



Dynamic Performance Evaluation of Grid-Connected Hybrid Renewable Energy-Based Power Generation for Stability and Power Quality Enhancement in Smart Grid

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Amir M, Prajapati AK and Refaat SS (2022) Dynamic Performance Evaluation of Grid-Connected Hybrid Renewable Energy-Based Power Generation for Stability and Power Quality Enhancement in Smart Grid. Front. Energy Res. 10:861282. doi: 10.3389/fenrg.2022.861282 Today's stochastic grid system is experiencing huge voltage fluctuations, which is responsible for power quality issues in the smart microgrid network due to its intermittent nature as well as penetration of hybrid renewable resources. Thus, the dynamic performance evaluation and their control are essential to sustaining the stability of the grid network. A d-q controller mechanism is suggested to maintain the balance of the distributed generation network and grid side network. A dynamic control mechanism of voltage source converter (VSC) is presented in the MPPT-based wind power generating station, where an induction generator generates the power by the optimal control of the wind energy-based subsystem. The distributed hybrid generation (solar PV and wind) subsystem's output terminal is linked to the DC bus bar's common link via the VSC. A VSI is utilized to convert the desired DC power to alternating current power. To regulate and improve the performance of the proposed hybrid power generating systems (HPGS), a supercapacitor (SC) is used to smooth out the ripple on the distribution side in the power grid. Furthermore, the dynamic stability of grid-connected solar PV and wind power generation systems is investigated. This article also proposed an effective control scheme for the SC in HPGS under the influence of weak grid conditions. This article aimed to validate the efficiency of the VSI topology; a PI controller stability enhancement approach is used in a proposed grid system under various disturbance conditions. Finally, the simulation results and FFT-based power quality response analysis are validated through the effective utilization of an SC.

Keywords: dynamic stability, HPGS, induction generator, photovoltaic array, supercapacitor, supply power grid, THD, smart grid

Abbreviations: IG, induction generator; PV, photovoltaic module; D_C , duty cycle (DC-DC boost converter); d, q, values of d and q axis; ρ , differential operator; Z_L , load impedance; V_C , I_C , values of AC-to-DC converter voltage and current; V_i , I_i , values of DC-to-AC inverter voltage and current; PV, values of solar PV module; NL, no-load case; L, values of loaded case; M, measured values; BES, battery energy system; DERs, distributed energy resources; VSC, voltage source converter; FFCL, feed-forward control loop; VSI, voltage source inverter; d-q, direct quadrature; P&O, perturbation and observation; MPPT, maximum power point tracking; HPGS, hybrid power generation system; X_C , commutation reactance; R_L , loading resistance; SC, supercapacitor; LPF, low-pass filter; THD, total harmonic distortion; PCC, point of common coupling.

1 INTRODUCTION

Recently, the use of grid-connected hybrid renewable energy resources (like solar, wind, and hydro) increases rapidly because of the huge expansion in the load demand on the distributed generating system (Kim et al., 2008). But on the other hand, it prompts significant issues and consequences by the intermittent nature of these hybrid energy resources (Salimi et al., 2021; Sun et al., 2021). Despite the availability of energy resources, grid modernization and consumers interest are increasing in the energy market (Amir and Khan, 2021). To meet the desired electricity demand, the most effective solution is to use affordable sustainable energy sources (Armghan et al., 2020; Sanguesa et al., 2021;). According to the energy reports (MNRE, 2019), in India, 56 percent of wind energy and 34 per cent of solar power supplies fulfill the consumer's electricity demand. That was generated by a key factor in promoting sustainable energy sources such as PV and wind-based resources and their interconnection with the on/off the grid (ToghaniHolari et al., 2020). The solar and wind energy hybrid power generation systems (HPGS) were primarily extended because the solar power plant's common accessibility and output generating power are dependent on the following environmental factors: individual irradiance accessibility, ambient temperature, and wind velocity (Okundamiya, 2021). These resources produce a discontinuous and irregular voltage as a consequence of the substantial impact of hybrid RES by the climate and weather variation (Ni et al., 2021; Praveen Kumar et al., 2021). To help with these issues, the hybrid renewable energy network is integrated with the grid network to identify resource reliability. Furthermore, the MPPT control technique is required to maximize the generation and control the most unusual power sources (BhatNempu and Jayalakshmi, 2020).

Various generation control methods have comprised MPPT control ((perturbation and observation (P and O), hillclimbing, etc.), intelligent-based genetic hybrid system, fuzzybased control, artificial neural network (Amir and Zaheeruddin, 2019), and other machine learning approaches to maximize the distributed power from variable renewable resources (Chandrasekaran et al., 2021). On the other hand, various research studies proposed the optimal controller design and optimization approach for the grid-connected their photovoltaic (PV) power generation system. Although the control approach was presented and mainly focused on the hybrid development of the wind-photovoltaic-based standalone system (Soliman et al., 2018), there is a huge challenge in the practical control design for large stand-alone photovoltaic and wind-based hybrid power generation systems. The dynamic stability of a grid-connected distributed generation station is coupled with different stages of the power transformer (Urooj et al., 2021) and converter mechanism, which is having a simultaneous diesel dynamo, and dynamic energy storage to utilize the maximum generation curve for the development of remote emergency backup systems.

The basic control structure of a grid-connected hybrid resource topology in a smart grid is shown in **Figure 1**. To improve the effectiveness of battery systems with HPGS, various research studies were performed (Amir and Srivastava, 2018; Shayeghi et al., 2020). A self-control approach of charge control was built with BSS at different levels. The utilization of reactive power (Q) limiters to increase the dynamic stability of the gridconnected wind-photovoltaic HPGS throughout the DC-link has been demonstrated in the study by Krishan and Suhag (2020).To stabilize the power, super capacitors (SC) were deployed in a hybrid microgrid system (Ponnuru et al., 2021). It should be taken into consideration that nearly all the suggested control



HPGS approaches had been mostly dependent on wind and PV subsystems. This is caused by the availability and potential of hybrid wind and PV to supplement other resources. However, in recent years, wind energy is usually an enormous renewable energy resource that has recently been (Sharma and Suhag, 2018) regularly evolving in the direction of commercialized distributed generation in smart microgrids (Kumar et al., 2021). It is stated that the utilization of ocean energy may come up with an important role in the emerging remote power generation technology in the coming years (Patel et al., 2020). Typically, among all types of the available power station like wind plants and ocean energy, the wind power station is usually having the key thing of many gripping as well as favorable options for power generation credited into its attributes regarding excessive potential stability in smart grid operation (Sanki et al., 2021). However, various control analyses of WPGS methods have been performed and demonstrated in the study by Punna et al. (2021). The research presented in the study by Argyrou et al. (2021) is a feasibility analysis of dynamic control of the photovoltaic-wind hybrid sustainable energy system in islands in Malaysia. In most cases (Wang and Lin, 2007), the investigation and implementation of SC for optimal performance evaluation are examined through the root-loci results and analysis under the inherent value of resources to control generating output. A schematic framework of an HPGS grid-connected system with SC is demonstrated in Figure 1.

