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# Single-photon counting detectors for diffraction-limited light sources

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When first introduced, single-photon counting detectors reshaped crystallography at synchrotrons. Their fast readout speed enabled, for example, shutter-less data collection and fine slicing of the rotation angle and boosted the development of new experimental techniques like ptychography. Under optimal conditions, single-photon counting detectors provide an unlimited dynamic range with image noise only limited by the Poisson statistics of the incoming photons. Counting the pulses from individual photons, essentially what made the detectors so successful, also causes the main drawback, which is the loss of efficiency at high photon fluxes due to pulse pileup in the analog front end. To fully take advantage of diffraction-limited light sources, the next-generation single-photon counters need to improve their count rate capabilities in the same order of magnitude as the increased flux. Moreover, fast frame rates (a few kHz) are required to cope with the shorter dwell time achievable, thanks to the higher flux. Detector architecture with multiple comparators and counters can open new possibilities for energy-resolved imaging, while interpixel communication can overcome the issues arising from charge sharing and reduce the loss of efficiency at the pixel corners. Coupling single-photon counting detectors to high-Z sensors for hard X-ray detection (> 20 keV) and to low-gain avalanche diodes (LGADs) for soft X-rays is also necessary to make use of the increased coherence of the new light sources over the full radiation spectrum. In this paper, we present possible strategies to improve the performance of single-photon counting detectors at the fourth-generation synchrotron sources and compare them to charge integrating detectors.

## KEYWORDS

hybrid detectors, x-ray detectors, pixel detectors, single-photon counting, synchrotron

## 1 Introduction

The idea of hybrid pixel detectors originated around the end of 1980s from within the particle physics community [1]. The first fully functional hybrid pixel detector was tested in the Omega-Ion experiment (WA94) in 1991 [2], and larger installations soon followed [3; 4]. Scientists realized early that hybrid detectors also would be a good fit for X-ray imaging [5; 6], and several groups [7; 8] including PSI Brönnimann et al. [9] started working on dedicated detectors. The development of single-photon counting (SPC) hybrid pixel detectors at PSI was motivated by the needs of diffraction applications at the Swiss Light Source, and most of the know how came from working on the original pixel chip for the CMS experiment at the LHC [10]. The PILATUS 1-M detector was the first large-

area SPC detector dedicated to macromolecular crystallography [11]. It comprised 18 multi-chip modules for a total of 288 readout chips covering  $21 \times 24 \text{ cm}^2$ .

Since their introduction, single-photon counting detectors have transformed data collection at synchrotrons. Replacing slower and less-sensitive detectors, they enable, for example, fine slicing of the rotation angle and shutter-less data collection [12; 13; 14] and now underpin many modern measurement techniques like ptychography [15]. Currently, most scattering beamlines at synchrotrons rely on SPC detectors like PILATUS [16; 17], EIGER [18; 19], XPAD3S [20], UFXC32k [21; 22], MEDIPIX [23; 24; 25], and MYTHEN [26]. SPC detectors can also be used for electron detection, for example, in low-energy electron microscopy and photoemission electron microscopy [27].

A pixel in a typical single-photon counting detector consists of a charge-sensitive preamplifier, shaper, comparator, and counter. When the analog signal exceeds a certain threshold, it is counted as a photon. If the threshold is high enough compared to the electronic noise (at least five times the RMS) and the photon energy is similarly higher than the threshold, it is possible to detect photons with a high efficiency and extremely good noise rejection [28]. Given that when a photon is absorbed between two pixels, its charge is shared between them, the threshold should be set at half of the photon energy in order to maximize the number of detected X-rays while avoiding double counts [16]. The electronic noise defines the minimum detectable energy at about 10 times the RMS of the electronic noise [26].

Another important parameter is the threshold dispersion, i.e., the accuracy of tuning the threshold at the same energy level for all the channels. Threshold equalization methods must be implemented to compensate for mismatches between channels in the analog chain, affecting mainly the gain and the baseline level. Since the threshold is normally set as a voltage, the threshold equalization circuitry is then implemented by adding an additional *trim* threshold to the global threshold common to all channels of the chip (or of the detector module). The *trim* digital-to-analog converter (DAC) has a resolution ranging between 3 and 6 bits, and usually, the range of the voltage available for trimming can be tuned by using an external voltage, as in [29]. Recently, trimming architectures capable of tuning the gain and offset of the analog chain by acting at the shaper level have been demonstrated by [30]. Threshold equalization strategies can either be implemented by optimizing the count dispersion using a flat illumination, as in [31], or by equalizing the gain at a certain energy, as in [32]. The energy resolution of the detector, i.e., the accuracy to define the threshold to discriminate photons of different energies, is given by the quadratic sum of the electronic noise and of the threshold dispersion [33]. This can be exploited, for example, in the case of fluorescence radiation emitted by the sample, which can be rejected by setting the threshold between the main beam energy and the fluorescence line.

Since SPC detectors do not provide any information about the energy of the photons, their performance is not ideal in case of a polychromatic radiation spectrum. However, also in this case, by assigning the same weight to each detected photon, their performance outdoes charge-integrating detectors, where the weight of the photons is proportional to their energy, resulting in a reduction of the image contrast [34] or in stronger high harmonic contamination in the case of diffraction applications.

