B_{r}(r) dV_{piezo}$$

$$K_{middle} = \int B_{r}^{T}(r) c_{s} B_{r}(r) dV_{middle}$$

$$C^{S^{-1}} = \sum_{i=1}^{2} \int B_{q}^{iT} \beta^{S} B_{q}^{i} dV$$

$$\theta = \sum_{i=1}^{2} \int B_{r}^{T}(r) hB_{q}^{i} dV$$

The eigenvalues of the matrix $(M_1 + M_2 + M_3)^{-1} (K_{piezo}^D + K_{middle} - \theta C^s \theta^T)$ are the squared values of the natural frequencies of the system. To investigate the system's dynamic response, we analyze the system in the frequency domain by applying the Laplace transform to Eq. 13 and obtaining the transfer function between deflection and force as well as voltage and force (at

r = 0). In addition, a damping term of $D = \alpha M + \beta K$ has been added to the system. The desired transfer functions are obtained as follows:

$$T_{uf} = [0\ 0\ 1]N_r(0)((M_1 + M_2 + M_3)s^2 + (K^D_{piezo} + K_{middle}))$$
$$+Ds - \theta C^s \theta^T)^{-1} \int N_r^T [0\ 0\ 1]^T dA \qquad (15)$$
$$T_{vf} = (B_v^T C^s \theta^T ((M_1 + M_2 + M_3)s^2 + (K^D_{piezo} + K_{middle})))$$
$$+Ds - \theta C^s \theta^T)^{-1} \int N_r^T [0\ 0\ 1]^T dA)/(2C)$$

It is notable that in the presence of the middle substrate, the output voltage is not simply proportional to a term which is purely related to material's properties, like what we observed in 33 mode which was g_{33} . In this case, for a static load, the output voltage would be described as follows:

$$V \propto \left| \frac{h_{13}}{\varepsilon_{33}^T \left(\bar{C} - \bar{h}^2 \right)} \right|,\tag{16}$$

where:

$$\begin{split} \bar{C} &= 8\beta_{33}^{S}C_{11}^{D(piezo)} + 8\beta_{33}^{S}C_{12}^{D(piezo)} + 12n\beta_{33}^{S}C_{11}^{D(piezo)} \\ &+ 12n\beta_{33}^{S}C_{12}^{D(piezo)} + 6n^{2}\beta_{33}^{S}C_{11}^{D(piezo)} + 6n^{2}\beta_{33}^{S}C_{12}^{D(piezo)} \\ &+ 6n^{3}\beta_{33}^{S}C_{11}^{(substrate)} + 6n^{3}\beta_{33}^{S}C_{12}^{(substrate)} \\ &\bar{h}^{2} = 24h_{13}^{2}n + 12h_{13}^{2} + 12h_{13}^{2}n^{2} \end{split}$$

And *n* is defined as:

$$n = \frac{t_{substrate}}{t_{piezo}}.$$

2.3 Prediction of effective properties of piezo-particulate composites

In order to evaluate the functionality of the sensors, it is critical to estimate the electrical and mechanical properties of the new material accurately. Here, we employ formulations developed based on the effective field method (Glushanin et al., 2006) to effectively estimate the mechanical and electrical properties of sensors composed of composite materials. This part aims to estimate piezo-particulate composites' electrical and mechanical properties.

There are several approaches to predict piezo-particulate composite properties. In this work, we use the results of a micromechanics approach known as the Mori-Tanaka mean-field theory (Khanbareh et al., 2019). Consider a composite material composed of particles with 0–3 connectivity (single isolated inclusions) surrounded by a matrix. It is assumed that the shape of inclusions in Cartesian coordinates is described as follows:

$$\left(\frac{x_1}{a_1}\right)^2 + \left(\frac{x_2}{a_2}\right)^2 + \left(\frac{x_3}{a_3}\right)^2 = 1,$$

where aspect ratio of the piezoelectric particles can be defined as $\rho = \frac{a_1}{a_3}$. The effective electromechanical properties matrix of a composite is defined as follows: (Khanbareh et al., 2019)

$$C_{9\times9}^{*} = \begin{bmatrix} C_{6\times6}^{*E} & e_{6\times3}^{*T} \\ e_{3\times6}^{*} & -\varepsilon_{3\times3}^{*S} \end{bmatrix},$$
(17)