Earlier, Al-Dhalaan et al. (1998) research work aimed the utilization of self-commutation-based inverters. The controlling parameters such as V/F are necessary for islanding detection. Some control techniques regarding islanding detection and their control are presented (Pinto and Panda, 2016). However, Ravikumar and Venkatanarayanan (2020) showed a framework of converter control and employed for grid-connected solar PV systems. This research investigates a power regulatory control in hybrid (solar and wind) distributed power systems based on the MPPT control. To ensure the effective utilization of hybrid resources, the Fuzzy-based FIS method is employed to achieve the MPPT for the hybrid (solar PV module and WT) subsystem. In this article, we will examine the transient stability and Q compensation control in the HPGS using a supercapacitor. It provides efficient utilization of energy resources and decreases consumption of battery storage, thus enhancing battery health (Faessler, 2021). The proposed system is investigated with the MATLAB Simulink tool.

This article offers the dynamic stability enhancement strategy and the performance steady-state characteristics of the HPGS, which is delivering power to the demand-side loads, for the improvement in the HPGS variables under grid (Amir and Srivastava, 2019). Also, a voltage stabilization scheme is presented for the supercapacitor (SC) in the HPGS under the influence of weak grid conditions. HES, based on the PV system connected to the weak grid, was considered for the evaluation of the system. The novel control scheme was designed to enhance quality power issues, also caused the reduction of ripples in the content through the VSI system with the DC link, and also improved the stability range of the HPGS by retaining the performance of the SC. Also, the proposed HES was designed and simulated in the MATLAB-R2021b tool, and the enhancement in the SC and PI controller scheme over variation in different parameters like improvisation using d-q control in DC link voltage was verified. Respectively, a significant drop in the ripple contents was present in VSI normal operation, and finally, the better dynamic performance of HPGS was integrated to the weak grid.

The remaining part of the article proceeds as follows: In **Section 2**, the design and mathematical modeling are presented. The control mechanism of the HPGS and converter topology is discussed in **Section 3**. The results of the suggested SC and controller topology for the proposed HPGS model are presented in **Section 4**. At last, the key conclusions and future framework of the HPGS are shown in the last section.

2 DESIGN OF THE GRID-CONNECTED HPGS MODEL

The wind energy subsystem and PV subsystems are coupled to the grid of the distributed network, as referred from Figure 1. The step-up converter connected to the PV module consists of an MPPT tracking algorithm. Although the power delivered to the grid via a hybrid system is fluctuating in nature, a supercapacitor (SC) is used to minimize the ripples and harmonics (Dai et al., 2016). The material used for constructing the supercapacitor has the characteristics of higher quick repose time, high power density, and does not require any cooling system (Urooj et al., 2020; Adil et al., 2022). The supercapacitor also maintained a balanced active power operation of the converter topology so that the interconnected converter and inverter can perform smoothly. The control mechanism for the VSI topology is presented in equations ((Eqs 1, 3, 5) for "d" parameter and (Eqs 2, 4, 6) for "q" parameter). The VSI control is shown in Figure 1, in which computation and control of grid power through LC filter (Naidu et al., 2019), a DC link network line, and a multilevel converter topology.

$$\mathbf{L}_{I} \times \mathbf{p} \times \mathbf{i}_{dI} = -\mathbf{R}_{I} \cdot \mathbf{i}_{dI} + \boldsymbol{\omega}_{e} \cdot \mathbf{L}_{I} \cdot \mathbf{i}_{qI} + \boldsymbol{v}_{dI} - \boldsymbol{v}_{dPCC}, \qquad (1)$$

$$L_I \times p \times i_{qI} = -R_I \cdot i_{qI} - \omega_e \cdot L_I \cdot i_{dI} + v_{qI} - v_{qPCC}, \qquad (2)$$

$$C_I \times p \times v_{dPCC} = i_{dI} - i_{dTL} + \omega_e \times C_I \times v_{qPCC}, \qquad (3)$$

$$C_I \times \mathbf{p} \times \mathbf{v}_{qPCC} = \mathbf{i}_{qI} - \mathbf{i}_{qTL} - \mathbf{\omega}_e \times C_I \times \mathbf{v}_{dPCC}, \qquad (4)$$

$$L_{TL} \times \mathbf{p} \times \mathbf{i}_{dTL} = -\mathbf{R}_{TL} \times \mathbf{i}_{dTL} + \boldsymbol{\omega}_{e} \cdot L_{TL} \cdot \mathbf{i}_{qTL} + \boldsymbol{v}_{dPCC} - \boldsymbol{v}_{dinf}, \quad (5)$$

$$(L_{TL})p(i_{qTL}) = -R_{TL} \times i_{qTL} - \omega_e L_{TL} \cdot i_{dTL} + v_{qPCC} - v_{qinf}.$$
 (6)

The dynamic control units of all these subsystems can be expressed from Eqs 1–6, where $(v_{dI} \text{ and } v_{dl})$ and $(i_{dI} \text{ and } i_{dq})$ are the voltage and currents of the VSI, respectively. Also, the point of common coupling (PCC) voltages $(v_{dPCC} \text{ together})$ with v_{qPCC} and voltage $(v_{dinf} \text{ together with } v_{qinf})$ were measured at the same level of the PCC. Current i_{dTL} and i_{qTL} are usually showing the interconnection range through the transmission line. The conventional full-bridge-based power electronic converter topology has been commonly implemented for the 3Ø HPGS because of the simplified circuit structure, lower cost, better efficiency, lower leakage



current, and so on (Estévez-Bén et al., 2020). For this research work, a two-level 3Ø-VSI is employed to control active (*P*) and reactive power (*Q*) in the grid integrated system. Which is divided into three models and controlled has 2 levels, a DC-to-DC boost stage conversion and further a DC-AC inversion level. The DC side subsystem model consists of a PV module-based DER, which is dependent on solar irradiance variation. The DC-DC converter is a necessary part to the interface to convert the lower V_{DC} output from the PV subsystem by the utility grid system *via* appropriate consumption voltage. The converter model consists of a converter and filter together with the grid control subsystem by controlling *V*, *F*, and synchronization of the distributed grid. The proposed model of the layout framework comprises the distributed grid and different loads.