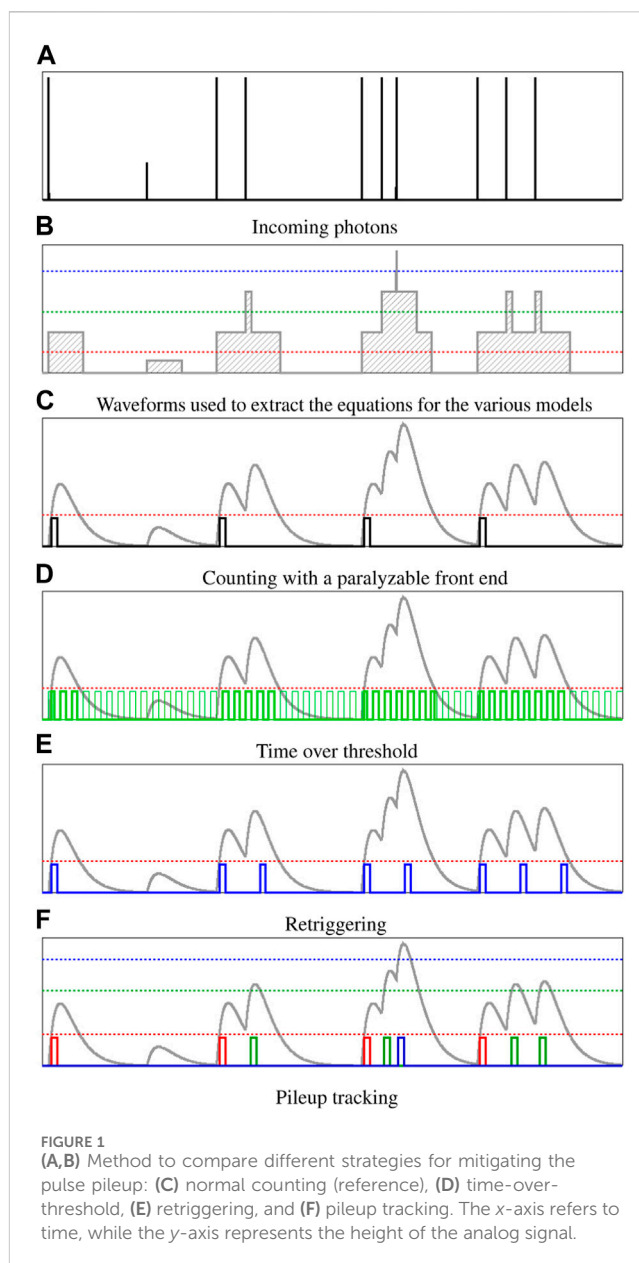


FIGURE 1 (A,B) Method to compare different strategies for mitigating the pulse pileup: (C) normal counting (reference), (D) time-over-threshold, (E) retriggering, and (F) pileup tracking. The x-axis refers to time, while the y-axis represents the height of the analog signal.

The readout of the detector is completely digital, and it does not add any noise to the data. Therefore, in ideal conditions, SPC detectors provide *noiseless* data, where the image quality is only limited by the Poisson fluctuations of the number of incident photons. SPC detectors can provide perfect linearity and virtually infinite dynamic range, only limited by the exposure time since it is possible to sum frames without adding noise. Lower bit depths or partial readout of the counters allow extremely high frame rates, for example, 20 kfps for EIGER [35] and 56 kfps for UFXC32k [21], while data compression can help achieve frame rates higher than 100 kHz [36]. A further improvement in speed can be achieved by defining a region of interest [37].

Using an electronic shutter, SPC detectors can be gated with a time resolution usually of a few tens of ns. This is of particular interest for pump-probe experiments since the signal from multiple probes can be accumulated at high frequencies without the need to

read out the detector in between the gates. With the hybrid fill pattern of synchrotrons, it is possible to gate the isolated bunch and obtain a time resolution limited by the duration of the bunch and by its jitter [21; 38].

Despite their success, single-photon counters still suffer from one major inherent weakness since the pulse processing front-end makes them susceptible to a loss of efficiency due to the pileup (Figure 1C). For this reason, SPC detectors require significant improvements in order to best use the increased brilliance of the fourth-generation synchrotron sources. In the following section, we will discuss the main ideas behind the development of the next-generation SPC detectors and look at their expected performance compared to charge-integrating detectors, initially developed for X-ray free-electron lasers (XFELs) but increasingly also used at synchrotrons. We will investigate the count rate capability (i.e., the linearity of the detector as a function of the number of photons per second), the spatial resolution, and strategies for extending the detectable energy range toward both lower- and higher-energy photons using advanced sensors.