where C^{*E} is the composite's stiffness matrix at the constant electrical field, ε^{*S} is the effective dielectric permittivity matrix at constant mechanical strain, and e^* is piezoelectric coefficient matrix of the composite. For a transversely isotropic piezoelectric material, the property matrices have the following forms: (Huang and Kuo, 1996)

$$C = \begin{bmatrix} C_{11} & C_{12} & C_{13} & 0 & 0 & 0 \\ & C_{11} & C_{13} & 0 & 0 & 0 \\ & C_{33} & 0 & 0 & 0 \\ & & C_{44} & 0 & 0 \\ & & & C_{44} & 0 \\ & & & & \frac{C_{11} - C_{12}}{2} \end{bmatrix}$$

$$e = \begin{bmatrix} 0 & 0 & 0 & 0 & e_{15} & 0 \\ 0 & 0 & 0 & e_{15} & 0 & 0 \\ e_{31} & e_{31} & e_{33} & 0 & 0 & 0 \end{bmatrix}$$

$$\varepsilon = \begin{bmatrix} \varepsilon_{11} & 0 & 0 \\ 0 & \varepsilon_{11} & 0 \\ 0 & 0 & \varepsilon_{33} \end{bmatrix}$$
(18)

0 0

Based on the effective field method, the matrix C* can be described as follows (Khanbareh et al., 2019)

$$C^* = C^{(2)} + m (C^{(1)} - C^{(2)}) [I + (1 - m) S C^{(2)^{-1}} (C^{(1)} - C^{(2)})]^{-1},$$
(19)

where superscript (1) denotes properties for piezo particles and superscript (2) corresponds to polymer properties. m is the volume fraction of piezoelectric particles, and S contains the electrostatic Eshelby tensor components. Assuming matrix S is known, we can calculate the composite's effective properties using Eq. 19. Matrix S was explicitly defined by Huang with nine independent components as following: (Huang and Kuo, 1996)

$$S = \begin{bmatrix} S_{11} & S_{12} & S_{13} & 0 & 0 & 0 & 0 & 0 & S_{19} \\ S_{12} & S_{11} & S_{13} & 0 & 0 & 0 & 0 & 0 & S_{19} \\ S_{31} & S_{31} & S_{33} & 0 & 0 & 0 & 0 & 0 & S_{39} \\ 0 & 0 & 0 & S_{44} & 0 & 0 & S_{48} & 0 \\ 0 & 0 & 0 & 0 & S_{44} & 0 & S_{48} & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & S_{66} & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & S_{75} & 0 & S_{77} & 0 & 0 \\ 0 & 0 & 0 & S_{75} & 0 & 0 & 0 & S_{79} \end{bmatrix}$$
(20)

Parameter	Value	Parameter	Value	
$C_{11}^{E}[Mpa]$	1.496	d ₃₁ [<i>pC/N</i>]	-5.7*	
$C^{E}_{33}[Mpa]$	3.644	d ₃₃ [<i>pC/N</i>]	14.6*	
$C_{12}^{E}[Mpa]$	0.691	d ₁₅ [<i>pC/N</i>]	0.7	
$C_{13}^{E}[Mpa]$	0.727	ε_{33}	10*	
$C_{55}^{E}[Mpa]$	0.982	ε_{11}	3.68	

TABLE 1 Properties of the micro-porous PU-PZT composite.

*Experimentally measured.

Note that this matrix is dependent solely on the properties of the composite's matrix (i.e., electromechanical properties) and the aspect ratio of particles. The particles are uniformly poled along the OX_3 axis, and therefore the poling direction of the whole composite will be the same as particles. Also, the expressions for S_{ij} are extracted from (Huang and Kuo, 1996) and (Huang and Yu, 1994).

3 Results

The purpose of this section is to make use of the formulations provided before to evaluate the performance of recently developed micro-porous PU-PZT composite as a candidate for designing large area soft sensors and compare it to commonly used piezoelectric materials. Further investigations are done in COMSOL[®], and a closer model to a large area sensor is provided.