2.1 Design of a Solar PV Subsystem With a Converter Circuit

Generally, the PV module shows a non-linear power characteristic, which varies as solar irradiation (in kW/m²) and ambient temperature vary. The model's classes of grid systems with PV cells have been described in several pieces of the literature (Estévez-Bén et al., 2020). As a distributed intermittent power resource, a solar PV module is reflected as the significant importance in the HPGS. The operational framework of a solar PV cell formulates the key characteristic in building a solar PV module by **Eq.** 7 (Patel, 2006). The photovoltaic module is made up of various PV cells interconnected in sequences of the array. The PV module is developed by a series of PV cells that are associated in a particular order. The schematic circuit of the stand-alone PV cell with the diode is shown in **Figure 2A**.

This circuit consists of a PV cell that is composed of the power supply current (I_{ph}) , a diode (D_j) , a parallel resistance (R_{par}) , and q is the electron charge; and a series resistance (R_{se}) increasing the unit of a PV cell to maximize the PV generated power range is shown within (Nwaigwe et al., 2019), and the

resulted current (in ampere) of the analyzed PV array range is indicated by Eq. 7.

$$I_{PV} = N_{ij}I_{ph} - N_{jk}I_{NL} \left\{ exp \left[\frac{q(V_{PV} + R_{P1}I_{PV})}{kATN_{se}N_{ij}} \right] - 1 \right\} - \left(\frac{(V_{PV} + R_{P2}I_{PV})}{R_{P2}} \right).$$
(7)

Here, V_{PV} is the voltage of a PV module, PV range, N_{ij} , and N_{jk} are typically the required PV units attached in parallel as distributed generation, individually (Zolfaghari et al., 2018), where N_{se} is the quantity of the sequence-linked cells in a PV module; the parallel resistances are R_{p1} and R_{p2} correspondingly, and I_{Ph} and I_{NL} are a particular phase current (in amp) and no-load current (in amp), respectively, in **Eqs 8**, **9**.

$$I_{PH} = \left[I_{T,n} + k_i \left(T_p - T_{pn}\right)\right] \cdot \left(\frac{G}{G_n}\right),\tag{8}$$

$$I_{NL} = I_{T,n} \cdot \left(\frac{T_p}{T_{pn}}\right)^3.$$
(9)

Additionally, generated current $(I_{0,k})$ is shown in Eq. 10.

$$I_{0,k} = \frac{I_{sc,k}}{\left\{ exp\left[\left(qV_{oc,k} \right) / \left(kAT_k N_{ij} \right) \right] - 1 \right\}},$$
(10)

where T is the cell temperature (in ${}^{o}K$) and G is the inversion (in K) and photovoltaic irradiance (in kW/m^2). Individually, for instance, the electron energy is related within semiconductors (e_v) . Also, A is the charge of the electron (in C), the Boltzmann constant (within J/K), in addition to the diode operation. **Figure 2C** displays the simplified analysis through the DC/DC converter, which is coupled with the PV module to the DC link connection. The switching behavior of a diode (D) as a switch (S) is demonstrated in **Figure 2B**. The chopper DC/DC converter is displayed in **Figure 2C**. Hence, the dynamic calculation is utilized to simulate the DC/DC converter.



$$F_c \times P_w \times V_{PV} = i_{PV} - i_{LP}, \qquad (11)$$

$$(F_L).(P_w).(i_{LP}) = -R_P i_{LP} + V_{PV} - (1 - K_P).V_{DC}, \qquad (12)$$

$$\mathbf{i}_{PV-DC} = (1 - K_p) \cdot \mathbf{i}_{LP}. \tag{13}$$

Here F_c is typically the filter capacitance, i_{PV-DC} is the current of dc-link through the PV module, L_C and R_C are the certain values of converter inductance and resistance, respectively. In addition to this, K_P will be the liable ratio from the current DC/DC converter. Also, i_{LP} is the inductor current and V_{DC} is the dc-link voltage.

The power and current characteristics with respect to the voltage of a PV module are demonstrated in Figure 3. The proposed solar PV module is simulated for average insolation (800 W/m^2) with a temperature of 27 °C.

2.2 Mathematical Modeling of Solar PV Inverters

The main objective is to feed extracted HPGS power via the boost conversion stage into the distributed grid by distributed grid-side converters. This is accomplished by the control of DC-link voltage as mentioned in the study by Hoon et al. (2016). In this research work, a VSI was linked to the distributed grid through an L-type filter for better response. Similarly, boost converters or the PV inverters have been controlled by the controlling of the duty cycles, and hence, output voltage was regulated. Various techniques based on PWM can be employed to generate the duty cycle for the inverters. A bipolar modulation has been preferred due to its low leakage current. This is important for transformer-less PV systems. There are two switching states in bipolar modulation: 1). SW1, SW4 are on, and SW2, SW3 are off, and 2). SW1, SW4 are off, and SW2, SW3 is on. Required expressions are given by Eq. 8.

$$\begin{cases}
L.\frac{di_{gr}}{dt} = v_{DC} - R \times i_{gr} - v_{gr}, & 0 \le t < d.T_{sw} \\
L.\frac{di_{gr}}{dt} = -v_{DC} - R \times i_{gr} - v_{gr}, & d.T_{sw} \le t < T_{sw}
\end{cases}$$
(14)

where v_{DC} is an input DC voltage from the VSI by boost stage with voltage (v_{qr}) and current (i_{qr}) parameters of the HPGS grid. For the design of a controller associated with the PV inverter, the transfer function in the frequency domain is to be achieved. So forth, a small-signal model was introduced to make the dynamic equations linear. Eq. 9 has been used to calculate the average value of the inductor voltage. Finally, using Eq. 10, the transfer function for the system can be obtained and thus used to design the controller for the associated PV inverters.

$$L\frac{d(i_{gr})}{dt} = (2d-1) \times (v_{DC}) - R.(i_{gr}) - v_{gr}, \qquad (15)$$

$$\hat{i}_{gr}(s) = \frac{2D - 1}{Ls + R} \cdot \hat{v}_{DC}(s) + \frac{2V_{DC}}{Ls + R} \cdot \hat{d}(s) - \frac{1}{Ls + R} \times \hat{v}_{gr}(s).$$
(16)

High-order filters have been used for the suppression of desired harmonics; it is a combination of L and C together. Therefore, higher frequency harmonics have been suppressed by the LCL filter. A damping-based resistor is generally employed in the design of an L-C-L filter to rectify the system resonance issues that may be responsible for instability in a hybrid grid system (Teodorescu, 2011).