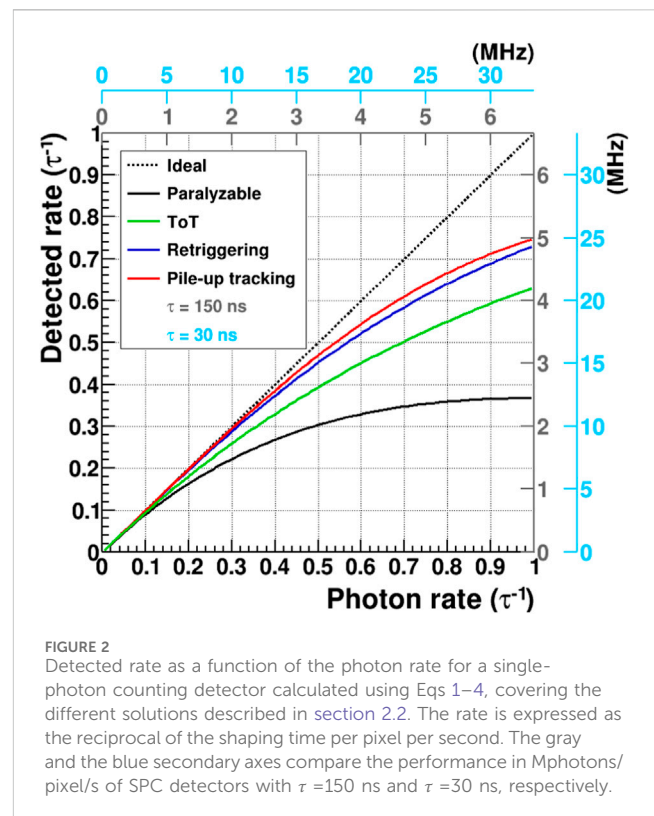
## 2 Discussion

The readout electronics of hybrid pixel detectors benefits from the advances in CMOS technology. By exploiting more advanced, smaller technology nodes with a higher transistor density, it is possible to integrate more functionalities in the pixel and target smaller pixel pitches, still with an acceptable power consumption. Moreover, by implementing advanced logic on-chip, taking advantage of the synthesized digital circuitry, the fully digital data can be processed in hardware to reduce the data throughput or speed up the readout.

In this section, we discuss the major trends in SPC detector development toward diffraction-limited light sources, namely, the possibility of having multiple comparators and counters per pixel with independent thresholds and enabling signals; the development of methods to reduce the loss of efficiency at high count rates; the goal of achieving a high spatial resolution either using smaller pixels or exploiting advanced inter-communication between pixels; and the possibility to combine SPC detectors with novel sensor technologies in order to cover an energy range spanning from soft to hard X-rays. Finally, we compare SPC with charge-integrating detectors, highlighting the strengths and weaknesses of both readout modes, and discuss the challenge of handling the data produced by fast large-area detectors.

### 2.1 Multiple comparators and counters

Thanks to the miniaturization of electronic components using advanced CMOS technologies, several comparators, each with an individual threshold, trimbits, and counter with independent gates can be allocated in the same pixel. The comparators and counters can be connected using the logic of varying complexity, and even inter-pixel communication can be implemented. The independent thresholds give access to different energy bins, the independent gates to different time windows.



The main application exploiting the independent thresholds is energy binning [39]. It can be used at polychromatic X-ray sources and is exploited in clinical CT systems [40]. However, energy binning can also be used at synchrotron beamlines for high harmonic and fluorescence suppression or detection. As long as the spectroscopic capabilities do not degrade at high fluxes, the energy discrimination capabilities can be used to operate in the pink beam mode, i.e., using a larger bandwidth without monochromatizing optics, to isolate the full undulator harmonic. The energy resolution depends not only on the noise and on the threshold dispersion but also on charge sharing. Therefore, larger pixel sizes and charge-sharing suppression methods (see section 2.3) are an advantage for energy binning. The multiple thresholds can also be exploited to improve the count rate capability, as described in section 2.2.3. Complex digital circuitry and inter-pixel communication can be used, for example, for interpolation to achieve a sub-pixel resolution, as highlighted in section 2.3.

The independent gates enabling the counters can be used to perform measurements in different time windows. Given that the reaction being studied happens on the same time scale, or slower, than the shortest gate that can be applied (normally tens of nanoseconds), it is possible to probe at multiple times and thus reduce the duration of the experiment proportionally to the number of counters compared to a single probe. Moreover, it is possible to acquire pumped and unpumped data alternating in time and, therefore, correct for possible low-frequency drifts of the system [41]. Ideally, a single comparator is connected to the multiple independent counters to minimize mismatches between different probes since the changes in the sample can be subtle (~1% or less).

## 2.2 Count rate capability

Single-photon counting detectors present a paralyzable behavior due to the pileup of the signal of photons arriving close to each other in time. This means that after detecting a photon, the detector is insensitive for a defined time (dead time), and a photon detected during this interval will not be counted and will also restart the dead time. As a consequence, with the increasing rate, the detector will reach a saturation point where it will be incapable of recording any event at all. To fully capitalize on the increased brilliance of the fourth-generation light sources, count rate capabilities up to 100 Mphotons/pixel/second are required. In the following section, we discuss and compare the new approaches that have been proposed to extend the usability of SPC detectors at next-generation synchrotron sources. They rely on the idea that when two or more photons pile up, the analog signal has a larger amplitude and longer duration. The working principle of various methods is shown in Figure 1, while Figure 2 shows the comparison of their performance in terms of count rate capability. The secondary axes quantify the performance in terms of impinging photon rate for 150-ns and 30-ns shaping times.

Usually, the count rate capability of an SPC detector is modeled according to the parameter  $\tau$ , which is proportional to the shaping time of the analog chain. It scales inversely with the shaping time of the analog signal, which is usually a compromise between the noise and gain of the detector. Therefore, the count rate performance tends to improve at high energies, where a lower gain can be used, and a loss of signal due to ballistic deficit can also be afforded. Another approach to obtain a fast shaping is active reset, as demonstrated by [42]. Faster shaping time and active reset cause a high power consumption, which is sometimes unacceptable for large readout chips with tens of thousands of pixels. Current single-photon counters exhibit a  $\tau$  of 30–150 ns, which translates to approximately 2–10 Mcounts/pixel/s. The KITE ASIC from DECTRIS has been designed for electron microscopy with extremely short signal pulses of 6 ns FWHM [36]. This is possible thanks to the fast settings and low gain allowed by the extremely large signal generated by high-energy electrons, as well as the small size with no buttability, which allows high power consumption and optimal power distribution.

