

## Robust State Estimation Method for Power System Based on Maximum Exponential Square and Data Fusion

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In order to solve the problem of the time offset between the supervisory control and data acquisition system and phasor measurement unit and the unknown distribution of non-Gauss noise, this paper proposes a robust state estimation method for power systems based on the Maximum Exponential Square and data fusion. Firstly, the robust Mahalanobis distance is used to detect system outliers and assign appropriate weights to the selected PMU buffer measurement. Then, the MES-based estimator uses these weights to filter out non-Gauss PMU measurement noise to generate a set of state estimation results. At the same time, the MES estimator is used to process the received SCADA measurement with unknown measurement noise, thereby generating another set of state estimation results. Finally, the two sets of estimation results from two independent MES estimators are fused by using data fusion theory to obtain the final optimal state estimation results. Based on IEEE-14 and 30-buses standard system, the simulation results prove the effectiveness and robustness of the method proposed in this paper.

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## **1 INTRODUCTION**

State Estimation technology can be used to improve the state awareness and defense against bad data in smart grid monitoring, reliability analysis and optimal load distribution, etc. (Liu Y et al., 2021). Reliable data foundation is the cornerstone of other advanced applications (Alnowibet et al., 2021). The purpose of SE is to obtain the best estimate of the current system node voltage modulus and phase angle given a set of redundant measurements and accurate network topology and parameters (Ma et al., 2021). Therefore, the performance of SE is highly dependent on the accuracy of the measurement and the estimation model established under the assumptions, and reasonable assumptions are crucial to the establishment of the model (Wang H et al., 2021). In SE, measurements can be affected by the errors of the measuring instruments, and even serious errors can occur due to instrument failures, impulse communication noise, etc. (Yu et al., 2021) Due to changes in the environment and temperature, the network parameters will also change over time, thus introducing uncertainty into the assumed SE model (Zanni et al., 2020). In addition, in the absence of sufficient field information or lack of calibration, the zero-injection information of the contact node may also be wrong (Zhao and Mili, 2018). Therefore, the measurement noise is unknown in the actual power system, and the uncertainty is strong, not the traditional Gauss noise.

However, in the existing research on power systems, measurement noise is often assumed to be Gaussian distribution, but actual measurement statistics show that measurement noise often deviates

from the assumed Gauss model, resulting in outliers. In this case, the performance of existing SE methods, which rely on the Gauss assumption that measurement noise is Gaussian, may be significantly degraded (Zhao and Mili, 2018). With the widespread deployment of PMUs, PMUs can provide more and more simultaneous voltage and current phasor measurements with time stamps and higher accuracy than traditional SCADA measurements, so are often used to improve the accuracy of SE as well as enhance Detection and identification of bad data. However, the sampling frequencies of the measurements from the SCADA system and the PMU are very different, leading to time offset issues, which must be unified under the same time profile (Thomas and Mili, 2007; Gandhi and Mili, 2010; Zhang et al., 2013; Murugesan et al., 2015; Zhao et al., 2015; Le et al., 2020).

To address this issue, PMU measurements can be buffered and hypothesis testing used to select the optimal buffer length (Thomas and Mili, 2007; Gandhi and Mili, 2010; Zhang et al., 2013; Murugesan et al., 2015; Zhao et al., 2015; Le et al., 2020). The sample mean and covariance of these buffered PMU measurements are then combined with SCADA measurements for SE. For PMU measurement, it is generally assumed that its error follows Gauss distribution. However, the latest statistical research shows that PMU measurement noise follows heavytailed distribution rather than Gaussian distribution (Zhao and Mili, 2018). Based on this, the literature (Zhao and Mili, 2018) improves the traditional static SE in the case of non-Gauss noise, but this literature does not consider the time offset problem between SCADA and PMU measurements.