The FFT-based THD analysis of voltage characteristics is shown in **Figure 4**. Initially, when the load is not increasing, the THD in the distributed voltage is found at 14.67% and the waveform has a nominal value of RMS voltage.

2.3 WPGS for VSC Designed

The design of the wind power generation control system is based on a mass-spring-damper control system (Kumar et al., 2021) that can be modeled as

$$\boldsymbol{p}(\boldsymbol{z}) = \boldsymbol{U}_{\boldsymbol{z}},\tag{17}$$

$$\boldsymbol{m}_t \times \boldsymbol{p} \times \boldsymbol{u}_z = \boldsymbol{F}_w - \boldsymbol{K}_s - \boldsymbol{K}_D \boldsymbol{.} \boldsymbol{u}_z - \boldsymbol{P}_f, \qquad (18)$$

where the gear operator is symbolized by p, the floater wind speed is denoted by z, U_z are the distances, m_t is the overall size of the specific floater, F_w is the certain force coming from the wind, and P_f is the pressure exerted on the floater. In **Eq. 18**, K_D along with K_s are the dissipating ratio and spring constants, respectively. The LPMG's representation based upon the d-q axis mechanism of WPG is expressed as (Penalba et al., 2017)

$$L_{dLG} \times p \times i_{dLG} = \omega_z L_{qLG} \cdot i_{qLG} - R_{LG} \cdot i_{dLG} - \nu_{dLG}, \qquad (19)$$





$$L_{qLG} \times \mathbf{p} \times \mathbf{i}_{qLG} = \omega_z \cdot \psi_{PM} - \omega_z \cdot L_{dLG} \cdot \mathbf{i}_{dLG} - \mathbf{R}_{LG} \cdot \mathbf{i}_{qLG} - \mathbf{v}_{qLG}.$$
 (20)

In **Eq. 19**, i_{dLG} , i_{qLG} , v_{dLG} , and v_{qLG} are both the d and q-axis voltages and current parameters, respectively. Also, R_{LG} is the resistance of stator-winding. In **Eq. 20**, L_{dLGG} and L_{qLG} include the *d*-*q*-axis inductance parameters individually.

$$\mathbf{i}_{qLG_ref} = \frac{(2\tau K_D U_z)}{(3\pi \psi_{PM})}.$$
(21)

The control mechanism of WPGS for VSC is designed based on *d* and *q* axis parameters related to the particular translation for the L_{PMG}, which is modified in the control operator associated with the VSC. The operation of the WPGS control operator is to get optimum energy coming from the wind energy resources and enhance the energy in the L_{PMG}. In research (Kulkarni et al., 2016), we can evaluate the energy loss in the particular L_{PMG} and the effective energy transformed from the wind resources to manage and simply by d and q axis parameters of the L_{PMG} (*i*_{dLG} and *i*_{qLG}), individually. As referred from **Figure 5**, the proposed control structure associated with the VSC to determine the d-q axis parameters of L_{PMG} as per reference indexes *i*_{dLG_ref} and *i*_{qLG_ref}, and optimal use of PI controller.

2.4 Implementation of DC-DC/DC-AC Topologies

Solar and wind are both energy resources that are intermittent in nature. Thus, stability can be improved by the hybrid resources of these two systems to comply with the remote areas. These systems can be further employed for the application of storage systems through the admittable converter systems, that is, AC-DC converters, then DC-DC converters, and again DC-AC converters. The load demands can be consummated by these renewable sources combined with storage systems and various converter/inverter systems. This hybrid system has shown diverse dominance to supply the energy demand by converter topology. **Eqs 21, 22** used for the DC-to-DC converter for boost operation are expressed as follows:

$$\begin{bmatrix} pi_E \\ pv_c \end{bmatrix} = \begin{bmatrix} -\frac{R_E + (1-D)(R_C//R)}{L_E} & -\frac{(1-D)R}{L_E(R+R_C)} \\ \frac{(1-D)R}{(R+R_C)C} & \frac{-1}{(R+R_C)C} \end{bmatrix} \cdot \begin{bmatrix} i_E \\ v_C \end{bmatrix} + \begin{bmatrix} \frac{1}{L_E} \\ 0 \end{bmatrix} v_g,$$
(22)

$$\mathbf{v}_E = \left[(1 - \mathbf{D}) \left(\mathbf{R}_C / / \mathbf{R} \right) \quad \frac{\mathbf{R}}{(\mathbf{R} + \mathbf{R}_C)} \right] \cdot \left[\begin{array}{c} \mathbf{i}_E \\ \mathbf{v}_C \end{array} \right].$$
(23)

Eqs 21, 22 are representing the converter state vector model and its output, i_E is the current through inductor L_E , v_c is the voltage within capacitor *C*, v_E is the voltage across the output terminals of the converter, and *D* is the duty ratio of the switching device to transfigure input voltage to other higher-level voltage. DC-AC inverter system linearization modeling is shown by the following equations.

The output voltage on a per-unit basis across the VSI terminals is given as follows, in **Eq. 23**:

$$V_{I} = V_{qI} \times \cos \gamma_{I} - \left(\frac{\pi}{6}\right) \times X_{CI} \times I_{I}.$$
 (24)

In Eq. 23, V_I and I_I are representing inverter input side voltage and current, respectively, γ_I is the excitation angle, X_{CI} depicts the commutation choke reactance of VSI, and V_{qI} is a voltage of the inverter output side. The system linearization equations for the HPGS along with converters and inverter topology can be simplified in the most appropriate formats as follows:

$$p[X] = f([X], [U], t),$$
(25)

where [X] is representing the state vector, [U] is the control input vector, and f is a non-linear vector of the concerned hybrid system. Eq. 24 can be written in the linearized format as follows:

$$\boldsymbol{p}[\Delta \boldsymbol{X}] = [\boldsymbol{A}][\Delta \boldsymbol{X}] + [\boldsymbol{B}][\Delta \boldsymbol{U}].$$
(26)

Here, [A] and [B] represent the HPGS and controller matrix, respectively, and $det([A] - \gamma[I])$ is the characteristic system linearization equation expressed as follows in **Eq. 27**.

$$det([A] - \gamma[I]) = 0, \qquad (27)$$

where [I] is the identity matrix of a suitable rank, and γ is the eigenvalue of the proposed system matrix [A]. When any characteristic root is being on the right side of the principal complex Laplacian plane, then the HPGS may be subjected to disturbance, and it may lead toward its overall system instability.