It is important to highlight that the count rate capability of SPC detectors depends on the photon distribution, i.e., on the filling pattern of the light source [43]. The actual time structure varies between synchrotrons and can, in many cases, be tuned, from a few isolated bunches to the quasi continuous mode, in order to optimize for certain experiments. Filling patterns with fewer bunches spaced more than the shaping time of the analog signal usually allow better performance, the rate correction present requires a simpler calibration of the parameters, and the methods explained in the following are more effective since they are less subjected to statistical fluctuations in the photon time distribution. However, in this case, the equations reported for the count rate corrections must be modified due to the different photon distribution.

For modeling the count-rate corrections presented in this section, a Poisson-like photon distribution is assumed, with no charge sharing and  $\tau$ -wide rectangular pulse shape, where  $\tau$  approximates the time-over-threshold of the analog pulse, as shown in Figure 1B. For a simple SPC detector, the paralyzable

detector model applies for converting from the impinging photon rate  $\phi$  to the measured flux  $\varphi$  [44]:

$$\varphi = \phi e^{-\phi\tau} \quad (1)$$

Despite being a simplified description of the pileup, this approach allows to model analytically the corrections needed to convert the detected signal into the impinging photon rate. The real corrections will depend also on other factors, such as the fill pattern of the photon source, charge sharing, and—in particular—on the shape of the analog signal, which should be optimized depending on the strategy chosen to improve the count rate capability.

### 2.2.1 Time-over-threshold

The time-over-threshold (ToT) readout measures the time duration of the analog signal above the threshold by providing a clock to each pixel and incrementing the counter while the signal is above the threshold (Figure 1D). This method is usually applied to measure the energy of the impinging particle, as in the TIMEPIX detectors [45]. However, to improve the energy resolution, slow shaping times are usually implemented, in contrast with the requirements to obtain a high count rate capability in SPC detectors. ToT, as a means to improve the count rate capability at high fluxes, has been demonstrated with an improvement of a factor 3–6, depending on the settings [46]. The count-rate correction equation is expressed as follows:

$$\varphi = \phi \frac{1 - e^{-\phi\tau}}{\phi\tau} \quad (2)$$

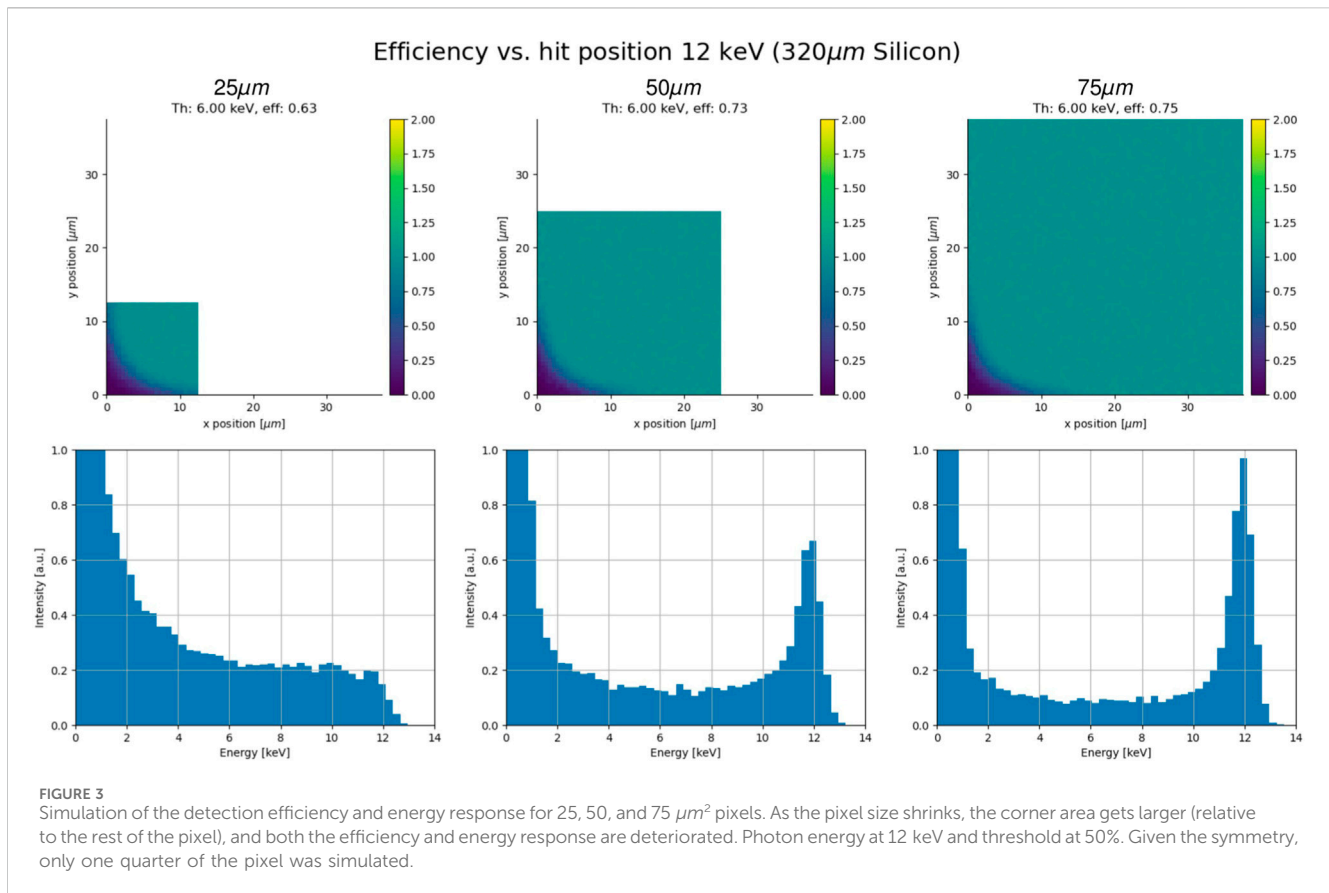
with little dependence on the frequency of the clock, as long as it is  $\geq 5\tau^{-1}$ . It is important to point out that in order to obtain the number of photons, it is necessary to normalize the counter value by the average number of counts per photon of that energy and threshold. The main disadvantage of the ToT approach is the distribution of the 10–100-MHz clock over the whole pixel matrix, which requires power and can generate the digital-to-analog crosstalk, increasing the noise. Optimized clock distribution solutions, such as the digital delay-locked loop (dDLL) [47], can help overcome these bottlenecks.