Data Fusion theory is a relatively new area in the fields of aerospace, information theory, and signal processing, dealing with processes or states monitored by different classes of sensors and performing information fusion. Each class of sensors may have common characteristics, such as underlying technology, level of accuracy, etc., and then combine the data generated by the various sensors in an optimal way, in a sense, relative to the information obtained from a single sensor; the resulting final fused information quality is improved. Taking SCADA and phasor measurements as different sources of information for monitoring the state of the same power grid, and applying data fusion technology to integrate PMU measurements into state estimation, SCADA and phasor data can be fused to improve the quality of the fusion data (Zheng, 2015; Li et al., 2021).

To improve the monitoring state awareness and deal with unknown non-Gauss noise and time offset between SCADA and PMU measurements, this paper proposes a robust SE method for power systems based on MES and data fusion. In this method, robust Mahalanobis distances are utilized to detect system anomalies and assign appropriate weights to each selected PMU buffering measure. These weights are further exploited based on the MES estimator to filter out non-Gauss PMU noise and help suppress outliers. At the same time, the MES estimator is also used to process the received SCADA measurements with unknown measurement noise, resulting in another set of SEs. Finally, the effectiveness and robustness of the proposed method are verified based on IEEE-14 and 30-node standard systems. The contributions of this paper are as follows: embed the zeroinjection expression in the MES estimation model to eliminate the equality constraints, which can improve the efficiency of the algorithm; calculate the PMU buffer based on the robust Mahalanobis distance, and unify the PMU and SCADA data to a unified cross-sectional scale, which can eliminate the time delay problem; use the state fusion based on data fusion theory which can obtain the optimal state estimation result.

#### **2 MEASUREMENT EQUATION**

By the Wiener approximation theorem, the non-Gauss distribution p(x) can be well approximated by the known Gaussian distribution, so the following model can be used to simulate the non-Gaussian distribution error:

$$p(x) = \sum_{i=1}^{N_A} a_i \Phi\left(\bar{x}_i, \sum_i\right) \tag{1}$$

Based on whether the known measurement error obeys the Gauss distribution, the above formula can be further expressed as follows (Zhao and Mili, 2018):

$$G(e) = (1 - \varepsilon)\Phi(e) + \varepsilon K(e)$$
(2)

where  $\Phi(e)$  is the majority distribution of measurement noise, usually modeled as Gauss distribution; K(e) is the unknown distribution, which is considered to be a heavy-tailed density, such as a Laplace density or Gauss density with large variance;  $0 \le e \le 0.5$ , contamination coefficient, which adjusts the contribution of non-Gauss components, e.g. for small  $\varepsilon$ , the model indicates that most of the errors follow a Gauss distribution, while maintaining a small fraction of non-Gauss errors.

The measurement equation under non-Gauss noise can be expressed as follows:

$$z = h(x) + G(e) \tag{3}$$

# 3 PMU OPTIMAL BUFFER LENGTH CALCULATION

Since the sampling frequency of the PMU is much higher than that of the SCADA measurement, there is a time skew problem for the PMU and SCADA measurements. In the existing research, such as the literature (Zhang et al., 2013) and (Murugesan et al., 2015), it is proposed to use the hypothesis test method to select the optimal buffer length of the PMU measurement between two SCADA measurement scans, but this is only for a single PMU measurement point, therefore, it is necessary to calculate each measuring point one by one, which is inefficient.

In this paper, based on Mahalanobis distance, the optimal buffer length of multiple PMU measuring points is determined (Zhao and Mili, 2018), and all PMU measuring points are taken as a whole.