2.4.1 Mathematical Modeling of a DC-DC Boost Converter

DC-to-DC converters bring forth the loads and PV modules. DCto-DC boost converters have inherent properties like higher power density and faster response during transient periods. The primary components of the DC-to-DC converters are inductance (*L*), a diode (*D*), filter capacitance (*C*), and a high-power semiconductor (MOSFET/IGBT) switch. Usually, they are used in PV inverter systems to find their application for the control of power extraction in PV systems with the appropriate implementation of MPPT algorithms (Elbarbary and Alranini, 2021). The purpose of MPPT



in the PV-based subsystem is to continuously extract maximum power from solar PV modules irrespective of the load condition or weather. In general, an efficient MPPT algorithm can be defined based on three main aspects: fast dynamic response, high accuracy under the steady-state condition, and robustness to disturbances. Considering these aspects, the Perturb and Observe (P&O) algorithm is widely employed for maximum power (Kollimalla and Mishra, 2013). To vary the duty cycle (D_s) of a boost converter to control the transferred power, an MPPT technique is required. The dynamic equations for the calculation of L and C are given by **Eqs 28, 29**.

$$L.\frac{di_L}{dt} = v_{sPV} - (1 - D_s) \times v_{DC}, \qquad (28)$$

$$C_{pv} \cdot \frac{d(v_{spv})}{dt} = (i_{s,c}) - \frac{v_{spv}}{R_{mpp}} - (i_L), \qquad (29)$$

where *L* is the inductance, i_L is the current through the inductor, *C* is the capacitance, and R_{mpp} corresponds to the resistance of the MPP. Another important variable is an output

current (i_{sPV}) through the solar PV system, which is derived as given by Eq. 30.

$$\mathbf{i}_{sPV} = \mathbf{i}_{S.C} - \frac{\mathbf{v}_{spv}}{\mathbf{R}_{MP}}.$$
(30)

Here, i_{sPV} is the current through the solar PV subsystem, i_{SC} is the respective short-circuit current, and v_{spv} is the voltage across the solar PV module.

2.5 Supercapacitor and DC-DC Converter

The equivalent circuit of the supercapacitor (SC) is shown in **Figure 6A**. In addition to the capacitance (C_{SC}) of the SC, the optimal placement of SC represents as well as holds into consideration reactive power compensation using SC. From **Figure 6A**, the dynamic computation of the SC is employed and represented by **Eq. 31**.

$$SC_c \times P_w = SC_{SCPSC_{SCVC}},$$
 (31)

$$V_{SC} = V_{qsc_{sc\,csc}},\tag{32}$$

where SC is the voltage (V_{SC}) and current (i_{SC}) , respectively, and V_{CSC} is the voltage across C_{SC} .

The simplified layout of the control schemes is presented in **Figure 6B**. The DC-DC converter comprises two switches (SW₁ and SW₂), which are operated in a switch operation. The converter is enabled by the boost setting of the circulation current. Switch SW₂ works as a switch, whereas SW₁ works as a diode, and the energy is fed by the SC to the DC link. Furthermore, once operating the DC-DC converter in the supercharged condition, the transitioning approach is usually overlooked. **Figures 6C,D** describe the active mean value of bidirectional DC-DC topology, which is step-down and setting of desired voltage. The mathematical expression of the converter mode topology is given in the following text.

a) In the boost mode:

$$L_s \times \boldsymbol{p} \times \boldsymbol{i}_{sc} = -\boldsymbol{R}_s \cdot \boldsymbol{i}_{sc} + \boldsymbol{V}_{sc} - (1 - \boldsymbol{D}_s) \times \boldsymbol{V}_{DC}, \quad (33)$$

$$\mathbf{i}_{SC-DC} = (1 - \mathbf{D}_s) \times \mathbf{i}_{sc}. \tag{34}$$

b) In the buck mode:

$$L_s \times \boldsymbol{p} \times \boldsymbol{i}_{sc} = -\boldsymbol{R}_s \cdot \boldsymbol{i}_{sc} + \boldsymbol{V}_{sc} - \boldsymbol{D}_s \times \boldsymbol{V}_{DC}, \qquad (35)$$

$$\mathbf{i}_{SC-DC} = \mathbf{D}_s \times \mathbf{i}_{sc},\tag{36}$$

where L_s is the inductance of the energy storage inductor and R_s is the resistance individually, i_{sc_DC} is the current feedthrough SC in the DC link.

The control configuration of a typical bilateral DC-to-DC converter *via* DC link and the SC is capable of maintaining the active power flow, as presented in **Figure 6E**. Generally, a bidirectional DC-to-DC converter-based controller is employed, which is dependent on current control loops. The external current control loop measures the V_{DC} through DC link as per V_{DC_ref} and the internal control loop for i_{SC} control *via* SC. Once the DC voltage reaches the typical desired range, then SC is followed by i_{SC_ref} , which is further controlled by the external loop *via* DC connection voltage.



3 CONTROL MECHANISM AND PERFORMANCE EVALUATION BY THE PROPOSED CONTROLLER

This section depicts the proposed control approach for the HPGS. The main goal of controlling the HPGS is to decrease harmonics and ripples in the grid, ensuring that the network's stability is maintained. The solar PV modules and wind power generating system reached to the distributed load *via* the DC-to-DC step-up-based converter and a controlled VSC. Although the bidirectional DC-to-DC converter maintains the DC side voltage, the controlled VSI is responsible for supply grid power management. The detailed study of the control mechanism is discussed for different aspects.

3.1 Control Mechanism of Step-Up Converter Integrated With the Solar PV Subsystem

The power generated by the solar PV module is fluctuating and changes with the solar irradiance and temperature. To operate the HPGS efficiently, the maximum power delivered by the PV module needs to monitor constantly. To achieve this objective, the step-up DC/DC converter associated with the PV module is regulated to control the V_{DC} of the module to get its desired maximum value. Various algorithms of MPPT are proposed in several stability studies (Hlaili and Mechergui, 2016). Furthermore, this article used the extensive MPPT control technique; that is, P&O is applied for controlling the operation of step-up DC/DC converter.