### 2.2.2 Retriggering

Retriggering was introduced using DECTRIS in the PILATUS3 detector [17] as a measure to increase the count rate capability and avoid the ambiguities of a paralyzable counter at extremely high photon rates. As the name implies, retriggering works by triggering an additional count after a certain time delay. The delay is started on the crossing of the threshold which gives retriggering a clear advantage over the non-synchronized ToT clock. Retriggering relies on knowing the pulse width for a single photon. The rate correction model is described in [48]:

$$\varphi = \frac{\phi}{e^{-\phi\tau} + \phi\tau_r} \quad (3)$$

where  $\tau_r$  is the time after which the retrigger is evaluated and an additional signal eventually generated. Ideally  $\tau_r = \tau$ , but  $\tau_r$  is usually larger to avoid double counts. The KITE ASIC from DECTRIS has been designed for electron detection, and it can support up to  $\sim 70$  Mcounts/pixel/s with a 43-keV



threshold for 200-keV electrons with a 10% loss of counting efficiency compared to approximately 20 Mcounts/pixel/s without retriggering [36].

### 2.2.3 Pileup tracking

Given a sufficient dynamic range of the front-end, another way to mitigate the loss of efficiency is to use multiple thresholds to count the pileup by placing the additional thresholds above the photon energy (e.g., 1.3 and 1.7  $\times E_{\text{photon}}$ , as shown in Figure 1F). This method is known as the *pileup trigger method* [49; 50] or *pileup tracking* [51]. Finding the right threshold depends on the analog shape of the signal and sensor geometry (charge sharing). Ideally, also, the higher thresholds are proportional to the energy, similar to the one at half energy. The count rate correction model can be found in [52]:

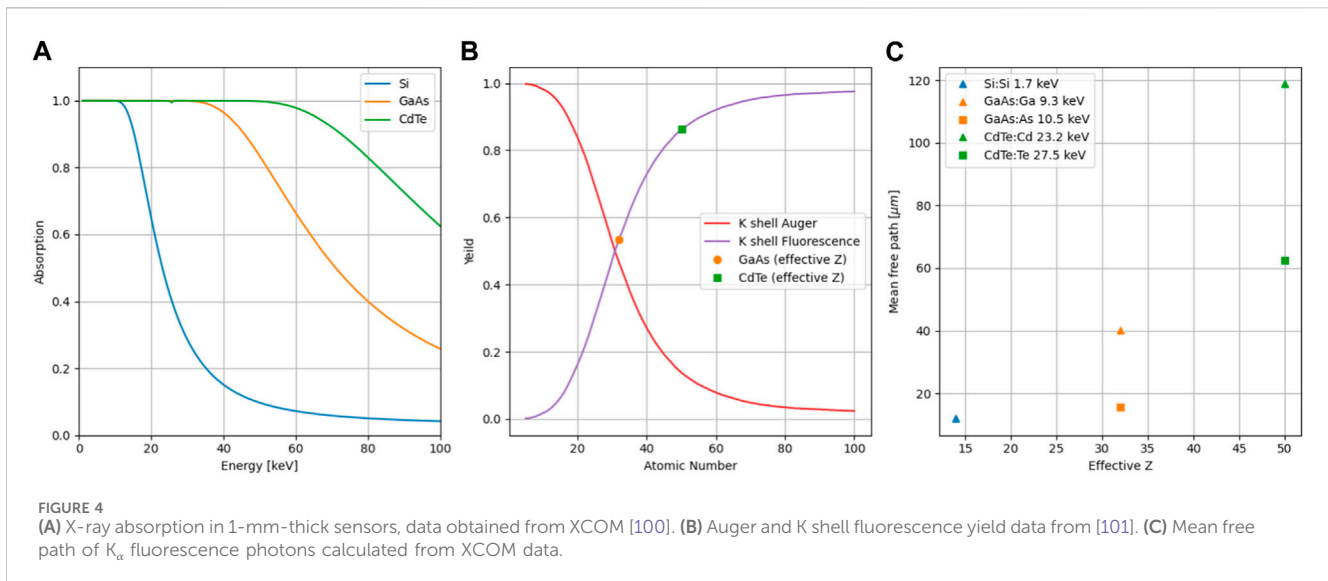
$$\varphi = \phi \sum_{i=0}^N (1 - e^{-\phi\tau})^i \quad (4)$$

where  $N$  is the number of comparators and counters. The improvement in the count rate capability is very pronounced up to three counters. For additional counters, however, there are diminishing returns, and one also has to consider issues like saturation of the preamplifier and available space in the pixel. The plot shown in Figure 2 refers to three counters. With MYTHEN3 we see an improvement of 4-6x in the count rate at 90% efficiency with three counters. For the new MATTERHORN detector, which is currently at the prototyping stage at PSI, we plan to use this approach with four counters.