3.1 Property estimation

Performing the property estimation is essential for analyzing the system dynamics proposed in the following

sections, and the effective field method was reviewed and implemented. As we need complete properties of the microporous piezoelectric material to evaluate its functionality and it is not feasible to find all of them using experiments, we can use properties of PZT5A4 and polyurethane separately to estimate the full properties of composite with the procedure described in the previous section using the Mori-Tanaka mean-field theory. Here, the aspect ratio of the particles is a tuning parameter and can be found by fitting the d_{31} volume fraction diagram to the experimental measurements of d_{31} . For this purpose, we use gradient descent method for finding the best value of ρ by minimizing the mean square error between measured and piezoelectric approximated coefficients, i.e., $MSE = \frac{1}{2n} \sum_{n=1}^{n} (d_{31} - \hat{d}_{31})^2$ where d_{31} and \hat{d}_{31} are the output piezoelectric coefficient obtained by the model and the experimentally measured piezoelectric coefficient, respectively. The piezoelectric coefficient d_{31} for this sensor has been measured experimentally at selected volume fractions. Using the experimental data and the procedure mentioned above, the aspect ratio ρ that represents data the best is obtained to be $\rho = 0.1326$. Using the estimated ρ , the remaining properties of the composite material of the sensor are approximated by Eq. 19 and reported in Table 1, which is required for the analysis of sensor performance as discussed in the following sections and especially for further usage in a simulating software. The dielectric constant of the composites was determined using an Agilent 4263B and 16034E -Inductance Capacitance Resistance Meter (LCR) by the parallel plate capacitor method at 1 V and 1 kHz. The d₃₃ of the poled samples were determined using the Piezotest PiezoMeter System PM300 - Berlincourt d₃₃ m, under a 10 N static force and a 0.25 N dynamic force at a frequency of 110 Hz. At least three samples of each composite were tested. The predicted and ground truth values for d₃₁ and d₃₃ are also shown in Figure 2.



FIGURE 2

Diagrams of piezoelectric charge coefficient estimations. (A) Estimation of d_{31} with the tuned value of ρ , (B) Estimation of d_{33} with the tuned value of ρ .

Material	$\boldsymbol{\varepsilon}_{33}^{T}$	$d_{33} [pC/N]$	$g_{33} \ [mV.m/N]$	Young's modules [MPa]
10% PZT Micro-porous composite (Khanbareh et al., 2017)	10	14.6	165	0.97
PVDF (Xu and Kim, 2012)	12	-33	-310.6	2e3
PZT Nano-composite (Almusallam et al., 2017)	262	98	42.2	131
50% PZT-PDMS composite (Babu and de With, 2014)	40	25	70.6	10
Fe ²⁺ incorporated MAPbI ₃ (Ippili et al., 2018)	107	17	17.9	-
PZT 5H	3,800	650	19.3	49e3
PZT 5A	1800	390	24.5	54e3
BaTiO ₃ (Xu and Kim, 2012)	1700	149	9.9	67e3
BiScO ₃ – PbTiO ₃ – Mn (Lee et al., 2014)	1,450	360	28	-
KNN-KCN (Lee et al., 2014)	290	90	35.1	-
BNT-BKT-BT-Mn (Lee et al., 2014)	510	96	21.3	-
Mn:PINMNT (Lee et al., 2014)	3,000	900	33.9	-

TABLE 2 A comparison of piezoelectric materials in 33 mode.

TABLE 3 A comparison of piezoelectric materials in 31 mode.

Material	Output voltage (31 mode) [V]	Output voltage (33 mode) [V]	Output voltage (total) [V]	Fundamental resonant frequency [Hz]	Maximum deflection [m]	g_{31} [mV.m/N]	Density [kg/m ³]
PU-PZT Micro- porous composite	0.87 (-1.2%)	0.165 top -0.0825 bottom	1.035 top 0.7875 bottom	92 (-1.1%)	1.12e-2 (3.1%)	-64.4	871
PVDF	3.18 (0.6%)	-0.3106 top 0.1553 bottom	2.8694 top 3.3353 bottom	3,085 (0.9%)	4.62e-6 (-1.5%)	233	1780
PZT 5A	0.22 (-6%)	0.0245 top -0.0123 bottom	0.2445 top 0.2077 bottom	8,336 (2.7%)	1.43e-7 (-7.1%)	-9.8	7,900
PZT 5H	0.266 (-6.7%)	0.0193 top -0.0097 bottom	0.2853 top 0.2563 bottom	8,491 (3.5%)	1.45e-7 (-7%)	-9	7,500

*The errors compared to COMSOL® results are shown in parentheses.

3.2 Evaluation of micro-porous PU-PZTbased sensor in 33 mode

According to the formulations, the piezoelectric voltage constant (g_{33}), affects the output voltage directly and can be used as a parameter to compare the performance of sensors. Accordingly, several piezoelectric materials are compared in the case we are using them in the 33 mode and the results are shown in Table 2.