Let the PMU sample  $n_r$ =30 samples/s per second, the SCADA sampling interval  $N_t = 5s$ , in a interval, the PMU sampling number  $N = N_t * n_r$ =150 samples, use Z to represent the sampling matrix, and divide it into  $n_{subset} = N_t = 5$  subsets;  $Z = [Y_1, Y_2, Y_3, Y_4, Y_5]$ , PMU Optimal Buffer Length calculated based on the following steps:

(1) Calculate the median value of the Z matrix:

$$Y' = median(z) = \left[y'_{1}, y'_{2}, y'_{3}, y'_{4}, y'_{5}\right]$$
(4)

- (2) Perform change point detection on  $Y_5$ , if there is a change point, use the last PMU data; otherwise,  $Y_5$  will be included in the PMU buffer set;
- (3) Based on the Mahalanobis distance, the system changes are detected for  $\eta_1 = [y'_5, y'_4], \eta_2 = [y'_4, y'_3], \dots, \eta_4 = [y'_2, y'_1]$ :

$$P_{i} = \max\left(\frac{\left|y_{i}^{T}v - median(y_{j}^{T}v)\right|}{\lambda \cdot median(\left|y_{k}^{T}v - median(y_{j}^{T}v)\right|)}\right), \|v\| = 1 \quad (5)$$

If all of Mahalanobis distance  $P_i$  in  $\eta_1$  are less than  $\zeta_1$ ,  $\zeta_1 = \chi^2_{2,0.975}$ , then contains  $Y_4$  in buffing set Z', otherwise the algorithm stops; and continue to check  $\eta_2 = [y'_4, y'_3], \ldots, \eta_4 = [y'_2, y'_1]$ , until the end.

(4) Calculate the measurement mean  $\bar{h}$  and variance  $\bar{C}$  in the PMU buffer set:

$$\bar{C} = \frac{\sum_{i=1}^{\alpha} (w_{i1}h_i - \bar{h}) (w_{i1}h_i - \bar{h})^T}{\sum_{i=1}^{\alpha} w_{i1} - 1}$$
(6)

$$\bar{h} = \frac{\sum_{i=1}^{\alpha} w_{i1} h_i}{\sum_{i=1}^{\alpha} w_{i1}}$$
(7)

Based on the above steps, the optimal buffer length of multiple PMU measuring points can be obtained. After calculating the mean  $\bar{h}$  and variance  $\bar{C}$ , it can be matched with the SCADA data under the same section.

## 4 ROBUST SE METHOD BASED ON MES AND DATA FUSION

#### 4.1 MES-Based SCADA Robust SE

Based on the MES estimation method, the MES estimation model is established by using SCADA measurement, and its expression is as follows:

$$\begin{cases} \max J_1(x) = \sum_{i=1}^m w_i \exp\left(-\frac{[z_i - h_i(x)]^2}{2\sigma^2}\right) \\ s.t.c(x) = 0 \end{cases}$$
(8)

The zero injection constraint in polar coordinates can be expressed as:

$$\begin{cases} x_B = \Psi(x_N) \\ \Psi = \Phi \Gamma \Phi^{-1} \end{cases}$$
(9)

 $\Gamma$  is:

$$\Gamma = -\begin{bmatrix} Y_{\mathrm{Im},BB} & Y_{Re,BB} \\ Y_{Re,BB} & -Y_{\mathrm{Im},BB} \end{bmatrix}^{-1} \begin{bmatrix} Y_{\mathrm{Im},BN} & Y_{Re,BN} \\ Y_{Re,BN} & -Y_{\mathrm{Im},BN} \end{bmatrix}$$
(10)

 $\Phi$  is:

$$\begin{cases} u_k = \sqrt{e_k^2 + f_k^2} \\ \theta_k = \arctan\left(\frac{f_k}{e_k}\right) \end{cases}$$
(11)

 $\Phi^{-1}$  is:

$$\begin{cases} e_k = u_k \cos\left(\theta_k\right) \\ f_k = u_k \sin\left(\theta_k\right) \end{cases}$$
(12)