## 2.3 Spatial resolution

Reducing the pixel size is an effective way to limit the incoming photon rate per pixel and improve the spatial resolution but only up to a certain extent. Single-photon counting detectors with relatively large pixels show an ideal MTF [53], when operated with the threshold at half the photon energy. As the pixel size shrinks, the charge cloud, which—in a typical 300- $\mu\text{m}$ -thick silicon sensor—is on the order of 10–20  $\mu\text{m}$  [54], becomes comparable to the pixel size, leading to worse energy response and a loss of detection efficiency in the corners of the pixel. Figure 3 shows a simulation of the energy response and detection efficiency as a function of the interaction position in the pixel for 25, 50, and 75  $\mu\text{m}^2$  pixels performed using GEANT4 [55] with custom drift-diffusion [56] implementation. As the pixel size becomes smaller, the corner area, where the charge generated by a photon is shared between four pixels, gets larger relative to the rest of the pixel, and both energy response and detection efficiency are degraded. Photons absorbed close to the pixel corner, whose signal does not exceed the 50% energy threshold in any of the four neighboring pixels, will not be detected. For pixels smaller than 30–40  $\mu\text{m}$ , it is not possible to work in the single-photon counting mode without a solution to compensate for charge sharing. Due to the absence of a threshold, charge-integrating detectors do not suffer from the corner effect, giving a flat response throughout the pixel [57].

For high-Z sensor materials (section 2.4.1), reducing the pixel size is even more difficult since the often thicker sensors provide a larger charge cloud, and in addition, X-ray fluorescence in the



material itself further spreads the charge. The fluorescence yield for CdTe is above 80%, and the mean free paths of the characteristic photons are 111 and  $58 \mu\text{m}$  [58]. This leads to distortions in the recorded energy spectrum and additional count rate load since, in many cases, the fluorescence photon is counted separately. Using a charge summing architecture, many of these disadvantages can be overcome [59; 60] but at the cost of the count rate capability [61]. It is possible to handle the corner effect digitally by using multiple comparators and evaluating the coincidence between neighboring pixels. This approach affects the count rate capability similarly to charge summation, but it does not affect the performance of the analog chain.

The best possible spatial resolution of hybrid detectors is reached using a charge-integrating detector under sparse illumination. In this case, charge sharing can be exploited to interpolate between neighboring pixels, reaching a spatial resolution in the micrometer range [62]. For single-photon counting detectors, processing has to happen in the pixel (or strip) since later, no energy information is available. With the MYTHEN3 SPC microstrip detector, it has been demonstrated in 1D that it is possible to use the digital circuitry of a single-photon counting detector with multiple comparators to obtain a spatial resolution better than the physical strip pitch [63]. Similar approaches for pixels are being evaluated, with the additional challenge of a more complex 2D coincidence logic and the need to use a very advanced technology node to shrink the whole circuitry within the necessarily small pixel size [64].

## 2.4 Advanced sensors

The pulse processing front-end in single-photon counting is more forgiving in terms of leakage current than a charge-integrating front-end and allows the use of sensors with a high and/or variable leakage current since the fluctuations are “filtered” out. This is helpful for detection of both low- and high-energy photons since low-gain avalanche diodes (LGADs) and many high-Z sensors exhibit this behavior. However, to allow full flexibility on the

sensor choice, the analog chain of the SPC detector must be designed to work both for hole collection, which is the standard for silicon and inverse LGADs, and electron collection, which is usually required for high-Z sensors and standard LGAD technologies.

### 2.4.1 High-Z materials

Since silicon becomes almost transparent above 20 keV (see Figure 4A), there is a strong need for sensor materials with higher atomic numbers (i.e., high-Z sensor materials) and thus increased photon cross section. Moving away from silicon, we are faced with intrinsic problems like increased X-ray fluorescence yield and range in the sensor material (Figures 4B,C) but also material defects coming from the fact that it is harder to grow high-quality crystals of compound semiconductors which additionally have not benefited from the massive investment from the electronics industry like silicon.

Over time, GaAs, CdTe, and CZT emerged as the most promising materials, and specifically, CdTe is applicable both for medical imaging [65; 66] and at synchrotrons (among others [67; 68; 69]). Larger-area detectors have also started to appear with, for example, the 16-M CdTe EIGER2X from DECTRIS at the P14 beamline at PETRA III. There was some interest in using germanium for SPC detectors [70], but due to the small band gap, they need to be cooled to  $\lesssim -100^{\circ}\text{C}$ , complicating the operation. As sensor materials have improved, the focus has shifted from understanding defects like tellurium inclusions [71] to optimizing for high flux and understanding dynamic effects [72; 73]. [74] and [75] offered a good overview of common sensor materials and their use within the Medipix community. At PSI, we have, for example, studied GaAs with JUNGFRÄU, probing an effective pixel size and understanding the negative signals observed when used with charge-integrating detectors [76].

In terms of new materials, perovskites [77] have attracted a lot of attention, showing a combination of high atomic number and good mobility-lifetime product. The production cost could be orders of magnitude lower than for CZT/CdTe due to cheap base material and a simpler manufacturing procedure. Although some experiments

have been done with photon counting [78], we are still far from observing large-area perovskite detectors at synchrotrons. One potential issue with the current materials is the relatively low mobility [79], which could cause problems in terms of ballistic deficit at the fast shaping times needed for high count rates.