Results show that the PZT micro-porous composite material has a high g_{33} value compared to the conventional piezo ceramic materials, and this high voltage constant leads to a higher output voltage. This characteristic and the flexibility of the PZT micro-porous composites make it advantageous for many applications, including sensors and energy harvesters. Although PVDF has a higher voltage constant than the PZT micro-porous material, it

has a relatively lower curie temperature and much higher coercive field, which makes it a less interesting candidate for high-temperature application. In addition, PVDF has a higher susceptibility to piezoelectric degradation, and it is hard to control the crystallinity and poling for PVDF. As a result, the PZT micro-porous composite outperforms other available materials in such situations.

3.3 Evaluation of micro-porous PU-PZTbased sensor in 31 mode

Analysing the bending mode, which activates 31 mode of the sensor, is an essential step for designing piezoelectric sensors and actuators. In this section, the behaviour of a sensor under the bending forces is studied to evaluate the



FIGURE 3

The first three bending mode shapes of the system and the convergence plot of their corresponding natural frequencies based on the number of shape functions considered [ω (i, j) is mode shape with radial mode number "i" and circular mode number "j"]. (A) 1st mode = ω (0,1) with resonant frequency of 92 Hz, (B) 2nd mode = ω (1,1) with resonant frequency of 274 Hz, (C) 3rd mode = ω (2,1) with resonant frequency of 512 Hz.



efficiency of PZT micro-porous material compared to the other commonly used materials in fabricating sensors. We use the model developed in Section 2 and analyze the microporous PU-PZT composites. First, the system's output is calculated, and the structure's resonant frequency is obtained and compared to other materials. Further, the changes in the structure, such as the sensor's thickness, diameter, and the effect of an extra middle layer on the response are investigated.

3.3.1 Sensor's output and resonant frequencies

To perform simulations, we have considered sensors to have a diameter of 2 cm and a thickness of 0.5 mm for each layer (As Figure 1). The maximum output voltage of the sensors (each layer) is calculated using the obtained transfer functions and under harmonic pressure of 2 kPa and frequency of 20 Hz. The first fundamental frequency of the system, which is essential in many applications, is also identified for the system. The results of this analysis are shown



FIGURE 5

The effect of middle (metal) layer thickness on the (A) maximum deflection, (B) first resonant frequency, and (C) on output voltage of the sensor. These studies were done with a dynamic load of 2 kPa and 3 Hz with a constant radius of 1 cm and piezoelectric thickness of 0.5 mm. Here middle layer is aluminum.



FIGURE 6

The effect of sensor's radius on the (A) maximum deflection, (B) first resonant frequency, and (C) on output voltage. These studies were done with a dynamic load of 2 kPa (on a constant area) and 3 Hz and piezoelectric thickness of 0.5 mm.





Effect of middle layer thickness with different young's modulus on (A) maximum deflection, (B) first resonant frequency, (C) maximum radial/ circumferential stress, and (D) maximum voltage. These studies were done with a dynamic load of 2 kPa and 3 Hz with a constant radius of 1 cm and piezoelectric thickness of 0.5 mm. Additionally, the unit of young's modulus is Pa.

in Table 3; Figure 3. We observe that while the output voltage of the micro-porous PU-PZT composite is not the highest, it is a good improvement from the bulk PZT-5A. Also, we have a lower fundamental resonant frequency compared to other materials, which can be beneficial when designing a resonator or energy harvester using this material. The high value of g₃₁, on the other hand, shows its unique performance when it is used as a bending beam or with stretching forces. Note that the flexibility of the sensor plays a vital role in integrating any system with that, and the micro-porous PU-PZT composite has the highest flexibility among all listed sensors. The results are also validated using COMSOL®. Besides the 31 mode voltage that is calculated based on previously presented transfer functions in Eq. 15, the 33 mode output voltage is calculated by substituting maximum stress in each layer in the Eq. 3. So, the total output voltage will be the summation of the 31 and 33 voltages with respect to their sign. Due to the stress concentration in the edges, there will be a very high

compression stresses in order of roughly 15 times greater than the applied pressure that is not considered in theory. Thus, the COMSOL[®] output voltage result for 33 mode will be much larger than theory, and its results is not provided.

Other mode shapes of the system can be obtained using the developed formulations. For example, for the same configuration of the micro-porous PU-PZT composite, the results are shown in Figure 3.