Substituting this constraint into the MES estimation model, we get:

$$\max J_{1}(x_{N}) = \sum_{i=1}^{m} w_{i} \exp\left(-\frac{\left[z_{i}-h_{i}(x_{N}, \Phi F \Phi^{-1}(x_{N}))\right]^{2}}{2\sigma^{2}}\right)$$
(13)

make:

$$r_{wi} = -\frac{[z_i - h_i(x)]}{\sqrt{2}\sigma} \tag{14}$$

$$f_i(x) = \exp\left(-r_{wi}^2\right) \tag{15}$$

Using Newton's method to solve, the modified equation is obtained:

$$\Delta x^{k} = \begin{bmatrix} \Delta x_{B}^{k} \\ \Delta x_{N}^{k} \end{bmatrix} = -QH^{T}F(z - h(x_{N}^{k}, \Phi\Gamma\Phi^{-1}(x_{N}^{k})))$$
(16)

Where F and Q are:

$$F(x) = diag\{F_{ii}(x)\} = diag\left\{\frac{w_i f_i(x)}{\sigma^2}\right\}$$
(17)

$$Q = -H^{T}F\left[I - diag\left(\frac{\left[z - h(x)^{2}\right]}{\sigma^{2}}\right)\right]H$$
(18)

The state quantity is corrected to:

$$x_N^{k+1} = x_N^k + \Delta x_N^k \tag{19}$$

$$x_B^{k+1} = \Phi F \Phi^{-1} \left( x_N^{k+1} \right)$$
 (20)

$$\mathbf{x}^{k+1} = \begin{bmatrix} \mathbf{x}_{B}^{k+1} \\ \mathbf{x}_{N}^{k+1} \end{bmatrix}$$
(21)

The iterative calculation flow of estimation based on MES is shown in **Figure 1**:

The calculation flow of MES estimator is as follows:

- (1) Initialization parameters, k = 0, set the initial value of  $x_{k,N}$
- (2) Calculate  $x_{k,B}$  from  $x_{k,N}$  according to Formula (20)
- (3) Calculate the matrices H, F and Q, solve Eq 16, get  $\Delta x_N^k$ , update  $x_N^{k+1} = x_N^k + \Delta x_N^k$ , and calculate  $x_B^{k+1}$
- (4) If max |Δx<sub>k</sub>| < ε, the state estimation is converged and the calculation is over, otherwise go to step (3).</p>



The state error variance matrix of SCADA-based MES estimation is:

$$\sum_{s} = E\left[ (x - \hat{x})(x - \hat{x})^{T} \right] = \left[ H^{T}(x)F(x)H(x) \right]^{-1}$$
(22)

It should be noted that this paper embeds the zero-injection expression in the MES estimation model, so that the equality constraints do not have to be explicitly expressed, and the equality constraints are eliminated. It is not necessary to use Lagrange multiplier method to improve the efficiency of the algorithm, which is one of the main tasks in this paper.

#### 4.2 MES-Based PMU Robust SE

Similar to SCADA robust SE, PMU measurement alone is used for robust estimation based on MES model (Min et al., 2021; Wang Q et al., 2021), but PMU measurement is a voltage and current phasor, and its measurement function is a linear function:

$$h = Ax + \varepsilon \tag{23}$$

Based on the MES model, we can get:

$$\max J_{1}(x_{N}) = \sum_{i=1}^{m} w_{i} \exp\left(-\frac{\left[z_{i} - \bar{h}(x_{N}, \Phi F \Phi^{-1}(x_{N}))\right]^{2}}{2\sigma^{2}}\right)$$
(24)

The Newton method is also used to solve the equation, and the modified equation is obtained:

$$\Delta x^{k} = \begin{bmatrix} \Delta x_{B}^{k} \\ \Delta x_{N}^{k} \end{bmatrix} - QA^{T}F(z - h(x_{N}^{k}, \Phi F \Phi^{-1}(x_{N}^{k})))$$
(25)

The state quantity correction is the same as formulas (19)-(21).