Looking at high-Z sensor material from the perspective of the readout ASIC, we observe a need for electron collection since most high-Z sensor materials have better transport properties for electrons than holes. Leakage current compensation is also important, given the higher leakage currents in most materials, for example,  $> 200$  pA per  $75 \times 75 \mu\text{m}^2$  pixel at room temperature using GaAs, as shown in [80]. The pixel size is also limited by the long range of X-ray fluorescence in the sensor layer, for example,  $110 \mu\text{m}$  in CdTe, requiring relatively large pixels even with interpixel communication for an optimal response. If the pixel size goes much below  $75 \mu\text{m}$ , one has to consider events where the fluorescent photon deposits the energy in a pixel that could be several pixels away from the initial interaction.

## 2.4.2 LGADs

Soft X-ray detection using hybrid detector technology is challenged by the shallow absorption and by the small charge generated by low-energy photons, resulting in low quantum efficiency and low signal-to-noise ratio, respectively. In particular, while charge-integrating detectors can be used for soft X-ray detection even without reaching single-photon resolution, SPC detectors require a SNR  $\geq 10$  for single photons, and they are consequently limited to energies above  $\sim 2$  keV when using standard silicon sensors [28]. The quantum efficiency below 2 keV can be improved by optimizing the entrance window of the silicon sensor, obtaining a performance comparable to state-of-the-art CCDs or CMOS imagers [81].

The signal-to-noise ratio can be improved by exploiting the internal amplification of LGAD sensors recently developed for high-energy physics application. They consist of a silicon sensor with a highly doped p-n junction, where charge carriers are multiplied by impact ionization, thanks to the high electric field. The main goal in particle physics is to use the fast avalanche to improve the timing performance and use the time of arrival of the particle for 4D tracking at high-luminosity colliders [82]. However, the requirements for photon science require additional developments, including small pixels  $\leq 100 \mu\text{m}$  and full sensitivity at the entrance window, in contrast to HEP detectors, with large pads and a thin sensitive layer on a thick substrate. Various LGAD fabrication technologies are described in detail in [83]. A few feasibility studies have been dedicated to X-ray detection [84; 85; 51], aimed at reducing the effective noise, thanks to the multiplication gain, while maximizing the fill factor, which is limited in most LGAD technologies due to the presence of regions without multiplication between the pixels. The performance of LGAD sensors combined with SPC readout electronics does not suffer much from the high leakage current due to the internal amplification, which, however, limits the maximum exposure time acceptable for charge-integrating detectors with LGAD sensors. Recently, the inverse LGAD sensors with optimized entrance window developed by [86] have been combined with the EIGER single-photon counting detector, allowing, for the first time, a SPC pixel detector to reach energies below 1 keV. The low

noise and high dynamic range allowed achieving unprecedented data quality in magnetic contrast soft X-ray ptychography, as described by [87].

LGADs can also be exploited for tender and hard X-ray detection to improve the count rate capability using a faster shaping of the analog signal since gain is provided in the sensor, and therefore, the signal can be sacrificed using fast analog settings and accepting a more ballistic deficit.

## 2.5 Single-photon counting versus charge integrating

With the introduction of charge-integrating detectors with dynamic gain switching (like JUNGFRÄU Mozzanica et al. [88]), it is possible to measure with an electronic noise below the Poisson limit from a single photon throughout the full dynamic range ( $10^4$ – $12$  keV photons/pixel). The maximum 2 kHz frame rate of JUNGFRÄU translates to a count rate capability of approximately 20 Mphotons/pixel/s at 12 keV. Since it is an integrating detector, the maximum supported flux scales inversely with the photon energy, providing even higher numbers at lower energies (e.g.,  $10^5$  photons/pixel/frame and 200 Mphotons/pixel/s at 1.2 keV). Different from SPC detectors, JUNGFRÄU is linear throughout the dynamic range independent of the photon rate (e.g., 40 Mphotons/pixel/s at 6 keV). [89] showed that data quality remained high even at full beamline transmission (thaumatin crystal at 6 keV) in contrast to previously published results with an EIGER 1-M detector [14].

For a charge-integrating detector, the highest continuous flux is determined by the dynamic range multiplied by the frame rate. We are currently developing the second-generation JUNGFRÄU detector with a target frame rate of 10 kHz, which, given the same dynamic range, could cope with 100 Mphotons/pixel/s at 12 keV. This would provide a solution for the most extreme fluxes at the cost of dealing with 12.5 GB/s per 500 kpixel module. However, since the performance of charge-integrating detectors degrades above a few hundred microseconds integration time (higher noise and lower dynamic range), longer exposure times must be achieved by summing up multiple images, with a consequent increase in the electronic noise, which scales with the square root of the number of frames.

At the other extremes, charge-integrating detectors have their place in photon-starved applications since with a low enough flux, it is possible to measure the charge deposited per pixel per photon, enabling both spectroscopic measurements and interpolation (Bergamaschi et al. [90]). However, in this application, the maximum supported flux is limited by the frame rate of the detector, which is typically of the order of a few kHz with one outstanding exception being the HEXITEC<sub>MHz</sub> [91]. For the same applications, the implementation of energy binning or a sub-pixel resolution in a single-photon counting chip could increase the maximum supported flux of up to three orders of magnitude since it would be limited by the  $\sim 100$  ns shaping time rather than by the  $\sim 500 \mu\text{s}$  readout time [63].