Since the Ritz method is used, the convergence of the resonant frequencies depends on the number of considered shape functions. The effect of the number of shape functions on the solution is also investigated and provided in Figure 3. It is observed that the first resonant frequency converges after adding the 6th shape function, the second resonant frequency converges after adding the 10th shape function, and the convergence of third frequency demands inclusion of more than 15 shape functions in the calculations. Hence, if the values of the higher resonant frequencies are needed, one should use more shape functions for an accurate result.



3.3.2 Effects of structural parameters on sensor response

Parameters like the sensor's diameter, thickness, and presence of an extra middle layer can affect the sensor's output voltage and resonant frequency. So, in this part, the effects of these parameters in our 31 system with microporous PU-PZT composite material is investigated and the results are provided in Figures 4, 5, 6.

According to the results, by increasing the piezoelectric layer thickness, the maximum deflection of the sensor and the output voltage decreases, but the resonant frequency increases linearly. The same trend is observed when an extra layer is added between piezoelectric layers, but in this case, changes will occur with higher intensity, and the variation of resonant frequency increase is not linear. By increasing the sensor diameter while keeping the applied pressure constant, the maximum deflection and output voltage are increased while the resonant frequency is decreased nonlinearly. It is notable that in all of the simulations, the dynamic load was fully distributed on the surface. When the effect of radius was investigated, the area where the load was applied on that was constant to have constant total force.

The radial and circumferential stresses in the layers can be written as: (Reddy, 2017):

$$\sigma_{rr}(r,z) = -\frac{Ez}{(1-\nu^2)} \left(\frac{d^2 u_z}{dr^2} + \frac{\nu}{r} \frac{du_z}{dr} \right)$$
(21)
$$\sigma_{\theta\theta}(r,z) = -\frac{Ez}{(1-\nu^2)} \left(\nu \frac{d^2 u_z}{dr^2} + \frac{1}{r} \frac{du_z}{dr} \right),$$

where z is the distance from the neutral plane. The highest value for these stresses occurs at the center of the sensor and the farthest

distance from the neutral plane in the *z*-direction. This maximum stress is reduced when we increase the piezoelectric thickness or thicken the middle layer, as shown in Figure 7. This decrease in the maximum stress causes a reduction in the output voltage. In contrast, by increasing the diameter, the maximum stress is also increased.

The material of the middle layer can vary in different applications. Here, the effect of the young's modulus of the middle layer (when the middle layer material is isotropic) is discovered and shown in Figure 8. The results show the same trend as what was observed before, but for materials with higher young's modulus, the speed of changes increases dramatically.

3.4 Micro-porous PU-PZT composite as a large area soft sensor

As discussed and analyzed above, the flexibility of the microporous composite enables the usage of this material as a large area sensor in many applications, including artificial soft skins. This feature encouraged us to propose and analyze a large area sensor in COMSOL[®] for further investigations. This sensor comprises a bottom layer of micro-porous PU-PZT composite and discrete circular electrodes on the top surface of that. A thin layer of PDMS is placed on the whole set for a better sense of touch, protection, and isolation of the electrodes. The schematic of the sensor is shown in Figure 9.

To better simulate the skin, it is necessary to use appropriate boundary conditions. It is assumed that all lower edges of the microporous PU-PZT composite material have simply supported constraint. Also, a spring foundation is placed under the sensor



FIGURE 10

The effect of piezoelectric layer thickness (t1) on the output voltage. (Applied force has the same position as Figure 9 and with the magnitude of 2 N and soft layer thickness is 0.5 mm). (A) low voltage outputs, (B) high voltage outputs.

to mimic the skin's properties. Different values for the young's modulus of human skin have been reported. In ref (Boyer et al., 2009), the young's modulus of the forearm skin was said to be 13.2 kPa–33.4 kPa in a dynamic indentation test with indentation depth of 1 μ m–10 μ m/100 μ m–500 μ m. In this application, a spring foundation with young's modules of 20 kPa and a depth of 2 cm is selected.

This sensor configuration can be integrated with other structures due to its flexibility and provides an excellent resolution. In this structure, there is no need for many small electrodes for a good resolution. Due to the flexibility of the system, electrodes can sense the force in any location. Further increases in sensor's resolution can be achieved by using neural networks for detecting the touch location, which can be investigated in our future works.

Some parameters affect the functionality and output of the proposed sensor. Some of them are the thickness of the piezoelectric layer, the thickness of the soft layer, and circuit resistance. For the analysis, the force applied to the surface is circular with a diameter of 1 cm to induce a finger-like sensation. The simulations were done in the frequency domain, and the applied force is harmonic. Of course, the static touch can be detected too, but there are some problems when we have long-term contacts due to the discharge of the piezoelectric material. We assign a number to each electrode for the following analysis and apply force to a fixed location as in Figure 9.