After iterative convergence, the state error variance matrix of the PMU-based MES estimation is:

$$\sum_{p} = E\left[\left(x - \hat{x}\right)\left(x - \hat{x}\right)^{T}\right] = \left[A^{T}F(x)A\right]^{-1}$$
(26)



# 4.3 State Fusion Based on Data Fusion Theory

In this paper, referring to data fusion theory, SCADA and PMU measurements are regarded as different data sources and their measurements are represented by  $z_S$  and  $z_P$ , respectively. The SCADA-based and PMU-based SE devices are called SSE and PSE, respectively. After processing  $z_S$  and  $z_P$  respectively, two independent SE results  $\hat{x}_s$  and  $\hat{x}_p$  can be generated, and the corresponding error variance matrix is represented by  $\sum_s$  and  $\sum_{p}$ .

<sup>7</sup> Based on a, b, c and d, the best state fusion result is obtained by data fusion:

$$\hat{x} = \frac{\sum_{p}}{\sum_{p} + \sum_{s}} \hat{x}_{s} + \frac{\sum_{s}}{\sum_{p} + \sum_{s}} \hat{x}_{p}$$
(27)

As can be seen from the above formula, the denominator of the weight of the SE of  $x_s$  and  $x_p$  is the same, and the numerators are  $\sum_p$  and  $\sum_s$  respectively. It means that the size of the weight is inversely proportional to the variance of the estimation result, that is, the estimation result with higher precision has a correspondingly larger weight, and the estimation result with lower precision has a correspondingly smaller weight. The above estimation method is called the data fusion SE method.

Combining data fusion with MES estimation model is the second main work of this paper.

The schematic diagram of the fusion SE calculation logic is shown in Figure 2.





## **5 CASE SIMULATION AND ANALYSIS**

#### 5.1 Basic Data and Simulation Conditions

In order to verify the performance of the robust SE method based on MES and data fusion proposed in this paper, simulations are carried out based on the IEEE-14 node and IEEE-30 node standard system, and the topology diagrams are shown in **Figures 3**, **4**, respectively.

For the IEEE-14 node standard system, three PMUs were deployed on nodes 4, 7, and 14, and 31 SCADA measurements were considered, including 16 branch power, 14 injected power, and one node voltage amplitude. For the IEEE-30 node standard system, 38 pairs of power measurements are considered, including 15 pairs of active and reactive power injection and 23 pairs of branch active and reactive power measurements; active and reactive power injection at nodes 11, 12, 24,27,30, and active and reactive power at branches 24-23, 25-26, 30-27 are key measurements, six PMUs are deployed on nodes 8, 9, 12, 24, 25, and 26.

At each layout point, PMU provides the voltage of the layout point and the branch current vector associated with the layout point; SCADA provides the voltage amplitude of the layout point, injected active power, reactive power and the branch active power, reactive power associated with the layout point.

- 1) Hybrid data WLS estimator (ie HWLS) (Li et al., 2013), which combines buffered PMU measurements with SCADA measurements for SE.
- 2) Two-stage WLS estimator (ie TWLS) (Liu X et al., 2021), which first performs traditional SE based on SCADA measurements, and its SE estimation results are further combined with PMU measurements to perform secondstage linear SE.

Using the MAE of the voltage modulus value and the angle estimate value as the overall performance evaluation index, the algorithm relationship between the method proposed in this paper and the HWLS and TWLS methods is shown in **Figure 5**.

# **5.2 Simulation Results and Analysis** 5.2.1 Under Gaussian Noise

First consider adding known Gauss noise to SCADA and PMU measurements to better understand the best performance of each method (Al-Saud et al., 2019; Mohamed et al., 2019).

For this, the noise of each SCADA and PMU measurement is assumed to be Gaussian random variables with zero mean and variance of  $2 \times 10^{-4}$  and  $2 \times 10^{-6}$  respectively. The mean absolute errors of the voltage modulus and angle estimated by each method are shown in **Tables 1**, **2**, respectively.