3.2 Control Mechanism of WPGS for VSC Designed

The controller layout of voltage d-q axis parameters is represented in **Figure 7A**. The controller is associated with the VSI, which is utilized to manage the distributed energy transmitted *via* the VSI, and also to stabilize the voltage value of the PCC at the grid side. To get the optimal control associated with the active grid elements, the active power of the VSI is carried out within the d-q axis lined up at the PCC. As a result, the energy transmitted using the voltage source inverter associated with the PCC is managed by the i_d and i_q d-q axis for VSI correspondingly. The controller associated with the VSI offers the rapid control system, as demonstrated in **Figure 7A**, where the external control loops deal with the effective energy flow from VSI at a different voltage.

The active power is controlled through the VSI *via* reference PCC of the HPGS (PV and wind power generating systems). Thus, the control mechanism allows the VSI to smoothly transmit the HPGS-based electricity to the grid. It can be achieved by applying the external control scheme as presented in **Figure 7B**. The developed HPGS power is transferred through the low-pass filter (LPF).

The time constant of low-pass filters is symbolized by T_1 and T_2 , respectively, as shown in **Figure 7B**. To minimize the power variation caused by wind, speed fluctuation (T_1) should be chosen equivalent to the wind duration (T_w). However, another constant can be chosen, that is, T_1 to control the power developed by P_{LG} and the PV module.

3.3 Control Through P&O-Based MPPT Controller

The voltage controller associated with a grid-connected inverter has been designed in three parts: voltage regulator for V_{PV} control, current controller for i_{PV} control, and PI controller for V_{dc} synchronization. Grid synchronizer has been designed to generate reference grid current so that *P* and *Q* can be under a specified limit (Zhang et al., 2010).

It is observed from **Figure 8** that $v_{g\beta}$ = grid voltage (v_g) , and v_g is in a similar quadrant as virtual voltage v_{ga} ; therefore,

$$\boldsymbol{v}_{mg} = \sqrt{\boldsymbol{v}_{g\beta}^2 + \boldsymbol{v}_{g\alpha}^2},\tag{37}$$

where magnitude v_m is denoted by $v_{m.g.}$. A controller for current control, proportional-resonant (PR) has been proposed in the study by Ngo and Santoso (2016). This PI controller is accomplished of desired reference signals, and it does not require an error value (Ozdemir, 2016). This task is accompanied only due to the reason that the PR controller has



FIGURE 8 | Inverter control block with PI and current controller.



a higher gain value even at the resonant frequency by utilizing the advantage of delay caused by the PWM generator (Kazimierczuk, 2000). The filter transfer function for the grid is obtained with the difference between the VSI output *V* and *I*. The current controller expression can be represented as

$$FC_{cc}(s) = FC_{PR}(s) + FC_{RC}(s)$$

= $k_{pr} + \frac{k_{ir}}{s^2 + w_0^2} + \frac{k_{rc}Q(s)e^{-sT_0}}{1 - Q(s)e^{-sT_0}} \cdot G_f(s).$ (38)

In **Eq. 38**, PR-based controller is signified by $FC_{PR}(s)$ for fundamental grid frequency controller, where k_{pr} is proportional control gain and k_{ir} is the resonant control gain to develop PRbased controller. F $FC_{PR}(s)$ is represented through values of w_0 and RC controller (k_{rc}) , and the fundamental time (T_0) is $\frac{2\pi}{\omega_0}$. Q(s) is an LPF and $G_f(s)$ is a phase-lead compensator which is equal to e^{sTc} with compensation time (T_c) (Jadoon et al., 2017).

3.4 DC Link Voltage Balancing Controller

In a grid-connected VSI, two control loops (external and internal) are typically employed (Hossain et al., 2017). The external control loop is employed to power control and control power fed into the distribution power grid through the DC link. However, the inner control loop is employed for current control and manages the power quality problems in faster response (Hossain et al., 2017). There are different types of proportional control schemes for the outer control power loop which depends on grid operation. But the most common control scheme is PI through the DC link voltage controller presented in **Figure 9**.

For the control of DC-link voltage, a closed-loop controller is needed to adjust within a specified voltage limit. The reference parameters of DC link voltages are adjusted by a voltage controller-based external loop, which controls the injected distributed grid current values. DC voltage was fed through a proposed PV module to regulate a reference current (i_{ref}) for injected grid current (i_{grid}). The DC voltage control is usually controlled using a PI-based controller. A PI controller is employed for DC link voltage balance. This PI controller representation is shown in **Figure 9**.

3.4.1 Converter Performance Evaluation Under Proposed PI Control Scheme

To decrease the error and enhance the dynamic stability of VSI (Dong et al., 2019), the operational capacity of the PI controller utilizes P control flow in VSI.

Figure 10A displays the duty ratio (in %) with respect to output voltage for the testing of VSI. From this response, it is observed that because the load variation scenario at t = 0.2 s characteristic is fluctuating, it further achieves a new stable state after t = 0.35 s.

3.5 Operational Control of SC for Stability Enhancement Under Weak Grid Condition

In Figure 11, the inverter configuration has been shown, which is connected through an inductor-capacitor-inductor (LCL) filter scheme. Here DC side of the inverter is interconnected with a supercapacitor (SC). A buck-boost (DC-to-DC) converter has taken advantage of connecting SC, which employs an inductor L_{sc} and bidirectional switches SW_3/SW_4 . In this configuration, i_{sc} is the current through the SC, d_{sc} duty cycle, and VSC is the voltage across the SC terminals. The designed LCL filter embodies two inductors L_1 and L_2 , and one capacitor C_f , as exhibited in Figure 11. At PCC, the grid has been connected to the hybrid system, where U_{Sabc} is the rated grid voltage, Z_a is the impedance of the grid, i_g is a current injected into the grid, i_C is the capacitor current, i_{inv} is the inverter current, and finally U_{aabc} is the voltage of the distributed grid at PCC, where *abc* is the reference frame (stationary) that was converted into a synchronous-based reference frame (d-q) by taking advantage of Park's transformation. The power developed by the solar subsystem is expressed as P^* , P_g is the injected power in the grid network, while K_P is the droop gain for developing the controller loop, w is the system frequency, w^* is the designed system frequency for the hybrid system, and finally θ corresponds to the synchronizing angle.

3.5.1 Control Scheme for the HPGS Integrated to a Weak Grid

Figure 12 exhibits the control structure for the control of DC link voltage and reactive power connected to the LCL filter, where U_{dc}^* provides a reference for the DC link voltage, i_{gd}^* is the based reference grid current (direct-axis), and i_{gq}^* is the based reference grid current (quadrature-axis). These values of references have been chosen to be zero to regulate the reactive power so that the power factor may lead toward a unity value.