The first parameter that is taken into account is piezoelectric layer thickness. As Figure 10, for electrodes near the applied force



FIGURE 11

The effect of (A) soft layer thickness, (B) spring foundation young's modulus, and (C) circuit resistance on the high output voltages. (Applied force has the same position as Figure 9 and with the magnitude of 2 N and for (A), the piezoelectric layer thickness is 1 mm, while for (B) and (C) piezoelectric layer thickness and soft layer thickness are 1 mm and 0.5 mm, respectively).

(electrodes 5 and 2), the output voltage will increase by increasing the piezoelectric thickness, but other voltages don't experience a notable change. In the previous part, it was observed that in the simpler mathematical model, by increasing the piezoelectric thickness, the 31 mode output voltage decreases. Also, according to Eq. 3, the 33 mode output voltage increases with increasing the piezoelectric thickness. Here, the electrode 2 is directly under the pressure. Consequently, increasing the thickness significantly increases the output voltage of the sensor (Figure 10) while electrode 5 does not experience much increase. We should mention that although increasing piezoelectric layer increases the effective output voltage, but a delicate and thin sensor is preferred for tactile applications.

The same analysis has been performed for the soft layer on the top of the piezoelectric layer as Figure 11A. The figure shows that by increasing the thickness of the soft layer, the output voltages increase significantly, but no notable trend was observed for electrodes far from the applied force.

Another parameter that can affect the outputs is the surface properties on which the sensor is mounted. This is considered in the model as the young's modulus for the spring foundation. The young's modulus for forearm skin was reported roughly between 10 kPa and 100 kPa (Boyer et al., 2007; Boyer et al., 2009; Khaothong, 2010). Figure 11B shows the effect of this parameter on sensor output. It is observed that for a more rigid surface, the output voltages will be reduced because of the lower bending deformation that occurs in the piezoelectric layer.

Another critical parameter is circuit resistance for piezoelectric sensors. This can affect the output voltage and power generated by an electrode. Figure 11C shows this dependence, and as we see, even though the voltage constantly increases with increasing the resistance, the power forms a peak at about 1,000 M Ω and then decreases.

4 Conclusion

This work has analyzed and evaluated the application of recently developed micro-porous PU-PZT composite material as a soft sensor. We studied the dynamic response of a PU-PZT sensors in 33 and 31 modes and compared it with other commonly utilized piezoelectric materials. The comprehensive properties of the micro-porous PU-PZT composite material required for dynamic analysis of the sensor response were estimated using the effective field method and partially available experimental data, and the optimum value for particles aspect ratio was identified as 0.1326. These properties were used further in our mathematical and finite element model. The functionality of the micro-porous PU-PZT composite material was investigated and it was observed that the output voltage of the micro-porous PU-PZT composite exhibits 295% improvement compared to the output voltage obtained from the bulk PZT-5A in a similar loading condition in 31 mode. In

addition, compared to the other materials, micro-porous PU-PZT composite has a lower fundamental resonant frequency, which is an important factor for designing resonators and mechanical energy harvesters. Notably, the flexibility of the sensor is crucial for integrating the sensor within a system, and the micro-porous PU-PZT composite has the highest flexibility among common materials used in sensor design. The effect of the piezoelectric layer, sensor diameter, and middle layer thickness on the performance of the PU-PZT based sensor was also investigated. Motivated by the application of particulate PU-PZT composites in large area sensors, a sensor consisting of discrete electrodes on a micro-porous PU-PZT composite layer was developed and analyzed in COMSOL®. Results of this study show that the micro-porous PU-PZT composite is a great candidate for applications in soft sensors, especially large area soft sensors, due to its flexibility and high sensitivity. This analysis provides a baseline for designing novel soft sensors using the micro-porous PU-PZT composite material for flexible, large area sensing implementations.

Data availability statement

The original contributions presented in the study are included in the article/supplementary material, further inquiries can be directed to the corresponding author.

Author contributions

AG performed the simulations, analyzed the data, and wrote the manuscript. HM and HK supervised the research and interpreted the results. All authors reviewed and approved the manuscript.

Conflict of interest

The authors declare that the research was conducted in the absence of any commercial or financial relationships that could be construed as a potential conflict of interest.

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