It can be seen from **Tables 1**, **2** that under the Gauss measurement noise, for the IEEE-14 node standard system, the voltage modulus value and the MAE of the phase angle of the HWLS are  $5.01 \times 10^{-4}$  and  $3.94 \times 10^{-2}$ , respectively. The voltage modulus of TWLS and the MAE of the phase angle are  $4.52 \times 10^{-4}$  and  $3.55 \times 10^{-2}$ , respectively. The voltage modulus value and the phase angle MAE of the method proposed in this paper are  $3.74 \times 10^{-4}$  and  $3.43 \times 10^{-2}$  respectively. For the IEEE-30 node standard system, the MAE of the HWLS voltage modulus and phase angle is  $3.77 \times 10^{-4}$  and  $2.71 \times 10^{-2}$ , respectively. The voltage modulus value and the phase angle of the HWLS voltage modulus and phase angle is  $3.77 \times 10^{-4}$  and  $2.71 \times 10^{-2}$ , respectively. The voltage modulus value and phase angle MAE of the HWLS are  $2.99 \times 10^{-4}$  and  $9.96 \times 10^{-3}$  respectively. The MAE of the voltage modulus and the phase angle of the method proposed in this paper are  $2.53 \times 10^{-4}$  and  $6.72 \times 10^{-3}$ , respectively.

Therefore, the method proposed in this paper is better than TWLS and outperforms the HWLS method. This is because both the proposed method and TWLS fully consider the buffered PMU measurement, which avoids the influence of SCADA with lower measurement accuracy on the PMU with higher measurement accuracy in the hybrid measurement, thereby improving the estimation accuracy.

To further evaluate the performance of various methods at another Gauss noise level. Assuming that the variance of each SCADA and PMU measurement is increased to  $6 \times 10^{-3}$  and  $6 \times 10^{-5}$ , respectively, the verification is performed based on the IEEE-30 node standard system, and the results are shown in **Table 3**.

It can be seen from **Table 3** that for HWLS, TWLS and the proposed method, the MAEs of the estimated voltage modulus values are  $1.57 \times 10^{-3}$ ,  $1.35 \times 10^{-3}$ , and  $1.21 \times 10^{-3}$ , respectively, and the MAEs of the estimated voltage phase angles are  $4.32 \times 10^{-2}$ ,  $3.33 \times 10^{-2}$  and  $1.94 \times 10^{-2}$ . It can be seen that the estimation error of all methods increases slightly as the noise level increases, because at a given measurement redundancy level, the estimation statistical efficiency will decrease if the noise level increases. However, the proposed method still outperforms the other two HWLS and TWLS methods.

#### 5.2.2 Under Non-Gaussian Noise

In practical applications, the noise statistics of SCADA and PMU measurements are usually unknown and deviate from Gauss assumptions due to the aging process of voltage and current transformers, changes in ambient temperature, communication



<b>TABLE 1</b> Average absolute error of voltage amplitude under Gaussian noise (pu).			
Standard System	HWLS	TWLS	Proposed method
EEE-14 node	5.01 × 10 <sup>-4</sup>	4.52 × 10-4	3.74 × 10 <sup>-4</sup>
IEEE-30 node	3.77 × 10 <sup>-4</sup>	2.99 × 10-4	$2.53 \times 10^{-4}$

Standard System	HWLS		TWLS	Proposed met
noise (pu).				
TABLE 4   Average al	bsolute error	of voltage a	amplitude	under non-Gaussian

Standard System	HWLS	TWLS	Proposed method
IEEE-14 node	9.41 × 10-3	8.37 × 10-3	6.75 × 10-3
IEEE-30 node	4.91 × 10-3	1.95 × 10-3	1.42 × 10-3

<b>TABLE 2</b> Average absolute error of voltage phase angle under Gaussian noise (°).				
Standard System	HWLS	TWLS	Proposed method	
IEEE-14 node	3.94 × 10 <sup>-2</sup>	3.55 × 10-2	$3.43 \times 10^{-2}$	
IEEE-30 node	2.71 × 10 <sup>-2</sup>	9.96 × 10-3	$6.72 \times 10^{-3}$	