 U_d^* and U_q^* are expressed to exhibit output for the grid current (i_g) controller loop and also employed as references to generate the switching signals for the inverters. Finally, w represents the frequency of the grid voltage/current. **Eq. 1** of desired voltages can be modeled as follows:

$$U_d = L \frac{di_{gd}}{dt} + U_{gd} - wLi_{gq}, \qquad (39)$$

$$U_q = L \frac{di_{gq}}{dt} + U_{gq} + \Delta i_{gd}, \qquad (40)$$

where $L = L_1 + L_2$ and reference voltages have been obtained using feed-forward control circuit as follows:

$$U_d^* = \Delta U_d + U_{gd} - \Delta i_{gq}, \qquad (41)$$

$$\boldsymbol{U}_{\boldsymbol{q}}^{*} = \Delta \boldsymbol{U}_{\boldsymbol{q}} + \boldsymbol{U}_{\boldsymbol{g}\boldsymbol{q}} + \Delta \boldsymbol{i}_{\boldsymbol{g}\boldsymbol{q}}. \tag{42}$$



FIGURE 10 | (A) Duty ratio and (B) VSI output.





The control scheme discussed previously is not appropriate for the poor grid connection. The following reasons are listed below in three steps.

$$\mathbf{i}_{gd}^* = \Delta \mathbf{i}_{gd} + \mathbf{i}_{gd}, \tag{43}$$

$$\mathbf{i}_{gq}^* = \Delta \mathbf{i}_{gq} + \mathbf{i}_{gq}, \tag{44}$$

a) Change in input current/power directly reflected into the reference currents is given in **Eqs 41**, **42**.

where Δi_{gd} and Δi_{gq} represent the DC values for the changes that support directly to the respective grid current and thus provides reference value under the required transient nature. But, at the



same time, these variations do not appear in the feed-forward controller loop and hence dynamic response is slower for the i_g control loop. The voltage instability is another main issue in the poor grid condition where impedance (z) of a proposed grid is not constant with respect to the operating system. The outer control loop is confined for the voltage control and the inner loop for grid current control.

b) The voltage control loop has a smaller bandwidth than the grid current control. So, in a poor grid-connected system where grid impedance is considered, the feed-forwarddecoupled control scheme for $-wLi_{gq} - wL_gi_{gq}$ and $wLi_{gd} + wL_gi_{gd}$. Therefore, **Eqs 41**, **42** are simplified as follows:

$$U_d^* = \Delta U_d + U_{gd} - wLi_{gq} - wL_g i_{gq}, \qquad (45)$$

$$\boldsymbol{U}_{q}^{*} = \Delta \boldsymbol{U}_{q} + \boldsymbol{U}_{gq} + \boldsymbol{w} \boldsymbol{L} \boldsymbol{i}_{gd} + \boldsymbol{w} \boldsymbol{L}_{g} \boldsymbol{i}_{gd}.$$
(46)

But the main issue is here that L_g is unknown values of grid inductance that cannot be rectified using FFCL, and hence, current control stability margin is reduced. So, for large grid inductance, the current control loop is severely affected by L_g . On the other hand, the outer voltage controller loop stability is also reduced because the PI controller used for it is tuned based on the inner current control loop and leads toward the voltage instability. Due to the occurrence of non-linear loads, steady-state lower order harmonics are also available in grid current. These harmonics cannot be suppressed by the conventional control system. So, i_g in the steady state can be represented as follows:

$$osc \ \mathbf{i}_{gd} = \ \mathbf{i}_{gd}^* + \ \mathbf{i}_{gd}, \tag{47}$$



rating SC to the effectively stored energy into the distributed grid. (B) Solar PV module via DC-to-AC inverter during no-load scenario. (C) V characteristics of wind energy subsystem. (D) Measured three phase V and I waveforms under no load. (E) Current stability response of HPGS under RL load. (F) Current stability response of the HPGS with SC operation under RL load.

$$i_{gq} = i_{gq}^* + \hat{i}_{gq}.$$
 (48)

In Eqs 47, 48, \hat{i}_{gd} and \hat{i}_{gq} represent steady-state harmonic currents. These ripples distort the steady-state grid current waveform. Similarly, reference voltage in the steady state can be simplified as follows:

$$U_d^* = \Delta U_d + U_{gd} - wLi_{gq} - wL\hat{i_{gq}}, \qquad (49)$$

$$\boldsymbol{U}_{\boldsymbol{q}}^{*} = \Delta \boldsymbol{U}_{\boldsymbol{q}} + \boldsymbol{U}_{\boldsymbol{g}\boldsymbol{q}} - \boldsymbol{w}\boldsymbol{L}\boldsymbol{i}_{\boldsymbol{g}\boldsymbol{d}} - \boldsymbol{w}\boldsymbol{L}\,\widehat{\boldsymbol{i}_{\boldsymbol{g}\boldsymbol{d}}}.$$
 (50)

c) Due to the presence of the terms $-wL\hat{i_{gq}}$ and $-wL\hat{i_{gd}}$, the waveform of the i_g and U_g will be distorted, and hence, the quality of the power will be worsened.

$$\mathbf{i}_{gd} = \mathbf{i}_{gd}^* + \hat{\mathbf{i}}_{gd}, \tag{51}$$

$$\mathbf{i}_{gq} = \mathbf{i}_{gq}^* + \hat{\mathbf{i}}_{gq}. \tag{52}$$

In **Eqs 51**, **52**, \hat{i}_{gd} and \hat{i}_{gq} represent steady state harmonic currents. These ripples distort the steady-state grid current waveform. Similarly, reference voltage in the steady state can be simplified as follows:

$$U_d^* = \Delta U_d + U_{gd} - wLi_{gq} - wL\hat{i_{gq}}, \qquad (53)$$

$$\boldsymbol{U}_{\boldsymbol{q}}^{*} = \Delta \boldsymbol{U}_{\boldsymbol{q}} + \boldsymbol{U}_{\boldsymbol{g}\boldsymbol{q}} - \boldsymbol{w}\boldsymbol{L}\boldsymbol{i}_{\boldsymbol{g}\boldsymbol{d}} - \boldsymbol{w}\boldsymbol{L}\,\widehat{\boldsymbol{i}_{\boldsymbol{g}\boldsymbol{d}}}.$$
 (54)

Due to the presence of the terms $-wL\hat{i_{gq}}$ and $-wL\hat{i_{gd}}$, the waveform of the grid voltage and current will be distorted, and hence, the quality of the power will be worsened.