An ideal detector should perform like a single-photon counter under low illumination (noiseless, stable, and low frame rate possible) but be capable of supporting high fluxes when

necessary. Charge-integrating detectors with a single-photon resolution converting the analog data to the number of photons on-the-fly and summing up frames to achieve long exposure times promise to achieve this goal, as shown in Leonarski et al. [92], but compared to SPC detectors, they require more developments in terms of chip design, firmware development, and data backend. Moreover, the integration of the leakage current requires the acquisition of frequent dark images to follow possible thermal drifts and the effect of radiation damage, which are usually filtered out in SPC detectors.

### 2.5.1 Mixed mode

An approach for obtaining ideal behavior from a large detector is to equip the more illuminated area of the detector (e.g., the central region for coherent scattering) with charge-integrating detector modules and a high-performance data back-end system and substitute single-photon counting modules in regions with lower illumination, synchronizing the acquisition of all the modules. Some challenges and artefacts might still arise when combining data from such different detector systems.

Charge removal architectures are a hybrid between single-photon counting and charge-integrating detectors [93; 94]. They have a fully digital readout, and ideally, they can provide a high gain (and low noise) on the whole virtually infinite dynamic range. High-flux detection is possible, only limited by the speed of charge removal (e.g., they are not usable at XFELs), while a slow frame rate readout is still feasible, with the disadvantage of large pedestal corrections due to the integration of the leakage current. Charge removal architectures might suffer from high noise levels at high intensities due to charge injection fluctuations during charge removal.

The possibility to statically configure the detector in the counting or integrating mode depending on the application (or on the illumination level) would add great flexibility, and users could benefit from the advantages of two detector systems, without the need to change the detector. This is possible by adding to the feedback circuit of the preamplifier of a SPC detector a reset switch. When the reset is open and the field-effect transistor (FET) controlling the feedback resistor is active, the detector can be operated in the SPC mode, feeding the preamplifier output to an additional shaper and comparator [95]. When the FET of the feedback resistor is off, the reset can be operated to integrate the charge on the feedback capacitor during the given exposure time and then sample on a storage capacitance. The comparator used in the SPC mode could be used in order to implement dynamic gain switching, while the shaper could be operated similar to the preamplifier for correlated double sampling (CDS). Such an architecture has been implemented and successfully tested in a MYTHEN3 microstrip prototype and will be adapted to pixel detectors. Ideally, the detector should dynamically adapt its behavior depending on the illumination, similar to dynamic gain switching, but this comes with additional challenges in the conversion of the signal into number of photons.

## 2.6 More photons, more data

Single-photon counting detectors present some advantages from the data handling point of view compared to charge-integrating

detectors since the detector readout already consists of the number of detected photons, without the need for additional processing. Moreover, the fully digital readout simplifies the data compression (e.g., zero suppression), making it possible already on the readout board or even on-chip.

Still, the higher frame rates driven by the increased brilliance will be a huge challenge for the beam lines. Already the previous generation detectors kicked off the “data deluge” [96], and with new detectors coming out with unprecedented data rates like MATTERHORN with 100 Gbit/s per 500k pixel module or TIMEPIX4 [97] with  $16 \times 10$  Gbit/s serializers per chip (448 Å— 512 pixels) saving (or even receiving), all raw data on the computing infrastructure will not be possible. The LEAPS<sup>1</sup> data strategy [98] outlines the importance of involving central IT services to take some of the burden off experimental groups which might not be used for large datasets, but with the explosion of data, we believe that measures also need to be taken to reduce data at the source.

We see two different approaches, of which one is to build a custom receiving system using FPGAs and GPUs to perform local processing like in [92] before streaming out the data, and the other is to move data reduction onto the readout board or even into the ASIC itself. Reducing the data closer to where it is produced has the added benefit that it lowers the demands on subsequent network and computing infrastructure at the cost of flexibility. For now, it remains an open but highly important field, and we need to see what works best in practice.

## 3 Conclusion

SPC detectors have been the most used position-sensitive detectors at synchrotron facilities for more than a decade, but the higher brilliance of the fourth-generation light sources requires a new generation with improved performance. There are several single-photon counting detectors under development which will be used at diffraction-limited sources including SPHIRD [94], MEDIPIX4 [99], and our own MATTERHORN. All these projects aim to significantly improve their count rate capability compared to existing systems, and in section 2.2, we presented some of the solutions which are being investigated to achieve this goal. Of the compared methods, pileup tracking shows potential to give the highest count rate, but one also has to consider other features of the detector like calibration, correction of mismatches between channels, spectral response, and radiation hardness. Therefore, the best solution will be the one which provides the most accurate data, something that could vary between applications and facilities.

Additional features like the possibility of using multiple comparators and counters and extending the usable energy range from soft X-rays to high-energy photons will open new possibilities for SPC detectors and can eventually pioneer novel experimental

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techniques. Despite not being the optimal solution for targeting high-resolution imaging applications, SPC hybrid detectors with intercommunication between pixels at the analog or at the digital level can further push the spatial resolution or at least improve some of the current flaws of SPC detectors with small pixels, like the corner effect.

Despite the challenge brought by the increased flux, we think that SPC detectors will have a bright future at diffraction-limited light sources and remain the workhorse detector in the near to mid-term future, thanks to their reliability and ease of use. For the highest rate applications, charge-integrating detectors will complement photon counting, until the mixed mode readout combining the advantage of both architecture and other novel readout methods will be available.

## Author contributions

EF: conceptualization and writing—original draft. AB: conceptualization and writing—original draft. BS: writing—review and editing.

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