**TABLE 3** | Average absolute error of voltage under Gaussian noise in three methods.

Standard System	HWLS	TWLS	Proposed method
Modulus value (pu)	1.57 × 10 <sup>-3</sup>	1.35 × 10 <sup>-3</sup>	1.21 × 10 <sup>-3</sup>
Phase angle (°)	4.32 × 10 <sup>-2</sup>	3.33 × 10 <sup>-2</sup>	1.94 × 10 <sup>-2</sup>

channel noise, etc. To evaluate the performance of each method against unknown statistical noise, non-Gauss noise models, namely Gauss mixture model and heavy-tailed Laplacian noise, were used for simulation. Specifically, for SCADA **TABLE 5** | Average absolute error of voltage phase angle under non-Gaussian noise (").

Standard System	HWLS	TWLS	Proposed method
IEEE-14 node	0.51	0.39	0.18
IEEE-30 node	0.34	0.27	0.13

measurements, a Gaussian mixture model with two Gauss components (non-Gaussian after mixing) is assumed, which is represented by zero mean and variance  $2 \times 10^{-4}$  and  $2 \times 10^{-3}$  with weights of 0.95 and 0.05, respectively. For the PMU measurement, the variances are assumed to be  $2 \times 10^{-6}$  and  $2 \times 10^{-4}$ , and the weights are 0.9 and 0.1. In the second case, a Laplacian distribution with zero mean and scale parameter  $10^{-3}$  is used for the noise of SCADA and PMU measurements.

Tables 4, 5 present the MAE results for the voltage modulo values and angles estimated by each method in the presence of





non-Gauss measurement noise simulated by the Gauss hybrid model, respectively. Under the Gauss mixed model noise, for the IEEE-14 node standard system, the voltage modulus value and phase angle MAE of HWLS are  $9.41 \times 10^{-3}$  and 0.51, respectively, and the voltage modulus value and phase angle MAE of TWLS are  $8.37 \times 10^{-3}$  and 0.39, the MAE of the voltage modulus and the phase angle of the method proposed in this paper are  $6.75 \times 10^{-3}$  and 0.18, respectively. For the IEEE-30 node standard system, the voltage modulus value and phase angle MAE of HWLS are  $4.91 \times 10^{-3}$  and 0.34, respectively, and the voltage modulus value and phase angle MAE of TWLS are  $4.91 \times 10^{-3}$  and 0.34, respectively, and the voltage modulus value and phase angle MAE of TWLS are  $1.95 \times 10^{-3}$  and 0.27, respectively. The voltage modulus value and the phase angle MAE of TWLS are  $1.95 \times 10^{-3}$  and 0.27, respectively.



FIGURE 8 | The absolute error of the voltage amplitude under different cases.



Compared with the Gauss noise cases in **Tables 1**, **2**, the estimation errors of all three methods are increased. But HWLS and TWLS have a larger increase in estimation error because they rely heavily on the Gauss assumption of SE. In contrast, the robust estimator proposed in this paper based on MES and data fusion is more robust and produces more reasonable SE results.

**Figures 6**, 7 show the absolute error values of HWLS, TWLS and the proposed method under Laplace measurement noise. As can be seen from the results in the figures, the results are in complete agreement with those shown in **Tables 4**, **5**. Therefore, it can be concluded that the robust estimator based on MES and data fusion proposed in this paper does not require the measurement noise to be Gauss noise, and has better robustness in dealing with non-Gauss noise.

#### 5.2.3 Robustness to Bad Data

In order to verify the robustness of the robust estimator proposed in this paper to various types of bad data, in the presence of bad data for bad leverage points, bad key measurements, and measurements of multiple interactions, based on the IEEE-30 node standard system simulation.

Set up the following four simulation schemes:

Scheme 1: In the case of a single bad data, change the active power injection into P8 of Section 8 to 0.1pu.