4 RESULTS ANALYSIS AND DISCUSSION 4.1 Validation Through Variation in Solar PV Power

The specific range of solar irradiance of the solar PV module is shown in **Figure 13A**. The variation in irradiance was taken for 0.5 s in the MATLAB simulation tool. The fluctuation of the solar power irradiance is supposed to consist of substantial imbalances due to the intermittent nature of the local atmospheric circumstances.

The dynamic response of the voltage of the PV module is shown in **Figure 13B**. The PV generation characteristic is displayed in **Figure 13B** and the resulting distributed energy of the PV module through controlled VSI boosts output and reduction of unwanted harmonics in the load side. Furthermore, it is observed the distributed energy of the particular PV module is certainly impacted by the variation of solar irradiance and maximize their value up to 10 kW.

Figure 13C displays the relative response of the battery charging energy provided to the distribution grid *via* a dc-link connection. Active response of the analyzed method is obtained within the variations of the solar power. It is noticed that if the PI control scheme is investigated together with the battery, then the effective energy is given directly into the distributed main grid. Furthermore, when the SC is generally absent, the distributed generated energy into the main grid fluctuates substantially.

4.2 Variation in the Capacity of SC

These subsections investigating the effect of different sizes of SC are used for stability evaluation. In **Figure 14A**, the relative dynamic response of the stored energy is given to distributed grid for three different ratings of SC. Three different



supercapacitors with a capacity of 60/30/95 kW are typically used in the stability analysis of the HPGS.

From the simulation results, validation of response characteristics illustrated in **Figures 14B,C**, it has resulted that at the 50 Hz constant frequency, under RL and without load conditions, the proposed HPGS generates an overrated RMS voltage of 960 V or 1.02 kV (peak). Now, the simulated loads are added at PCC through a DC bus system.

Here in **Figure 14C**, we can examine that specific voltage drops in the HPGS demonstrated at 0.25–0.6 s. After 0.6 s duration, the load will be removed, and we have noticed that the reduced rated voltage is now stabilized.

From **Figure 14D**, the stability of a grid-connected HPGS has been successfully investigated at no load; also, during 3–4 s, I waveforms tend to stabilize. To reduce the fluctuation of a connected grid system, a supercapacitor-based energy storage technology was deployed. Furthermore, a d-q control technique is proposed for maintaining the power flow balance and obtaining maximum power from an HPGS coupled to the grid system. The d-q based control scheme is proposed to suppress power variation from the distributed energy resources to the distribution grid, while maximum energy extracts were obtained through a P&O-based MPPT controller. The results presented for the power quality are considerably higher by enabling the integration of an effective distributed HES approach operated by the PI controller.

To stabilize the current drop phenomena, the proposed RL load is presented to the HPGS referred from Figure 14E. It was

observed that SC was integrated with DC bus along with RL load at 0.67 s, as depicted from **Figure 14F**. When SC is placed into the HPGS along with sudden RL load, the proposed system maintained their current and frequency values at rated values of 425 V (RMS value) or 590 V (peak value). Furthermore, these results show that the optimal placement of SC by VSI voltage is controlled, even though the DC bus voltage is controlled. It is depicted that, in the current response, roughly transients are there because of switching time in the SC. Thus, nevertheless, of these slight transients, the simulated results show a stability enhancement approach in the performance of the proposed HPGS.

4.3 Power Quality Analysis of the Proposed HPGS

The grid-side harmonics of the proposed HPGS model is reduced, that is, 1.06% THD. **Figure 15A** shows the output power was ensured comparable with IEEE standards 519–2014 (Kanjiya et al., 2015). Since the voltage of the DC buses was modified by modifying the resonance frequency using a PI controller, it should also be specified that the recommended converter has the highest converter efficiency for boost operation.

The FFT algorithm based THD of the V and I in the proposed simulated HPGS as shown in **Figures 15B,C** under the RL load scenario was examined as per the IEEE THD range (<5%) (Kanjiya et al., 2015).

Simulated results of the proposed PI control scheme conclude that the SC enables the HPGS to achieve the optimal

implementation and limits reactive power at a steady state. The PI controller was utilized to achieve the active power flow through the DC link in the HPGS and stored SC energy to maintain the load level. The PI controller uses different modes to get optimal results from the HPGS.

5 CONCLUSION

This article has successfully contributed to the stable dynamic responses of SC and validated the power quality results for the proposed HPGS model. The designed wind PMSG coupled with a suitable solar PV module which supplied energy to an RL load through the controlled application of a VSI for DC/AC, a battery storage subsystem, a DC/DC boost converter, and an LCL based filter for harmonic elimination. The intermittent characteristics of wind speed and solar irradiation obtained from sunlight have no adverse effect over the integrated load and fed continuous stable load voltage and current profiles associated through the DC/AC inverter. During modeling eigenvalue analysis of the hybrid system, it was established that the designed SC has controlled damping oscillations and supplemented the stability of HPGS under the influence of sudden load changing operating conditions. The proposed dc-link voltage control strategy has improved the active power flow with the utilization of SC for HPGS under the influence of the different grid scenarios (weak/ sudden load change). This article has analyzed the optimal oscillation control approach to enhance the dynamic grid performance and power quality control of HPGS under a scenario of a weak grid. Simulated results of HPGS have proven the effective utilization of the HPGS under a steadystate scenario. Analysis with the comparison of obtained results from the PI controller approach reveals that the transient/ dynamic responses have better flexible controllability. The dclink voltage controller can be robust for the wide range of grid voltage and impedance control under multi microgrid systems. The proposed novel controller approach is very efficient in

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controlling the constant bus voltage and fed ripples free power under the influence of sudden disturbances. So forth, the tuning of the proposed control scheme can be improved further to maintain the quality of power being supplied to AC loads integrated with remote control microgrid. In a future study, the real-time implementation of the HPGG model and its dynamic performance analysis to evaluate the optimal sizing of the HPGS with an intelligent controller (AL-Jumaili et al., 2021) scheme for the smart microgrid systems.

DATA AVAILABILITY STATEMENT

The original contributions presented in the study are included in the article/supplementary material; further inquiries can be directed to the corresponding authors.

AUTHOR CONTRIBUTIONS

Conceptualization and writing original draft; MA: formal analysis and editing; AP: investigation and supervision; SR. All authors listed have made a substantial, direct, and intellectual contribution to the work and approved it for publication.

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