Solution 1: For bad lever measurement, change P19 and P19-20 to 0.1pu.

Solution 2: For bad key measurement, change P11 to 0.2pu. Scheme 3: For bad data affected by multiple interactions, change P2 and Q2 to the original opposite numbers.

**Figure 8, 9** show the simulation results of different types of bad data under the four schemes. Obviously, since the robustness of the robust estimator based on MES and data fusion proposed in this paper is provided by the robust Mahalanobis distance and MES estimator. The influence of various types of bad data is bounded, resulting in a very small estimation deviation. The maximum deviation of the voltage modulus value estimation is 0.00094pu, and the maximum deviation of the voltage phase angle estimation is 0.0075°, which can be ignored in practical applications.

## **6 CONCLUSION**

To solve the problem of unknown non-Gauss noise and time offset of SCADA and PMU measurements, this paper proposes a robust SE method for power systems based on MES and data fusion based on data fusion theory and MES estimation model. This method has the following characteristics:

- The zero injection expression is embedded in the MES estimation model, so that the equality constraints do not need to be expressed explicitly, and the equality constraints are eliminated, and the Lagrange multiplier method is no longer required to process, thereby improving the efficiency of the algorithm;
- (2) PMU buffer measurement based on robust Mahalanobis distance selection, and assign weights, further based on MES SE method, which can filter non-Gauss measurement noise and outliers;
- (3) Based on MES and data fusion theory, perform SE for PMU and SCADA measurements with unknown measurement noise.

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Al-Saud, M., Eltamaly, A. M., Mohamed, M. A., and Kavousi-Fard, A. (2019). An Intelligent Data-Driven Model to Secure Intravehicle Communications Based on Machine Learning. *IEEE Trans. Ind. Electron.* 67 (6), 5112–5119. doi:10. 1109/TIE.2019.2924870 Based on the two sets of estimation results, the estimation results from two independent MES estimators are fused by using data fusion theory to obtain the final optimal SE.

Based on the IEEE-14 and IEEE-30 node standard systems, the simulation results show that the proposed method is effective and robust, and has obvious advantages compared with the existing methods.

## DATA AVAILABILITY STATEMENT

The raw data supporting the conclusion of this article will be made available by the authors, without undue reservation.

## AUTHOR CONTRIBUTIONS

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## NOMENCLATURE

Supervisory control and data acquisition system SCADA
Phasor measurement unit PMU
State Estimation SE
Data Fusion DF
Maximum Exponential Square MES
Mean absolute error MAE
Variables
The weight of measurement $(\sum_{i=1}^{N_A} a_i = 1) a_i$
The number of Gauss elements $N_A$
The mean of measurement $\bar{x}_i$
The variance of measurement $\sum_i$
The non-Gauss distribution $p(x)$
The Gauss distribution $\Phi(\bar{x}_i, \sum_i)$
The unknown distribution $K(e)$
Contamination coefficient $\varepsilon$
The measurement z
The nonlinear measurement function $h(x)$
The measurement error $G(e)$
Mahalanobis distance P <sub>i</sub>

The buffing set Z' The number of columns of  $Z' \alpha$ The weight of the *i*th column of Z',  $w_{i1} = \min(1, \frac{\varsigma_2}{P^2}), \ \varsigma_2 = \chi^2_{\alpha, 0.975} \ w_{i1}$ The measurement of  $Z' h_i$ The state quantity x The measurement function  $h_i(x)$ The *i*th measurement weight  $w_i$ The *i*th measurement  $z_i$ The Parzen window width  $\sigma$ The zero injection node constraint c(x) = 0The zero injection power node B The non-zero injection node N The transformation relationship between  $x_B$  and  $x_N \Psi$ The measurement Jacobian matrix H The mean value of the measurement in the PMU buffer set  $\bar{h}$ The relationship matrix between the PMU and the state quantity A The PMU measurement error variance in the PMU buffer set  $\bar{C}$